

**MAX77985/MAX77986****19VIN, 3.5/5.5A 1-Cell Li+ Battery Charger with Smart Power Selector and OTG for USBC PD****General Description**

The MAX77985/MAX77986 is a high-performance high input range 3.5/5.5A fast charger with Smart Power Selector™. The IC can operate as a reverse boost without an additional inductor, allowing the battery to share its power through the charging port with voltage programmable from 5V to 12V. The device features fully integrated low-loss power switches to provide a small solution size and high efficiency, even at high input voltage and high charging current. Its high switching frequency allows the use of a smaller sized inductor. Li-ion, Li-polymer, and LiFePO<sub>4</sub> battery chemistries are supported. The IC is compatible with USB Power Delivery sources with input current capability up to 5A. It can also provide fast unplug detection of 5V, 9V, and 15V sources for dynamic management of system load. The IC also helps mitigate EMI with the spread spectrum feature. The IC features true load disconnection in reverse boost mode and has an adjustable output current protection limit. The device is highly flexible and programmable through I<sup>2</sup>C configuration.

The battery charger includes a Smart Power Selector to accommodate a wide range of battery sizes and system loads. The Smart Power Selector allows the system to start up gracefully as soon as an input source is available, even when the battery is deeply discharged (dead battery) or missing. It can be configured so that when power is applied to the charger input, the battery charging can automatically start.

**Note:** For parts with chip revision 0b001 (PASS1), contact Analog Devices for Rev 0 of the data sheet. This information can be read from the CHIP\_REVISION (0x01) register.

**Applications**

- Gaming Devices
- VR Applications
- mPOS
- Tablet PCs

**Benefits and Features**

- High-Efficiency Single-Cell Switching Charger
  - Up to 5.5A Charging with MAX77986
  - Up to 3.5A Charging with MAX77985
  - 92% Buck Efficiency at 4A, 12V Input
  - 92% Charging Efficiency at 3.5A, 9V Input
  - Optimized for High Voltage Input Operation
  - Accelerate Charge Time by Monitoring Kelvin Sensing Battery Voltage
  - Up to 5.5A Input Current Limit with AICL (MAX77986)
  - Up to 3.5A Input Current Limit with AICL (MAX77985)
- +28V Absolute Maximum Input Voltage Rating
- 4.7V to 19V Input Operating Voltage Range
- Reverse Boost with Programmable Output Voltage Options up to 12V
  - Up to 18W for MAX77986
  - Up to 12W for MAX77985
- Integrated Battery True-Disconnect FET
  - RDSON = 7.7mΩ
  - Programmable Discharge Current Limit up to 10A
  - Support Shipping Mode and Low Battery Leakage Current
  - 1.3MHz/2.6MHz Switching Frequency with 1µH/0.47µH Inductor
  - Disconnect Input (DISQBAT)
- Safety
  - Battery Temperature Sensing and Charge Safety Timer
  - JEITA Guideline Compliant
  - Thermal Regulation and Thermal Shutdown
  - System Voltage OVLO/UVLO
- Spread Spectrum for Noise Sensitive Applications
- Programmable Unplug Detection for 9V and 15V Sources
- Charge Status Output for LED
- Push-Button Input for Exiting from Ship Mode
- External Discharge FET Enable Output
- Dedicated Input for Suspend Mode (SUSPND)
- I<sup>2</sup>C Interface
- 4mm x 4mm FC2QFN

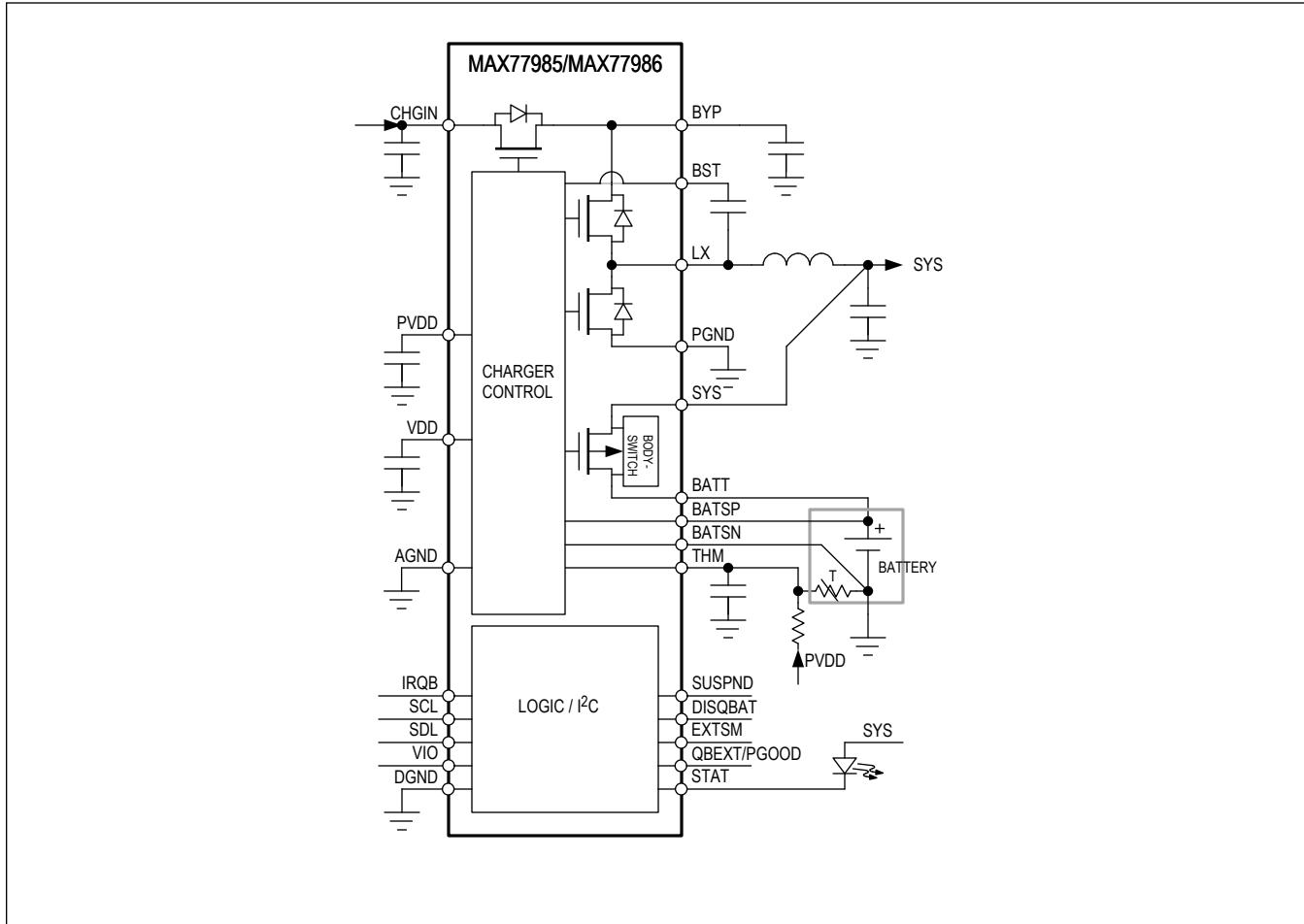
Smart Power Selector is a trademark of Maxim Integrated Products, Inc.  
USB Type-C is a registered trademark of USB Implementers Forum.  
PowerPath is a trademark of Linear Technology Corporation.

