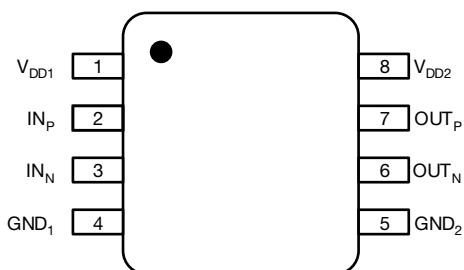


## High Thermal Stability Isolation Amplifier



### LINKS TO ADDITIONAL RESOURCES


[Product Page](#)

[Application Notes](#)

[3D Models](#)

[SPICE Models](#)

[Ultra Librarian EDA/CAD](#)

[Design Tools](#)

### AGENCY APPROVALS

- UL
- cUL
- DIN EN 60747-5-5 (VDE 0884-5)
- CQC

### DESCRIPTION

The VIA0250DD is a high performance differential output isolated amplifier ideally suited for shunt based current sensing. The device is based on proprietary capacitive isolation technology and has a linear differential input signal range of  $\pm 250$  mV.

The low offset drift typ.  $0.15$   $\mu$ V/ $^{\circ}$ C and low gain drift typ.  $15$  ppm/ $^{\circ}$ C ensures a good accuracy over the entire temperature range. The devices unmatched CMTI of  $100$  kV/ $\mu$ s allows accurate measurements in the noisy environment.

### FEATURES

- Isolation test voltage:  $5000$  V<sub>RMS</sub>
- $\pm 250$  mV linear input voltage range
- Fixed gain: 8.2
- gain error and drift:  $\pm 0.05$  % typ.,  $\pm 15$  ppm/ $^{\circ}$ C typ.
- Low non-linearity and drift:  $\pm 0.03$  % max.,  $\pm 1$  ppm/ $^{\circ}$ C typ.
- SNR: 72 dB (typ., BW = 100 kHz)
- Wide bandwidth: 250 kHz
- High CMTI: 150 kV/ $\mu$ s typ.
- Inbuilt common-mode overvoltage detection
- Operating temperature:  $-40$   $^{\circ}$ C to  $+125$   $^{\circ}$ C
- Material categorization: for definitions of compliance please see [www.vishay.com/doc?99912](http://www.vishay.com/doc?99912)



**RoHS**  
COMPLIANT  
HALOGEN  
**FREE**  
**GREEN**  
(5-2008)

### APPLICATIONS

- Isolated current measurement in:
  - Motor control applications
  - Power supplies
  - Charging stations
  - EV powertrain

#### Note

- For automotive qualification please contact our local sales.

### DEVICE INFORMATION

PART NUMBER	PACKAGE	BODY SIZE
VIA0250DD	SOP-8 (300 mil)	5.85 mm x 7.50 mm

### ORDERING INFORMATION

PART NUMBER	ISOLATION RATING (kV)	LINEAR INPUT RANGE (mV)	MOISTURE SENSITIVITY LEVEL	TEMPERATURE (°C)	AUTOMOTIVE	PACKAGE TYPE	SPQ
VIA0250DD	5	-250 to +250	Level 3	-40 to +125	No	SOP-8 (300 mil)	1000

<b>ABSOLUTE MAXIMUM RATINGS</b> ( $T_{amb} = 25^{\circ}C$ , unless otherwise specified)			
PARAMETER	SYMBOL	VALUE	UNIT
Power supply voltage	$V_{DD1}, V_{DD2}$	-0.3 to 6.5	V
Input voltage	$IN_P, IN_N$	GND <sub>1-6</sub> to $V_{DD1+0.5}$	V
Output voltage	$OUT_P, OUT_N$	GND <sub>2-0.5</sub> to $V_{DD2+0.5}$	V
Output current per output pin	$I_o$	-10 to +10	mA
Operating temperature	$T_{amb}$	-40 to +125	$^{\circ}C$
Junction temperature	$T_j$	-40 to +150	$^{\circ}C$
Storage temperature	$T_{stg}$	-55 to +150	$^{\circ}C$
Electrostatic discharge	HBM <sup>(1)</sup>	$\pm 2000$	V
	CDM <sup>(2)</sup>	$\pm 1000$	V

**Notes**
<sup>(1)</sup> Human body model (HBM), per AEC-Q100-002-RevD

<sup>(2)</sup> Charged device model (CDM), per AEC-Q100-011-RevB

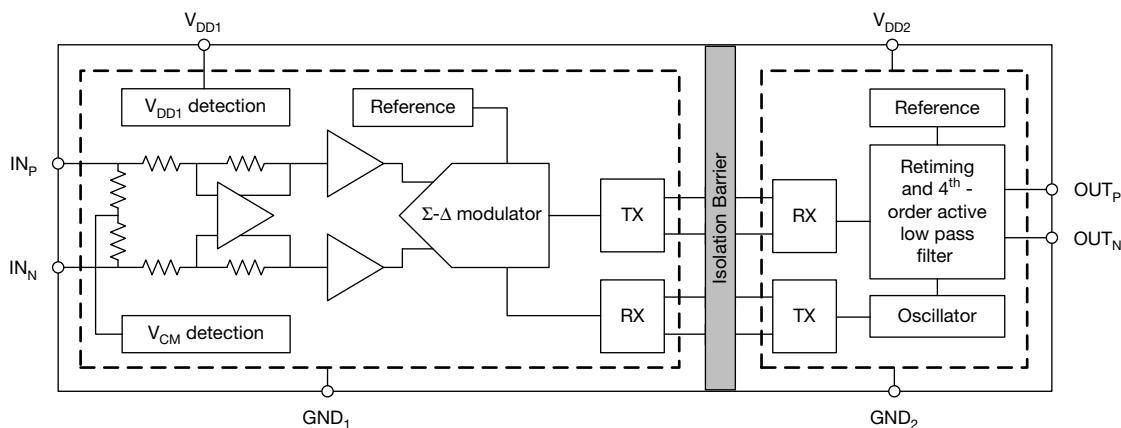
**FUNCTIONAL BLOCK DIAGRAM**


Fig. 1 - VIA0250DD Block Diagram

<b>RECOMMENDED OPERATING CONDITIONS</b> ( $T_{amb} = 25^{\circ}C$ , unless otherwise specified)					
PARAMETER	SYMBOL	MIN.	TYP.	MAX.	UNIT
Side 1 power supply	$V_{DD1}$	3.0	5.0	5.5	V
Side 2 power supply	$V_{DD2}$	3.0	3.3	5.5	V
VIA0250DD	Differential input voltage before clipping output	$V_{clipping}$	-	$\pm 320$	mV
	Linear differential input full scale voltage	$V_{FSR}$	-250	-	+250
	Operating common-mode input voltage	$V_{CM}$	-0.16	-	+0.8
Operating ambient temperature	$T_{amb}$	-40	-	+125	$^{\circ}C$

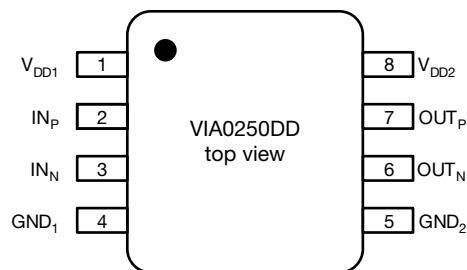
**PIN CONFIGURATION AND FUNCTIONS**


Fig. 2 - VIA0250DD Package

<b>PIN CONFIGURATION AND DESCRIPTION</b>		
<b>PIN NO.</b>	<b>SYMBOL</b>	<b>FUNCTION</b>
1	$V_{DD1}$	Power supply for isolator side 1 (3.0 V to 5.5 V)
2	$IN_P$	Positive analog input ( $\pm 250$ mV recommended for VIA0250DD)
3	$IN_N$	Negative analog input
4	$GND_1$	Ground 1, the ground reference for isolator side 1
5	$GND_2$	Ground 2, the ground reference for isolator side 2
6	$OUT_N$	Negative output
7	$OUT_P$	Positive output
8	$V_{DD2}$	Power supply for isolator side 2 (3.0 V to 5.5 V)

<b>ELECTRICAL CHARACTERISTICS: VIA0250DD</b> ( $V_{DD1}, V_{DD2} = 3 \text{ V to } 5.5 \text{ V}$ , $IN_P = -250 \text{ mV to } +250 \text{ mV}$ , $IN_N = GND_1 = 0 \text{ V}$ , $T_{amb} = -40 \text{ }^{\circ}\text{C to } +125 \text{ }^{\circ}\text{C}$ ( $V_{DD1} = 5 \text{ V}$ , $V_{DD2} = 3.3 \text{ V}$ , $T_{amb} = 25 \text{ }^{\circ}\text{C}$ , unless otherwise specified))						
<b>POWER SUPPLY</b>						
PARAMETER	TEST CONDITION	SYMBOL	MIN.	TYP.	MAX.	UNIT
<b>POWER SUPPLY</b>						
Side 1 supply voltage		$V_{DD1}$	3.0	5.0	5.5	V
Side 2 supply voltage		$V_{DD2}$	3.0	3.3	5.5	V
Side 1 supply current		$I_{DD1}$	-	11.4	15.1	mA
Side 2 supply current		$I_{DD2}$	-	6.3	8.4	mA
$V_{DD1}$ undervoltage detection threshold voltage	$V_{DD1}$ falling	$V_{DD1\_UV}$	1.8	2.3	2.7	V
<b>ANALOG INPUT</b>						
Common-mode overvoltage detection level	Detection level has a typical hysteresis of 96 mV	$V_{CMov}$	0.9	-	-	V
Input offset voltage	$IN_P = IN_N = GND_1$	$V_{OS}$	-0.2	$\pm 0.01$	+0.2	mV
Input offset drift		$TCV_{OS}$	-2	+1	+4	$\mu\text{V}/^{\circ}\text{C}$
Common-mode rejection ratio	$IN_P = IN_N$ , $f_{IN} = 0 \text{ Hz}$ , $V_{CM} \text{ min.} \leq V_{IN} \leq V_{CM} \text{ max.}$	$CMRR_{DC}$	-	-106	-	dB
	$IN_P = IN_N$ , $f_{IN} = 10 \text{ Hz}$ , $V_{CM} \text{ min.} \leq V_{IN} \leq V_{CM} \text{ max.}$	$CMRR_{AC}$	-	-106	-	dB
Single-ended input resistance	$IN_N = GND_1$	$R_{IN}$	-	19	-	$\text{k}\Omega$
Differential input resistance		$R_{IND}$	-	22	-	$\text{k}\Omega$
Input capacitance		$C_I$	-	2	-	$\text{pF}$
Input bias current	$IN_P = IN_N = GND_1$ , $I_{IB} = (I_{IBP} + I_{IBN})/2$	$I_{IB}$	-24	-18	-12	$\mu\text{A}$
Input bias current drift		$TCI_{IB}$	-	$\pm 1$	-	$\text{nA}/^{\circ}\text{C}$
<b>ANALOG OUTPUT</b>						
Nominal gain			-	8.2	-	V/V
Gain error		$E_G$	-0.3	$\pm 0.05$	+0.3	%
Gain error thermal drift		$TCE_G$	-50	$\pm 15$	+50	$\text{ppm}/^{\circ}\text{C}$
Non-linearity			-0.03	$\pm 0.01$	+0.03	%
Non-linearity drift			-	$\pm 1$	-	$\text{ppm}/^{\circ}\text{C}$
Total harmonic distortion	$V_{IN} = 500 \text{ mV}_{pp}$ , $f_{IN} = 10 \text{ kHz}$ , $BW = 100 \text{ kHz}$	THD		-85	-	dB
Output noise	$IN_P = IN_N = GND_1$ , $BW = 100 \text{ kHz}$		-	195	-	$\mu\text{V}_{RMS}$
Signal to noise ratio	$V_{IN} = 500 \text{ mV}_{pp}$ , $f_{IN} = 1 \text{ kHz}$ , $BW = 10 \text{ kHz}$	SNR	80	86	-	dB
	$V_{IN} = 500 \text{ mV}_{pp}$ , $f_{IN} = 10 \text{ kHz}$ , $BW = 100 \text{ kHz}$		-	72	-	dB
Common-mode output voltage		$V_{CMout}$	1.39	1.44	1.49	V
Fail-safe differential output voltage	$V_{CM} > V_{CMov}$ , or $V_{DD1}$ missing	$V_{Fail-Safe}$	-	-2.6	-2.5	V
Output bandwidth		$BW$	250	310	-	$\text{kHz}$
Power supply rejection ratio <sup>(1)</sup>	PSRR vs. $V_{DD1}$ , at DC	$PSRR_{DC}$	-	-104	-	dB
	PSRR vs. $V_{DD1}$ , 100 mV and 10 kHz ripple	$PSRR_{AC}$	-	-102	-	dB
	PSRR vs. $V_{DD2}$ , at DC	$PSRR_{DC}$	-	-90	-	dB
	PSRR vs. $V_{DD2}$ , 100 mV and 10 kHz ripple	$PSRR_{AC}$	-	-85	-	dB
Output resistance		$R_{OUT}$	-	< 0.2	-	$\Omega$
Output short-circuit current		$I_{OUT,OC}$	-	$\pm 13$	-	mA
Common-mode transient immunity		$CMTI$	100	150	-	$\text{kV}/\mu\text{s}$
<b>TIMING</b>						
Rising time of $OUT_P$ , $OUT_N$		$t_r$	-	1.3	-	$\mu\text{s}$
Falling time of $OUT_P$ , $OUT_N$		$t_f$	-	1.3	-	$\mu\text{s}$
$IN_P$ , $IN_N$ to $OUT_P$ , $OUT_N$ signal delay (50 % to 50 %)		$t_{PD}$	-	1.6	2.1	$\mu\text{s}$
Analog setting time	$V_{DD1}$ step to 3.0 V with $V_{DD2} \geq 3.0 \text{ V}$ , to $OUT_P$ , $OUT_N$ valid, 0.1 % settling	$t_{AS}$	-	0.5	-	ms