[Ordering Information](#) appears at end of data sheet.

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**Simplified Block Diagram**

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## Absolute Maximum Ratings

CHGIN to PGND.....	-0.3V to +28V	DISQBAT, SUSPEND, QBEXT to AGND .....	-0.3V to +6.0V
BYP to PGND .....	-0.3V to +28V	EXTSM to AGND .....	-0.3V to $V_{BATT}$ +0.3V
BYP to CHGIN .....	-0.3V to +16V	IRQB, STAT to AGND.....	-0.3V to +6.0V
BYP to LX .....	-0.3V to +28V	THM to AGND .....	-0.3V to $V_{PVDD}$ +0.3V
LX to PGND .....	-0.3V to +22V	SDA, SCL to AGND .....	-0.3V to +6.0VV
BST to PVDD .....	-0.3V to +22V	CHGIN Continuous Current .....	5.6A <sub>RMS</sub>
BST to LX .....	-0.3V to +2.2V	BYP Continuous Current.....	3.4A <sub>RMS</sub>
SYS to AGND .....	-0.3V to +6.0V	PGND Continuous Current .....	6.5A <sub>RMS</sub>
BATT to AGND .....	-0.3V to +6.0V	LX Continuous Current .....	9.8A <sub>RMS</sub>
BATSP to AGND .....	-0.3V to $V_{BATT}$ +0.3V	SYS, BATT Continuous Current .....	10.0A <sub>RMS</sub>
BATSP to BATT .....	-0.3V to +0.3V	Continuous Power Dissipation (Multilayer Board) ( $T_A = +70^\circ\text{C}$ , deration is 35.34mW/°C above +70°C).....	mW to 2826.86mW
BATSN to AGND .....	-0.3V to +0.3V	Operating Temperature Range .....	-40°C to +85°C
PGND to AGND .....	-0.3V to +0.3V	Junction Temperature .....	+150°C
DGND to AGND .....	-0.3V to +0.3V	Storage Temperature Range .....	-65°C to +150°C
PVDD to PGND .....	-0.3V to +2.2V	Soldering Temperature (reflow) .....	+260°C
VDD to AGND.....	-0.3V to +2.2V		
VIO to AGND .....	-0.3V to +6.0V		

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

## Package Information

### FCQFN

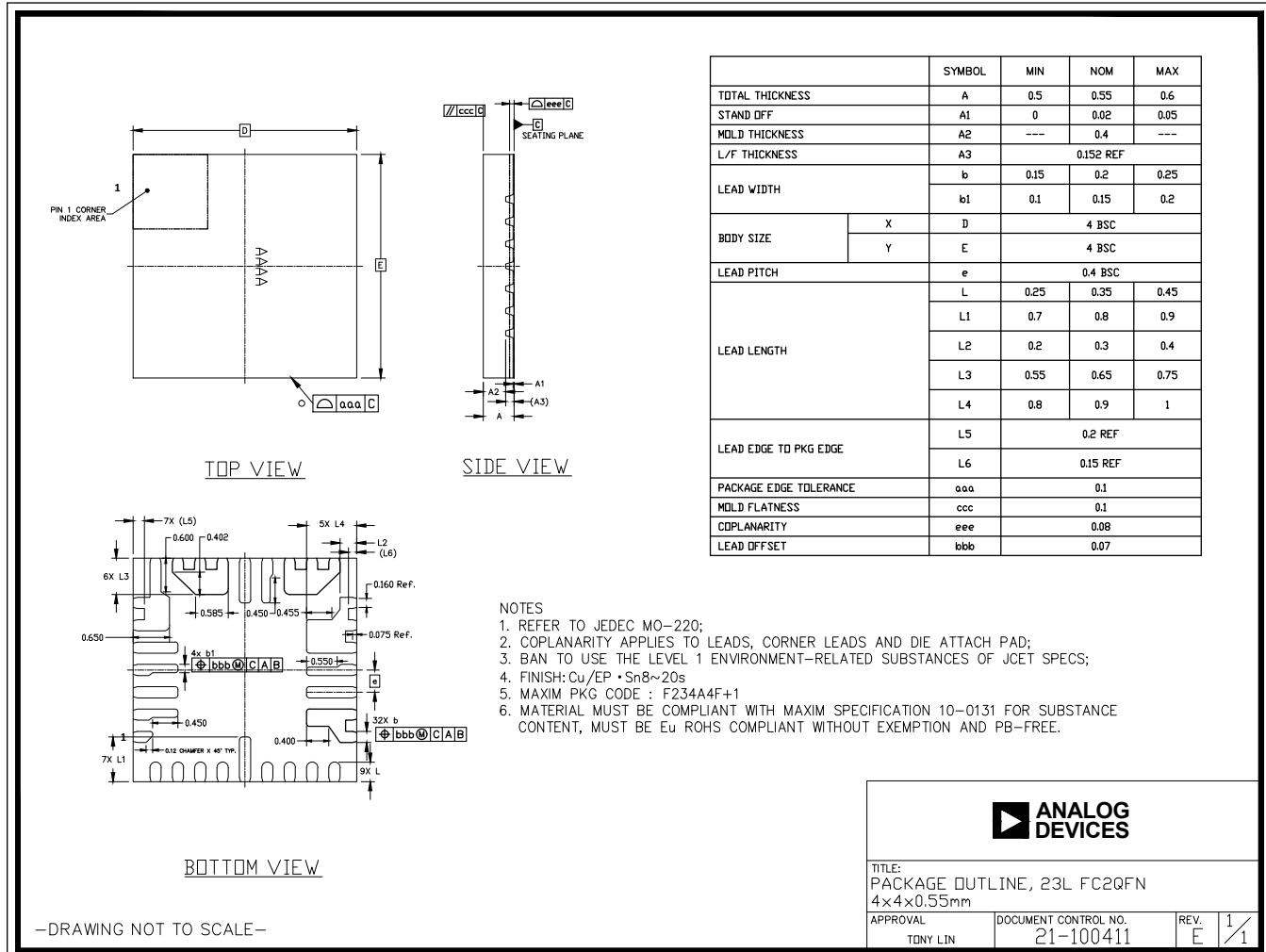
Package Code	F234A4F+1
Outline Number	<a href="#">21-100411</a>
Land Pattern Number	<a href="#">90-100145</a>
Thermal Resistance, Four-Layer Board:	
Junction-to-Ambient ( $\theta_{JA}$ )	28.30°C/W
Junction-to-Case Thermal Resistance ( $\theta_{JC}$ )	6.65°C/W
Junction-to-Board ( $\theta_{JB}$ )	12°C/W

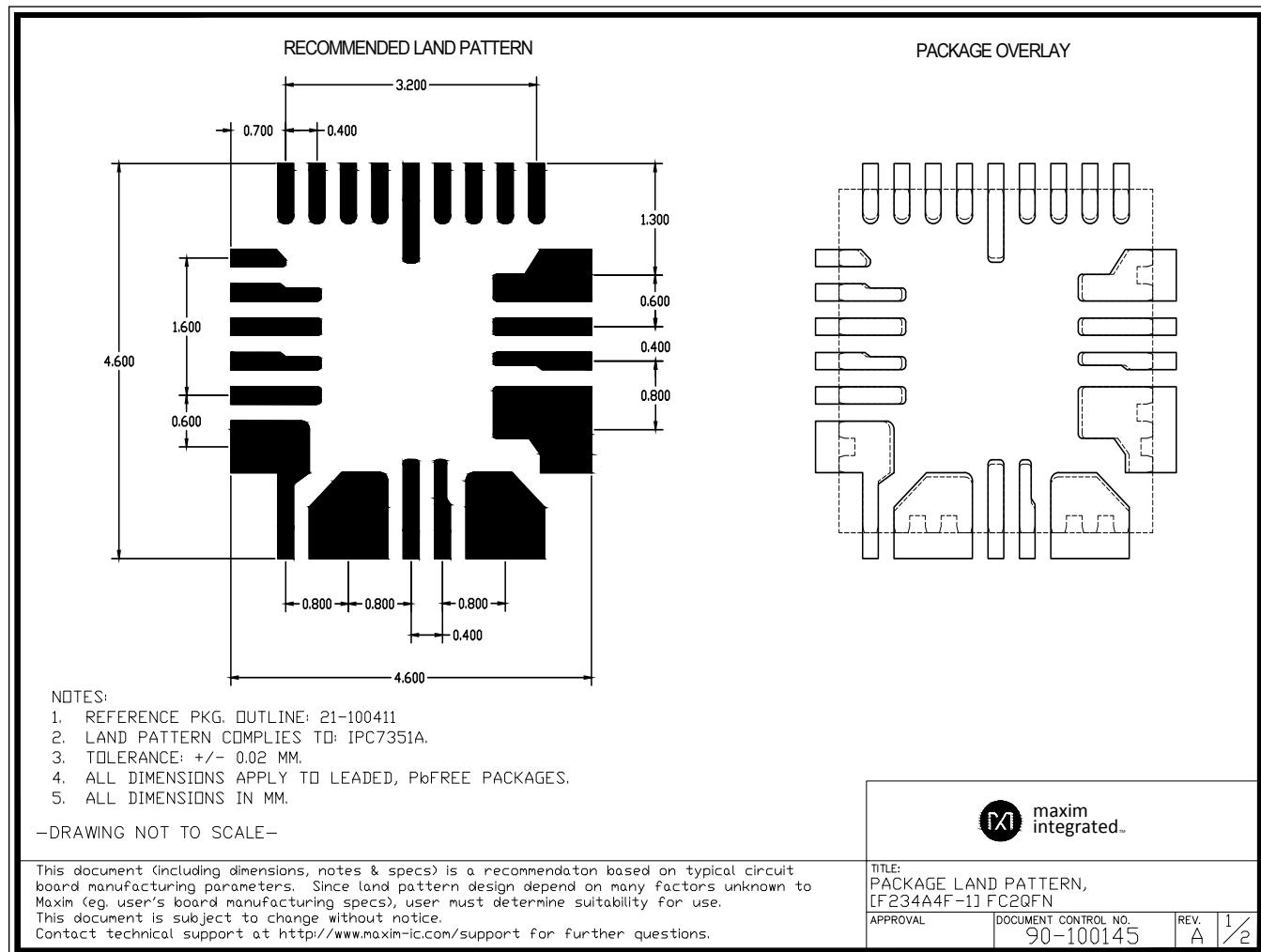
For the latest package outline information and land patterns (footprints), go to [www.maximintegrated.com/packages](http://www.maximintegrated.com/packages). Note that a "+", "#", or "-" in the package code indicates RoHS status only. Package drawings may show a different suffix character, but the drawing pertains to the package regardless of RoHS status.