**Note**
<sup>(1)</sup> Input referred

**THERMAL INFORMATION**

PARAMETER	SYMBOL	VALUE	UNIT
Junction to ambient thermal resistance	$R_{\theta JA}$	86	°C/W
Junction to case (top) thermal resistance	$R_{\theta JC(\text{top})}$	28	°C/W
Junction to board thermal resistance	$R_{\theta JB}$	42	°C/W
Junction to top characterization parameter	$\Psi_{JT}$	4	°C/W
Junction to board characterization parameter	$\Psi_{JB}$	42	°C/W

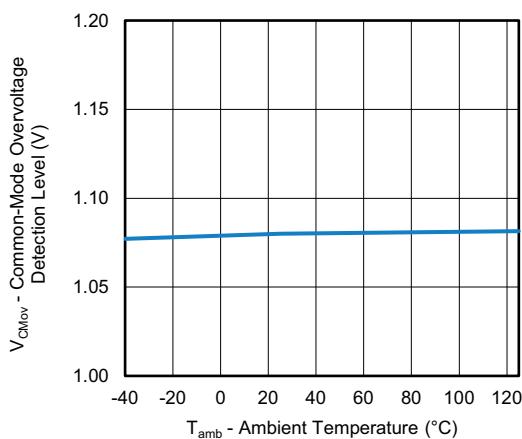
**TYPICAL CHARACTERISTICS**
 $(V_{DD1} = 5 \text{ V}, V_{DD2} = 3.3 \text{ V}, V_{IN} = -250 \text{ mV to } +250 \text{ mV})$ 