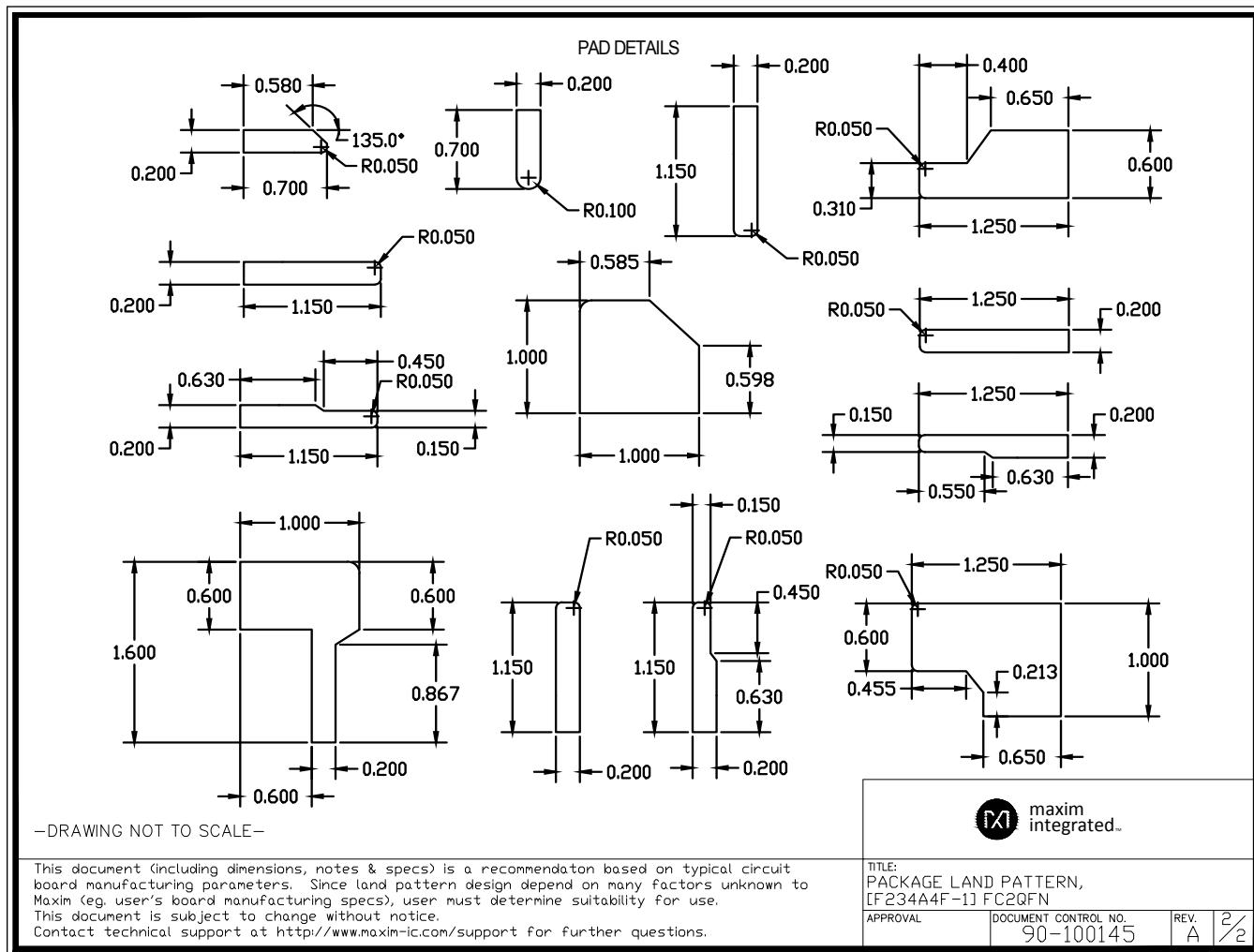
Package thermal resistances were obtained using the method described in JEDEC specification JESD51-7, using a four-layer board. For detailed information on package thermal considerations, refer to [www.maximintegrated.com/thermal-tutorial](http://www.maximintegrated.com/thermal-tutorial).

## MAX77985/MAX77986

## 19VIN, 3.5/5.5A 1-Cell Li+ Battery Charger with Smart Power Selector and OTG for USBC PD







## Electrical Characteristics

( $V_{SYS} = 3.8V$ ,  $V_{BATT} = 3.8V$ ,  $V_{VIO} = 1.8V$ ,  $V_{CHGIN} = 5V$ , unless otherwise specified. Limits are production tested at  $T_A = +25^\circ C$ . Limits over the operating temperature range and relevant supply voltage range are guaranteed by design and characterization.)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
<b>GENERAL ELECTRICAL CHARACTERISTICS</b>						
CHGIN Quiescent Current	$I_{CHGIN}$	$V_{CHGIN} = 5.0V$ , SUSPEND pin digital high or MODE = 0, DEEP_SUSP_DIS = 1	0.19	0.32		mA
		$V_{CHGIN} = 5.0V$ , SUSPEND pin digital high or MODE = 0, DEEP_SUSP_DIS = 0	85			$\mu A$
		$V_{CHGIN} = 5.0V$ , $V_{BATT} = 4.2V$ , MODE = 5, DONE state ( $V_{SYS} = 4.35V$ ), $I_{SYS} = 0A$ ( <a href="#">Note 3</a> )	2.2	3.3		mA
Input Undervoltage Supply Current	$I_{IN}$	$V_{CHGIN} = 2.4V$ , the input is undervoltage	0.035			mA