Fig. 3 - Common-Mode Overvoltage Detection Level vs. Ambient Temperature

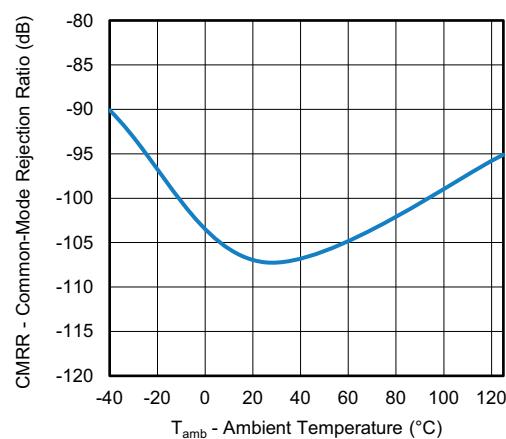


Fig. 5 - Common-Mode Rejection Ratio vs. Ambient Temperature

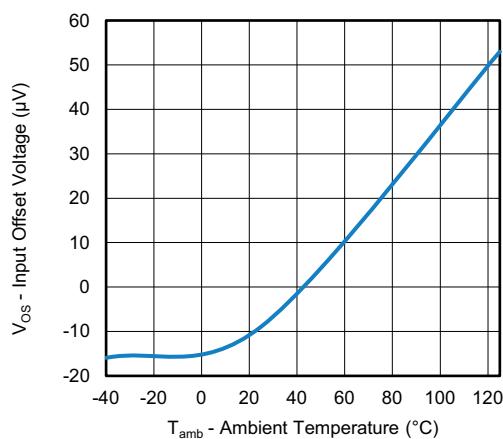


Fig. 4 - Input Offset Voltage vs. Ambient Temperature

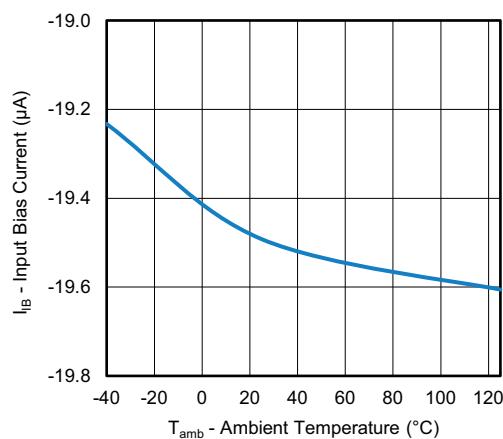


Fig. 6 - Input Bias Current vs. Ambient Temperature

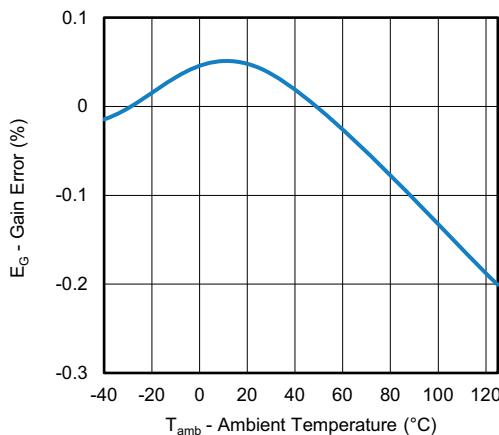


Fig. 7 - Gain Error vs. Ambient Temperature

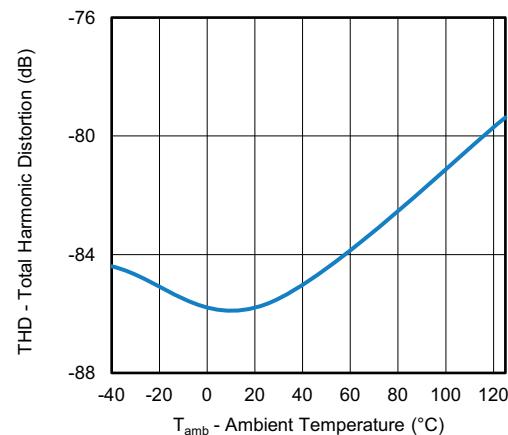


Fig. 10 - THD vs. Ambient Temperature

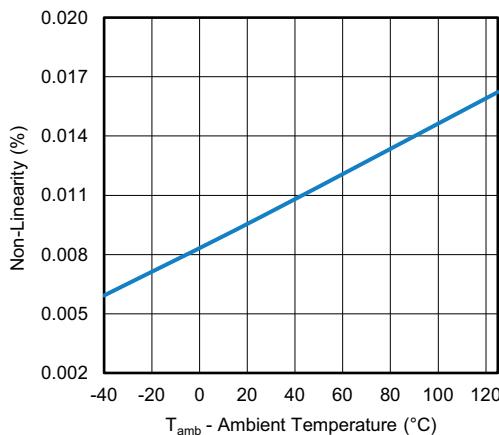


Fig. 8 - Non-Linearity vs. Ambient Temperature

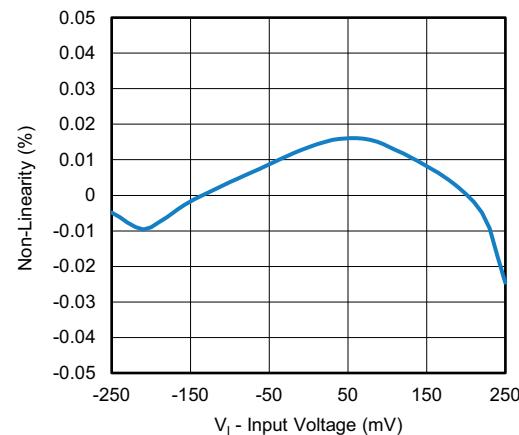


Fig. 11 - Non-Linearity vs. Input Voltage

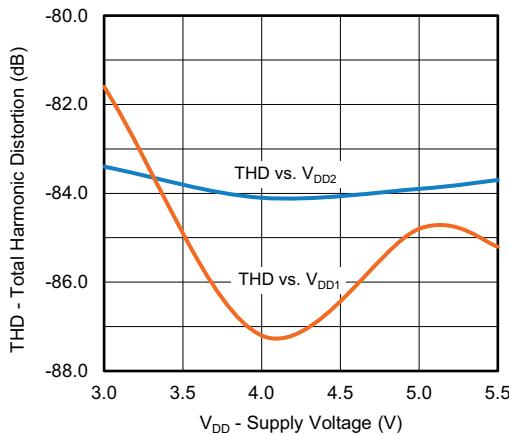


Fig. 9 - THD vs. Supply Voltage

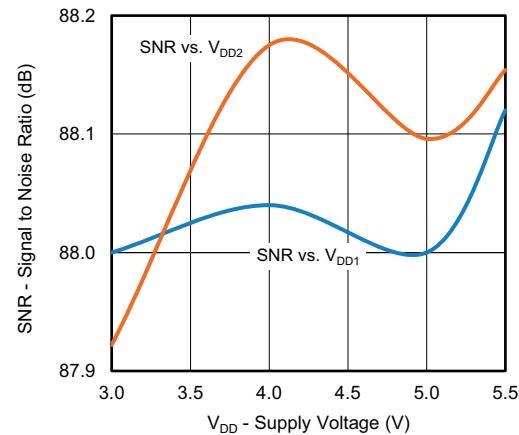


Fig. 12 - SNR vs. Supply Voltage

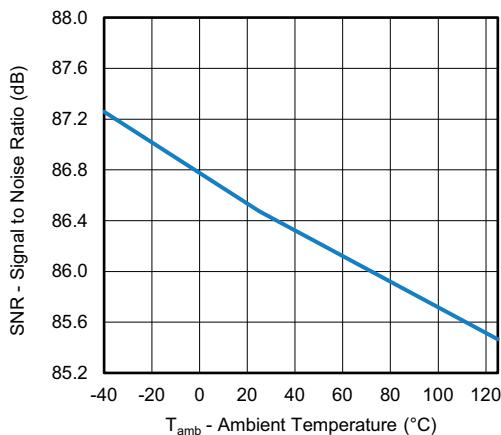


Fig. 13 - SNR vs. Ambient Temperature

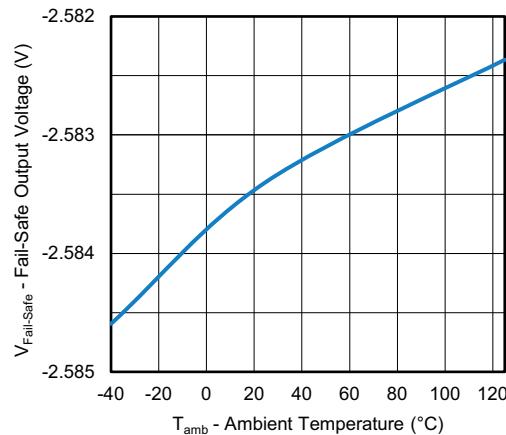


Fig. 16 - Fail-Safe Output Voltage vs. Ambient Temperature

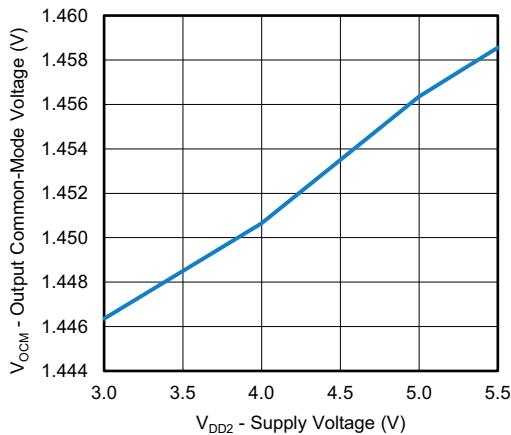


Fig. 14 - Output Common-Mode Voltage vs. Side 2 Supply Voltage

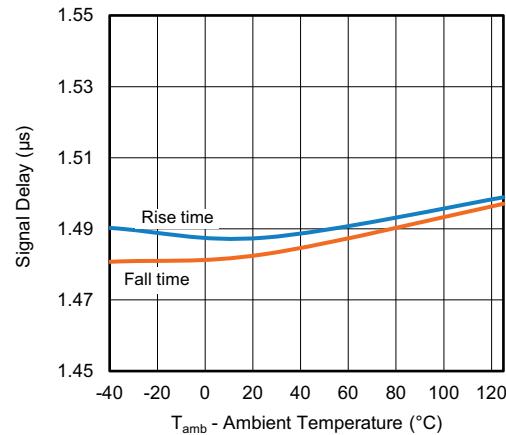


Fig. 17 -  $V_{IN}$  to  $V_{OUT}$  Delay vs. Ambient Temperature

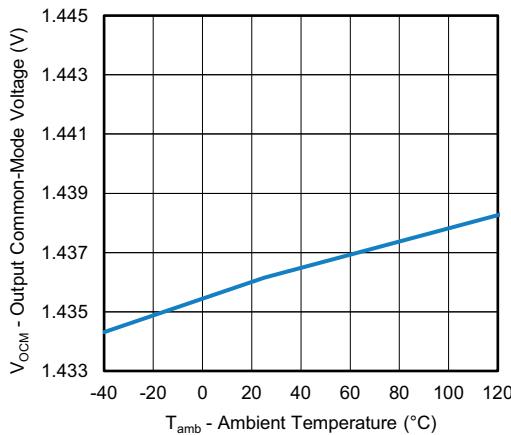


Fig. 15 - Output Common-Mode Voltage vs. Ambient Temperature

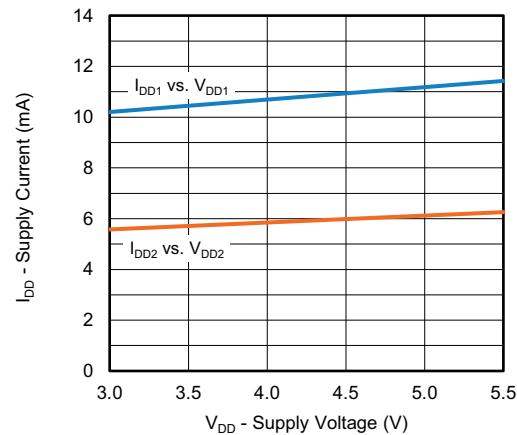


Fig. 18 - Supply Current vs. Supply Voltage

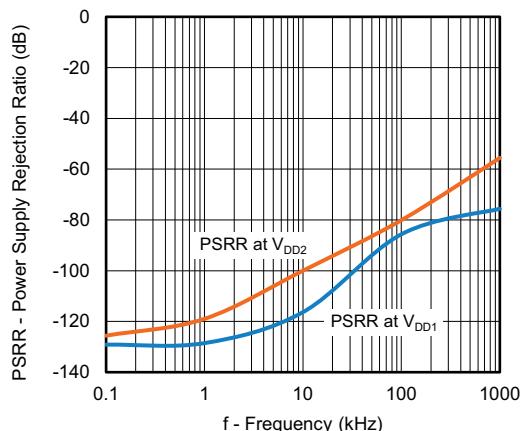


Fig. 19 - Power Supply Rejection Ratio vs. Ripple Frequency

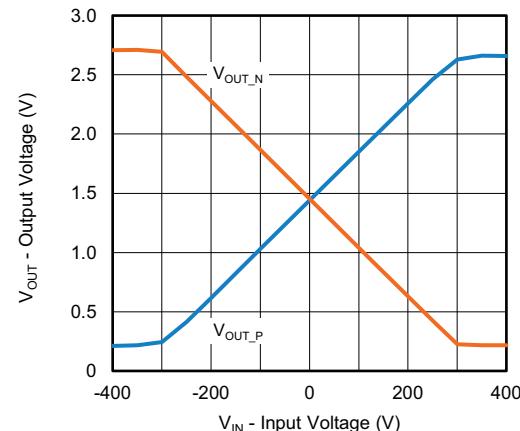


Fig. 21 - Output Voltage vs. Input Voltage

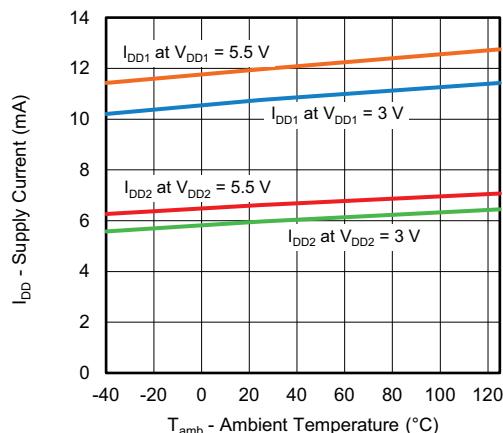


Fig. 20 - Supply Current vs. Ambient Temperature

## PARAMETER MEASUREMENT INFORMATION

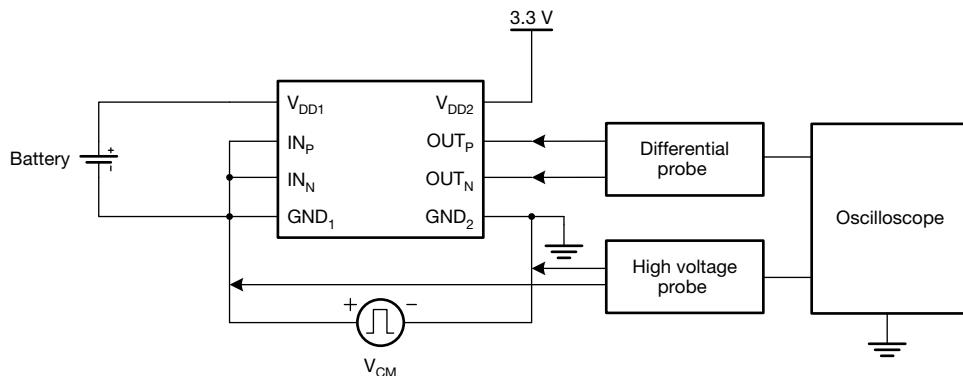


Fig. 22 - Common-Mode Transient Immunity Test Circuit

SAFETY AND INSULATION RATINGS				
PARAMETER	TEST CONDITION	SYMBOL	VALUE	UNIT
Climatic classification	According to IEC 68 part 1		40 / 125 / 21	
Comparative tracking index	DIN EN 60112 (VDE 0303-11); IEC 60112	CTI	> 600	
Maximum rated withstanding isolation voltage	$V_{TEST} = V_{ISO}$ , $t = 1$ min (qualification); $V_{TEST} = 1.2 \times V_{ISO}$ , $t = 1$ s (100 % production test)	$V_{ISO}$	5000	$V_{RMS}$
Maximum transient isolation voltage	$t = 1$ min	$V_{IOTM}$	8000	$V_{peak}$
Maximum repetitive isolation voltage		$V_{IORM}$	2121	$V_{peak}$
Maximum surge isolation voltage	Test method per IEC 60065, 1.2/50 $\mu$ s waveform, $V_{TEST} = V_{IOSM} \times 1.6$	$V_{IOSM}$	6250	$V_{peak}$
Maximum working isolation voltage	AC voltage	$V_{IOWM}$	1500	$V_{RMS}$