## Electrical Characteristics (continued)

( $V_{SYS} = 3.8V$ ,  $V_{BATT} = 3.8V$ ,  $V_{VIO} = 1.8V$ ,  $V_{CHGIN} = 5V$ , unless otherwise specified. Limits are production tested at  $T_A = +25^\circ C$ . Limits over the operating temperature range and relevant supply voltage range are guaranteed by design and characterization.)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
BAT Quiescent Current	$I_{BAT}$	$V_{CHGIN} = 0V$ , $V_{BATT} = 3.6V$ , $Q_{BATT}$ FET is on, $B2SOVRC\_CTRL = 0$ , $LPM = 0$ , $I_{SYS} = 0A$		29		$\mu A$
BAT Quiescent Current in Low-Power Mode	$I_{BAT}$	$V_{CHGIN} = 0V$ , $V_{BATT} = 3.6V$ , $Q_{BATT}$ FET is on, $B2SOVRC = 0$ , $LPM = 1$ , $I_{SYS} = 0A$		22		$\mu A$
BAT Quiescent Current in Factory-Ship Mode	$I_{BAT}$	$V_{CHGIN} = 0V$ , $V_{BATT} = 3.6V$ , $Q_{BATT}$ FET is off, $V_{SYS} = V_{VDD} = 0V$ , factory-ship mode	3.0	4.5		$\mu A$
BAT Quiescent Current in Done State	$I_{MBDN}$	$V_{CHGIN} = 5V$ , $I_{BYP} = 0A$ , $V_{BATT} = 4.2V$ , $I_{SYS} = 0A$ , $Q_{BATT}$ FET is off, $B2SOVRC = 0$ , $MODE = 5$ , done state		7.5	10.5	$\mu A$
SYS Operating Voltage	$V_{SYS}$	Guaranteed by $V_{SYS\_UVLO\_R}$ and $V_{SYS\_OVLO\_R}$	$V_{SYS\_U}$ $V_{VLO\_R}$	$V_{SYS\_O}$ $V_{VLO\_R}$		V
VIO Operating Voltage Range	$V_{VIO}$		1.62		5.5	V
SCL, SDA Input Low Level	$V_{SCL\_SDA\_IN\_L}$	$T_A = +25^\circ C$			$0.3 \times V_{VIO}$	V
SCL, SDA Input High Level	$V_{SCL\_SDA\_IN\_H}$	$T_A = +25^\circ C$			$0.7 \times V_{VIO}$	V
SCL, SDA Input Hysteresis	$V_{SCL\_SDA\_HY\_S}$	$T_A = +25^\circ C$			$0.05 \times V_{VIO}$	V
SCL, SDA Logic Input Current	$I_{SCL\_SDA}$	$V_{SCL} = V_{SDA} = V_{VIO} = 1.9V$	-10		+10	$\mu A$
SDA Output Low Voltage	$V_{SDA\_OUT\_L}$	$I_{SDA} = 20mA$ sinking			0.4	V
IRQB Output Low Voltage	$V_{IRQB\_OUT\_L}$	$I_{IRQB} = 1mA$ sinking			0.4	V
IRQB Output High Leakage	$I_{IRQB\_H}$	$V_{IRQB} = 5.5V$ , $T_A = +25^\circ C$	-1	0	+1	$\mu A$
		$V_{IRQB} = 5.5V$ , $T_A = +85^\circ C$			0.1	
<b>CHGIN INPUT LIMITER</b>						
CHGIN Operating Voltage Range	$V_{CHGIN}$	$V_{CHGIN}$ must be less than $V_{CHGIN\_OVLO}$ and greater than both $V_{CHGIN\_UVLO}$ and $(V_{SYS} + V_{CHGIN2SYS\_TH})$ for the charger to turn-on	$V_{CHGIN\_UVLO}$	$V_{CHGIN\_OVLO}$		V
CHGIN Overvoltage Threshold	$V_{CHGIN\_OVLO}$	$V_{CHGIN}$ rising	19	19.5	20	V
CHGIN Overvoltage Threshold Hysteresis	$V_{CHGIN\_OVLO\_HYS}$				500	mV
CHGIN Undervoltage Threshold Setting Range	$V_{CHGIN\_UVLO}$	$V_{CHGIN}$ rising, 20% hysteresis, programmable at 4.7V, 4.8V, 4.9V, 5.05V	4.7		5.05	V
CHGIN Undervoltage Threshold Accuracy	$V_{CHGIN\_UVLO\_ACC}$	$V_{CHGIN}$ rising, 4.7V setting	4.625	4.7	4.775	V
CHGIN to SYS Undervoltage Threshold Rising	$V_{CHGIN2SYS\_TH}$	$V_{CHGIN} - V_{SYS}$ , rising	0.15	0.20	0.25	V

## Electrical Characteristics (continued)

( $V_{SYS} = 3.8V$ ,  $V_{BATT} = 3.8V$ ,  $V_{VIO} = 1.8V$ ,  $V_{CHGIN} = 5V$ , unless otherwise specified. Limits are production tested at  $T_A = +25^\circ C$ . Limits over the operating temperature range and relevant supply voltage range are guaranteed by design and characterization.)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
CHGIN Turn-On Threshold Validation Delay	$t_{D-UVLO}$	Delay from $V_{CHGIN} > V_{CHGIN\_UVLO}$ to $Q_{CHGIN}$ FET enable		8		ms
CHGIN Switching Start Delay	$t_{START}$	Delay from Input Validation to LX switching (if charge or buck mode is selected and charger is not suspended); see the Input Validation section for input validation conditions		150		ms
CHGIN Adaptive Voltage Regulation Threshold Setting Range	$V_{CHGIN\_REG}$	Programmable at 4.5V, 4.6V, 4.7V, 4.85V. The input voltage regulation loop decreases the input current to regulate $V_{CHGIN}$ at $V_{CHGIN\_REG}$ under weak input source conditions. If the input current is decreased to $I_{IULO\_DET}$ and the input voltage is equal or below $V_{CHGIN\_REG}$ , then the charger input is turned off.	4.5		4.85	V
CHGIN Adaptive Voltage Regulation Threshold Accuracy	$V_{CHGIN\_REG\_ACC}$	4.5V setting	4.4	4.5	4.6	V
CHGIN Input Current Limit Setting Range	$I_{INLIMIT}$	Programmable, 500mA default, 50mA step, production tested at 5V (500mA, 2700mA), 9V (1500mA), 12V(1000mA), 15V(500mA) (MAX77985)	0.1		3.5	A
		Programmable, 500mA default, 50mA step, production tested at 5V (500mA, 2700mA, 5000mA), 9V (1500mA, 4000mA), 12V (1000mA, 3000mA), 15V (500mA, 2400mA) (MAX77986)	0.1		5.5	
CHGIN Input Current Limit Accuracy	$I_{INLIMIT\_ACC}$	Charger enabled, $CHGIN = 5V, 9V, 12V$ , and $15V$ , $0.5A \leq CHGIN\_ILIM \leq 3A$ , $T_A = -5^\circ C$ to $+85^\circ C$ ( <a href="#">Note 3</a> )	-10		-2	%
CHGIN Input Current Low Threshold	$I_{IULO\_DET}$	Charger enabled, 3200mA input current limit setting		60		mA
CHGIN Unplug Detection Threshold Accuracy (5V)	$V_{CHGIN\_UNPL\_G\_ACC\_5V}$ (AICL enabled)	$V_{CHGIN}$ falling, $DIS\_AICL=0b0$ , $UNPLUG\_TH[1:0]=0b01$ , $V_{CHGIN\_REG}[1:0]=0b01$ ( <a href="#">Note 1</a> , <a href="#">Note 3</a> )	4.5	4.6	4.7	V
	$I_{CHGIN\_UNPL\_G\_ACC\_5V}$ (AICL enabled)	$I_{CHGIN}$ : $DIS\_AICL=0b0$ , $UNPLUG\_TH[1:0]=0b01$ , $V_{CHGIN\_REG}[1:0]=0b01$ , 1ms debounced ( <a href="#">Note 1</a> , <a href="#">Note 3</a> )	30	60	90	mA
	$V_{CHGIN\_UNPL\_G\_ACC\_5V}$ (AICL disabled)	$V_{CHGIN}$ falling, $DIS\_AICL=0b1$ , $UNPLUG\_TH[1:0]=0b01$ ( <a href="#">Note 2</a> , <a href="#">Note 3</a> )	3.54	3.84	4.14	V

## Electrical Characteristics (continued)

( $V_{SYS} = 3.8V$ ,  $V_{BATT} = 3.8V$ ,  $V_{VIO} = 1.8V$ ,  $V_{CHGIN} = 5V$ , unless otherwise specified. Limits are production tested at  $T_A = +25^\circ C$ . Limits over the operating temperature range and relevant supply voltage range are guaranteed by design and characterization.)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS	
CHGIN Unplug Detection Threshold Accuracy (9V, 15V)	$V_{CHGIN\_UNPL\_G\_ACC\_9V}$	$V_{CHGIN}$ falling, $UNPLUG\_TH[1:0]=0b10$ ( <a href="#">Note 3</a> )	7.30	7.45	7.60	V	
	$V_{CHGIN\_UNPL\_G\_ACC\_15V}$	$V_{CHGIN}$ falling, $UNPLUG\_TH[1:0]=0b11$ ( <a href="#">Note 3</a> )	12.925	13.125	13.325		
CHGIN Unplug Detection Threshold Hysteresis	$V_{CHGIN\_UNPL\_G\_HYS}$			500			
<b>SYSTEM BUCK</b>							
Buck Output Voltage Setting Range (Tracking Disabled, MODE = 0x4, Variant A)	$V_{SYSREG}$	Programmable 4.15V to 4.5375V in 12.5mV steps (5-bits); production tested at 4.2V only	4.15	4.5375		V	
Buck Output Voltage Setting Range (Tracking Disabled, MODE = 0x4, Variant B)	$V_{SYSREG}$	Programmable 3.50V to 4.275V in 25mV steps (5-bits); production tested at 4.2V only	3.50	4.275		V	
Buck Output Voltage Setting Range (Tracking Disabled, MODE = 0x6, Variant A)	$V_{SYSREG}$	Programmable 4.65V to 5.00V in 50mV steps (3-bits); production tested at 5.0V only	4.65	5.00		V	
Buck Output Voltage Setting Range (Tracking Disabled, MODE = 0x6, Variant B)	$V_{SYSREG}$	Programmable 4.50V to 5.00V in 100mV steps (3-bits); production tested at 5.0V only	4.50	5.00		V	
Buck Output Voltage Accuracy (Tracking Disabled)	$V_{SYSREG\_AC\_C}$	Buck only, charging disabled	-3	+3		%	
Buck Output Voltage (Tracking Enabled)	$V_{SYSREG\_TR\_K\_MIN}$	MODE = 4, MINSYS setting = 3.60V, SYS tracking mode enabled, $V_{BATT} < V_{SYS\_MIN}/1.04$	3.48	3.60	3.72	V	
	$V_{SYSREG\_TR\_K}$	MODE = 4, SYS tracking mode enabled, $V_{BATT} \geq MINSYS - 4\% * V_{BATT}$ , measured of $V_{SYS} - V_{BATT}$ , $V_{SYSREG\_TRK}$ represented as a percentage of $V_{BATT}$ ( <a href="#">Note 3</a> )	4		%		
		MODE=6, SYS Tracking mode enabled, $V_{BATT} \geq BCKSYS - 4\% * V_{BATT}$ , measured of $V_{SYS} - V_{BATT}$ , $V_{SYSREG\_TRK}$ represented as a percentage of $V_{BATT}$ ( <a href="#">Note 3</a> )	4				
Buck Inductor Current Limit	$I_{HSILIM}$	For MAX77986	10.0	11.1	12.2	A	
	$I_{HSILIM}$	For MAX77985	7.5	8.3	9.1		
Buck Minimum On Time	$t_{ON-MIN}$	Measured on LX	100		ns		
Buck Minimum Off Time	$t_{OFF-MIN}$	Measured on LX	100		ns		
System Power-Up Current (from BYP)	$I_{SYSPU\_BYP}$	Charger present, $V_{SYS} < V_{SYS\_UVLO\_R}$	50	75	100	mA	

## Electrical Characteristics (continued)

( $V_{SYS} = 3.8V$ ,  $V_{BATT} = 3.8V$ ,  $V_{VIO} = 1.8V$ ,  $V_{CHGIN} = 5V$ , unless otherwise specified. Limits are production tested at  $T_A = +25^\circ C$ . Limits over the operating temperature range and relevant supply voltage range are guaranteed by design and characterization.)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
System Power-Up Time-Out (from BYP)	$t_{SYSPU\_BYP}$			150		ms
Spread Spectrum Modulation Envelope	$\Delta F_{ss}$	$SS\_ENV = 0b01$		$\pm 4$		%
		$SS\_ENV = 0b10$		$\pm 8$		
		$SS\_ENV = 0b11$		$\pm 16$		
<b>CHARGER</b>						
Precharge Charge Current	$I_{PRECHG}$	$V_{BATT} < V_{PRECHG}$	40	55	80	mA
Precharge Voltage Threshold	$V_{PRECHG}$	$V_{BATT}$ rising	2.425	2.5	2.575	V
Precharge Voltage Threshold Hysteresis	$V_{PRECHG\_HY\_S}$			500		mV
Trickle Charge Current	$I_{TRICKLE}$	$TKEN = 1$ by default, $V_{PRECHG} < V_{BATT} < V_{TRICKLE}$	270	300	330	mA
Trickle Charge Voltage Threshold	$V_{TRICKLE}$	$V_{BATT}$ rising, $TKEN = 1$ by default	3.0	3.1	3.2	V
Trickle Charge Voltage Threshold Hysteresis	$V_{TRICKLE\_HY\_S}$	$TKEN = 1$ by default		100		mV
Prequalification Time	$t_{PQ}$	Applies to the total time of precharge and trickle charge mode		30		min
Fast-Charge Current Setting Range	$I_{FC}$	100mA to 5500mA in 50mA steps; production tested at 500mA, 1000mA, 3000mA, and 5000mA settings (MAX77986 only)	0.1		5.5	A
		100mA to 3500mA in 50mA steps; production tested at 500mA, 1000mA, and 3000mA settings (MAX77985 only)	0.1		3.5	
Fast-Charge Current Accuracy	$I_{FC\_ACC}$	Programmed $I_{FC} \geq 500mA$ , $V_{BATT} > V_{SYSMIN}$ , $T_A = +25^\circ C$	-3.5		+3.5	%
		Programmed $I_{FC} \geq 500mA$ , $V_{BATT} > V_{SYSMIN}$ , $T_A = 0^\circ C$ to $+85^\circ C$	-6		+6	
		Programmed $I_{FC} \geq 500mA$ , $V_{TRICKLE} < V_{BATT} < V_{SYSMIN}$ (LDO mode), $T_A = -5^\circ C$ to $+85^\circ C$	-10		+10	
Fast-Charge Current Thermal Regulation Setting Range	$T_{REG}$	Junction temperature when charge current starts to reduce for thermal regulation; programmable from $+85^\circ C$ to $+130^\circ C$ in $5^\circ C$ steps; the default value is $+115^\circ C$	85		130	°C