	DC voltage		2121	$V_{DC}$
Isolation resistance	$T_{amb} = 25$ °C, $V_{IO} = 500$ V	$R_{IO}$	$> 10^{12}$	$\Omega$
	$T_{amb} = 125$ °C, $V_{IO} = 500$ V	$R_{IO}$	$> 10^{10}$	$\Omega$
	$T_{amb} = 150$ °C, $V_{IO} = 500$ V	$R_{IO}$	$> 10^9$	$\Omega$
Total power dissipation at 25 °C	$\theta_{JA} = 86$ °C/W, $V_I = 5.5$ V, $T_j = 150$ °C, $T_{amb} = 25$ °C	$P_S$	1430	mW
Safety input, output, or supply current	$\theta_{JA} = 86$ °C/W, $V_I = 5.5$ V, $T_j = 150$ °C, $T_{amb} = 25$ °C	$I_S$	260	mA
Maximum safety temperature		$T_S$	150	°C
Creepage distance	SOP-8 (300 mils)		$\geq 8$	mm
Clearance distance			$\geq 8$	mm
Insulation thickness	Distance through insulation	$DTI$	32	$\mu$ m
Material group	IEC 60664-1		I	
For rated mains voltage $\leq 150$ $V_{RMS}$			I to IV	
For rated mains voltage $\leq 300$ $V_{RMS}$			I to IV	
For rated mains voltage $\leq 400$ $V_{RMS}$			I to IV	
Pollution degree per DIN VDE 0110, table 1			2	
Input to output test voltage, method B1	$V_{IORM} \times 1.875 = V_{pd(m)}$ , 100 % production test; $t_{ini} = t_m = 1$ s, partial discharge $< 5$ pC	$V_{pd(m)}$	3977	$V_{peak}$
After environmental tests subgroup 1	$V_{IORM} \times 1.6 = V_{pd(m)}$ , $t_{ini} = 1$ s, $t_m = 10$ s, partial discharge $< 5$ pC	$V_{pd(m)}$	3394	$V_{peak}$
Isolation capacitance	$f = 1$ MHz	$C_{IO}$	0.8	pF

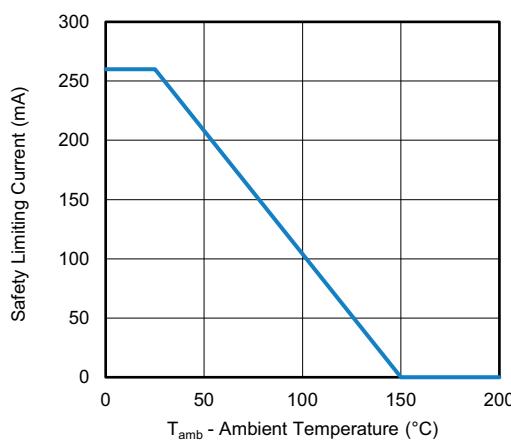


Fig. 23 - VIA0250DD Thermal Derating Curve, Dependence of Safety Limiting Values With Case Temperature per DIN VOE V0884-11

## FUNCTION DESCRIPTION

### Overview

The VIA0250DD is a high performance isolated amplifier that accept fully-differential input. The fully-differential input is ideally suited to shunt current monitoring in high voltage applications where isolation is required. The analog input is continuously sampled by a second-order - modulator in the device, which is driven by a pre-stage fully-differential amplifier in the device. With the internal voltage reference and clock generator, the modulator convert the analog input signal to a digital bitstream. The output of the modulator is transferred by the drivers (called TX in the functional block diagram) across the isolation barrier that separates the isolated side 1 and side 2 voltage. The received bitstream and clock are synchronized and processed, as shown in the functional block diagram, by a fourth-order analog filter on the side 2 and has a differential output.

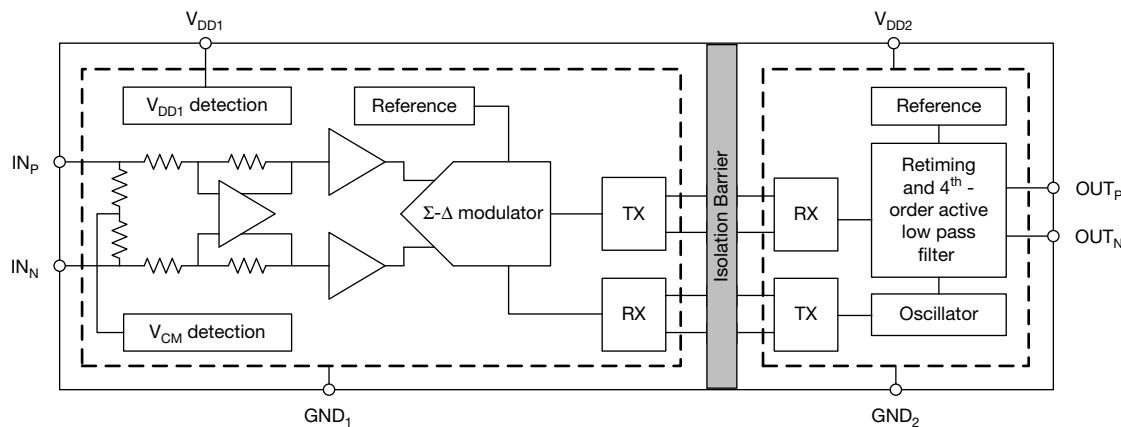


Fig. 24 - Function Block Diagram

### Analog Input

Below mentioned are the restrictions on the analog input signal ( $V_{IN}$ ).

1. If the input voltage exceeds the range  $GND_1 - 6\text{ V}$  to  $V_{DD1} + 0.5\text{ V}$ , the input current must be limited to 10 mA because the device input electrostatic discharge (ESD) diodes turn on
2. The linearity and noise performance of the device are ensured only when the analog input voltage remains within the specified linear full-scale range (FSR) and within the specified common-mode input voltage range

### Analog Output

For linear input range, the analog output of VIA0250DD has a fixed gain of 8.2. If a full-scale input signal is applied to the VIA0250DD ( $V_I \geq V_{anosing}$ ), the analog output will be clipped (typically, 2.45 V for positive clipping and -2.45 V for negative clipping).

In addition, VIA0250DD integrates some diagnostic measures and offers a fail-safe output to simplify system-level design. The fail safe output is a negative differential output voltage that is activated in the conditions mentioned below. Please note that the fail safe output does not occur during normal operation.

1. When the undervoltage of  $V_{DD1}$  is detected ( $V_{DD1} < V_{DD1uv}$ )
2. When the overvoltage of common-mode input voltage is detected ( $V_{aw} > V_{cewo}$ )

### APPLICATION NOTE

#### Typical Application Circuit

VIA0250DD is ideally suited to shunt resistor-based current sensing in high voltage applications such as frequency inverters. The typical application circuit is shown in Fig. 25.

The voltage across the shunt resistor  $R_{sense}$  is applied to the differential input of VIA0250DD through a RC filter. The differential output of the isolated amplifier is converted to a single-ended analog output using an operational amplifier. Suggest to add  $> 1$  k $\Omega$  resistor on the OUT<sub>P</sub> and OUT<sub>N</sub> pin to prevent output over-current. An analog to digital converter receives the analog output and converts to digital signal for controller processing.