**Electrical Characteristics (continued)**

( $V_{SYS} = 3.8V$ ,  $V_{BATT} = 3.8V$ ,  $V_{VIO} = 1.8V$ ,  $V_{CHGIN} = 5V$ , unless otherwise specified. Limits are production tested at  $T_A = +25^\circ C$ . Limits over the operating temperature range and relevant supply voltage range are guaranteed by design and characterization.)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
Fast-Charge Current Thermal Regulation Gain	$A_{TJREG}$	The charge current is decreased by 5.73% of the fast-charge current full-scale for every degree that the junction temperature exceeds the thermal regulation temperature. This slope ensures that the full-scale current of 5.5A is reduced to 0A by the time the junction temperature is $+17.5^\circ C$ above the programmed loop set point. For lower programmed charge currents such as 480mA, this slope is valid for charge current reductions down to 80mA; below 100mA the slope becomes shallower but the charge current is reduced to 0A if the junction temperature is $+20^\circ C$ above the programmed loop set point.		-315		$mA^\circ C$
Fast-Charge Termination Voltage Setting Range (Variant A)	$V_{BATTREG}$	Programmable from 4.15V to 4.5375V in 12.5mV steps (5-bits); production tested at 4.2V and 4.35V only (MAX77985A and MAX77986A)	4.150	4.5375		V
Fast-Charge Termination Voltage Setting Range (Variant B)	$V_{BATTREG}$	Programmable from 3.500V to 4.275V in 25mV steps (5-bits); production tested at 3.6V and 4.2V only (MAX77985B and MAX77986B)	3.500	4.275		V
Fast-Charge Termination Voltage Accuracy at Room Temp (Variant A)	$V_{BATTREG\_AC\_C}$	$V_{BATTREG} = 4.35V$ setting, represented as percentage of $V_{BATTREG}$ ; $T_A = +25^\circ C$	-0.6	-0.3	+0.0	%
Fast-Charge Termination Voltage Accuracy (Variant A)	$V_{BATTREG\_AC\_C}$	$V_{BATTREG} = 4.35V$ setting, represented as percentage of $V_{BATTREG}$ ; $T_A = -5^\circ C$ to $+85^\circ C$	-0.8	-0.3	+0.2	%
Fast-Charge Termination Voltage Accuracy at Room Temp (Variant B)	$V_{BATTREG\_AC\_C}$	$V_{BATTREG} = 4.20V$ setting, represented as percentage of $V_{BATTREG}$ ; $T_A = +25^\circ C$	-0.6	-0.3	+0.0	%
Fast-Charge Termination Voltage Accuracy (Variant B)	$V_{BATTREG\_AC\_C}$	$V_{BATTREG} = 4.20V$ setting, represented as percentage of $V_{BATTREG}$ ; $T_A = -5^\circ C$ to $+85^\circ C$	-0.8	-0.3	+0.2	%
Fast-Charge Termination Debounce Time	$t_{TERM}$	( <a href="#">Note 3</a> )		100		ms
Fast-Charge Constant Current + Constant Voltage Safety Time	$t_{FC}$	Adjustable from 3hrs, 4hrs, 5hrs, 6hrs, 7hrs, and 8hrs including a disable setting; 5hrs default		5		hrs
Top-Off Current Setting Range	$I_{TO}$	Programmable from 150mA to 850mA with 50mA in 16 steps; production tested at 150mA, 200mA, 500mA, and 850mA settings	150	850		mA

### Electrical Characteristics (continued)

( $V_{SYS} = 3.8V$ ,  $V_{BATT} = 3.8V$ ,  $V_{VIO} = 1.8V$ ,  $V_{CHGIN} = 5V$ , unless otherwise specified. Limits are production tested at  $T_A = +25^\circ C$ . Limits over the operating temperature range and relevant supply voltage range are guaranteed by design and characterization.)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
Top-Off Current Accuracy	$I_{TO\_ACC}$	150mA setting	122.5	150	177.5	mA
		200mA setting	170	200	230	
		500mA setting	455	500	545	
		850mA setting	787.5	850	912.5	
Top-Off Time	$t_{TO}$	Adjustable from 30sec to 70min in 10min steps; default setting is 30min		30		min
Charge Restart Threshold Setting Range	$V_{RSTRT}$	Adjustable at 100mV, 150mV, and 200mV; it can also be disabled	100	150	200	mV
Charge Restart Debounce Time	$t_{CRDG}$			130		ms
Charge State Change Interrupt Debounce Time	$t_{SCIDG}$	Excludes transition to timer fault state, watchdog timer state		30		ms
Charge Watchdog Time	$t_{WD}$			80		s
Charge Timers Accuracy	$t_{ACC}$		-20		+20	%
Charge-Overshoot Threshold	$V_{COV}$	$V_{BAT\_SP} - V_{BAT\_SN}$ , relative to $V_{CHG\_CV\_PRM}$		200		mV
Remote Sense BAT_SP Input Current in Charging Mode	$I_{BAT\_SP\_CHG}$	$V_{BATT\_SP} = V_{BATT} = 3.8V$ , MODE = 5, $T_A = +25^\circ C$		14		$\mu A$
Remote Sense BAT_SN Input Current in Charging Mode	$I_{BAT\_SN\_CHG}$	$V_{BATT\_SN} = 0$ , MODE = 5, $T_A = +25^\circ C$		10		$\mu A$
<b>SMART POWER SELECTOR</b>						
System Regulation Voltage Setting Range (Charging Enabled, Low Battery, Variant A)	$V_{SYSMIN\_RNG}$	Charging enabled, $V_{BATT} < V_{SYSMIN}$ , programmable from 3.4V to 3.7V in 0.1V steps (MAX77985A and MAX77986A)	3.4		3.7	V
System Regulation Voltage Setting Range (Charging Enabled, Low Battery, Variant B)	$V_{SYSMIN\_RNG}$	Charging enabled, $V_{BATT} < V_{SYSMIN}$ , programmable from 3.0V, 3.1V, 3.5V or 3.6V (MAX77985B and MAX77986B)	3.0		3.6	V
System Regulation Voltage Accuracy (Charging Enabled, Low Battery)	$V_{SYSMIN\_ACC}$	Charging enabled, $V_{BATT} < V_{SYSMIN}$ , production tested at 3.60V only	-3		+3	%