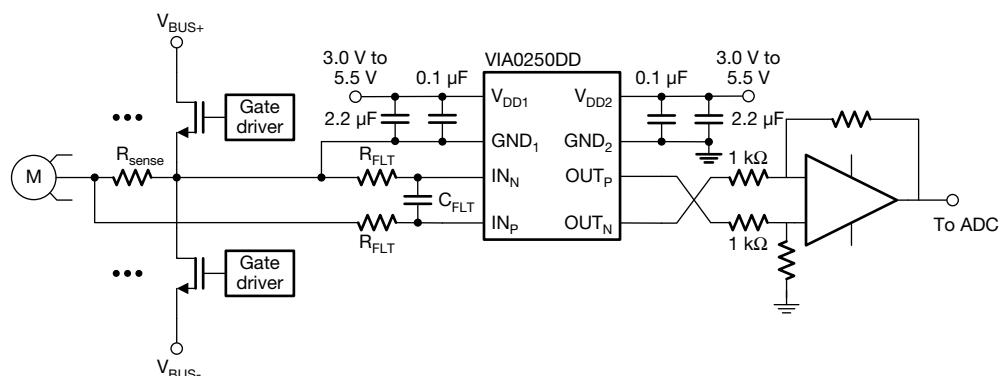


Fig. 25 - Typical Application Circuit in Phase Current Sensing

### Shunt Resistor Selection

Choosing a particular shunt resistor is usually a compromise between minimizing power dissipation and maximizing accuracy. Smaller sense resistor decreases power dissipation, while larger sense resistor can improve measure accuracy by utilizing the full input range of isolated amplifier.

There are two other factors should be considered when selecting the shunt resistor:

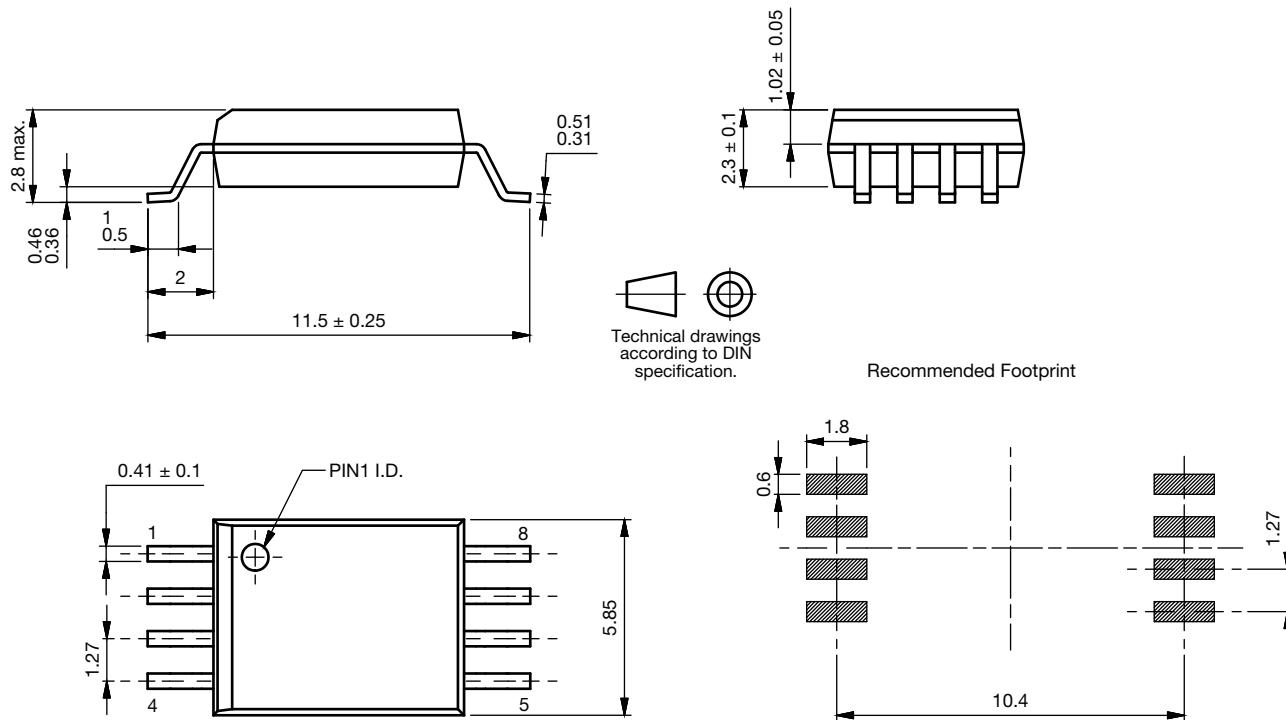
- The voltage-drop caused by the rated current range must not exceed the recommended linear input voltage range:  $V_{SHUNT} \leq FSR$
- The voltage-drop caused by the maximum allowed overcurrent must not exceed the input voltage that causes a clipping output:  $V_{SHUNT} \leq V_{clipping}$

### PCB Layout

There are some key guidelines or considerations for optimizing performance in PCB layout:

- VIA0250DD requires a 0.1  $\mu$ F bypass capacitor between  $V_{DD1}$  and  $GND_1$ ,  $V_{DD2}$  and  $GND_2$ . The capacitor should be placed as close as possible to the  $V_{DD}$  pin. If better filtering is required, an additional 1  $\mu$ F to 10  $\mu$ F capacitor may be used.
- Kelvin rules is recommended for the connection between shunt resistor to VIA0250DD. Because of the Kelvin connection, any voltage drops across the trace and leads should have no impact on the measured voltage.
- Place the shunt resistor close to the  $IN_P$  and  $IN_N$  inputs and keep the layout of both connections symmetrical and run very close to each other to the input of the VIA0250DD. This minimizes the loop area of the connection and reduces the possibility of stray magnetic fields from interfering with the measured signal.

### PACKAGE DIMENSIONS (in millimeters)



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Fig. 26 - SOP-8 (300 mil) Package Shape and Dimensions

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