## Electrical Characteristics (continued)

( $V_{SYS} = 3.8V$ ,  $V_{BATT} = 3.8V$ ,  $V_{VIO} = 1.8V$ ,  $V_{CHGIN} = 5V$ , unless otherwise specified. Limits are production tested at  $T_A = +25^\circ C$ . Limits over the operating temperature range and relevant supply voltage range are guaranteed by design and characterization.)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
System Regulation Voltage (Charging Enabled, Low Battery)	$V_{SYSTRK}$	Charging enabled, $V_{SYSMIN} - V_{SYSTRK} < V_{BATT} < V_{SYSMIN}$ , measure of $V_{SYS} - V_{BATT}$ , $V_{SYSREG\_TRK}$ represented as a percentage of $V_{BATT}$ [For MINSYS settings 3.0V, 3.1V, 3.2V, 3.3V] ( <a href="#">Note 3</a> )		15.5		%
		Charging enabled, $V_{SYSMIN} - V_{SYSTRK} < V_{BATT} < V_{SYSMIN}$ , measure of $V_{SYS} - V_{BATT}$ , $V_{SYSREG\_TRK}$ represented as a percentage of $V_{BATT}$ [For MINSYS settings 3.4V, 3.5V, 3.6V, 3.7V] ( <a href="#">Note 3</a> )		13.5		
BATT to SYS Reverse Regulation Voltage	$V_{BSREG}$	Measure of $V_{SYS} - V_{BATT}$ ; production tested at 10mA and 2A	-100			mV
SYS Self-Discharge Resistor	$R_{SYSSD}$	Switching is disabled, $Q_{BATT}$ FET is off, $V_{SYS} < V_{SYSUVLO\_F}$	600			$\Omega$
<b>BATTERY OVERCURRENT PROTECTION</b>						
Battery Overcurrent Protection Quiescent Current	$I_{Q\_OVRC}$	$B2SOVRC\_CTRL = 0$ ; $I_{BATT}$ represented in units of $\mu A$	3 + $I_{BATT}/75000$			$\mu A$
Battery Overcurrent Protection Setting Range	$I_{BOVRC}$	Programmable from 3A to 10A with 0.5A steps; can be disabled	3	10		A
Battery Overcurrent Protection Accuracy	$I_{BOVRC\_ACC}$	$B2SOVRC$ setting 0x4 (4.5A) and below; production tested at 3.0A setting	-15	+15		%
		$B2SOVRC$ setting 0x5 (5.0A) and above; production tested at 5.0A setting	-10	+10		
Battery Overcurrent Debounce Time	$t_{BOVRC}$	$B2SOVRC\_CTRL = 1$ ; time between battery over-current event and $BAT\_I$ interrupt generation	105			$\mu s$
Battery Overcurrent Delay	$t_{OCP}$	Delay from IRQB toggling low to $Q_{BATT}$ FET opening ( $B2SOVRC\_DTC = 0$ )	105			$\mu s$
		Delay from IRQB toggling low to $Q_{BATT}$ FET opening ( $B2SOVRC\_DTC = 1$ )	10			ms
Battery Overcurrent Retry Timer	$t_{OCP\_RETRY}$	Retry is one time	150			ms
System Power-Up Current (from BATT)	$I_{SYSPU\_BAT}$	$V_{CHGIN} = 0V$	35	50	80	mA
System Power-Up Voltage (from BATT)	$V_{SYSPU\_BAT}$	$V_{SYS}$ rising, 100mV hysteresis	1.9	2.0	2.1	V
System Power-Up Time-Out (from BATT)	$t_{SYSPU\_BAT}$		150			ms

**Electrical Characteristics (continued)**

( $V_{SYS} = 3.8V$ ,  $V_{BATT} = 3.8V$ ,  $V_{VIO} = 1.8V$ ,  $V_{CHGIN} = 5V$ , unless otherwise specified. Limits are production tested at  $T_A = +25^\circ C$ . Limits over the operating temperature range and relevant supply voltage range are guaranteed by design and characterization.)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
<b>REVERSE BOOST</b>						
Reverse Boost Quiescent Current		$V_{BYP} = 5.1V$ , $V_{BATT} = 3.8V$ , MODE = 0x0A, $V_{BYPSET} = 0x1$		2.5		mA
Reverse Boost Output Voltage Setting Range	$V_{BYP\_OTG}$	Measured on BYP pin, $2.5V < V_{BATT} < 4.5V$ ; adjustable from 5V to 12V with 0.1V step; production tested at 5V and 12V	5		12	V
Reverse Boost Output Voltage Accuracy	$V_{BYP\_ACC}$	Measured on BYP, MODE = 0x0A, $V_{BYPSET} = 0x1$	4.95	5.10	5.25	V
Reverse Boost Inductor Current Limit	$I_{LSILIM}$	For MAX77986	8.5	9.5	10.5	A
	$I_{LSILIM}$	For MAX77985	5.95	7.00	8.05	
<b>CHGIN OUTPUT LIMITER</b>						
OTG Output Current Limit Setting Range (MAX77985)	$I_{CHGIN\_OTG\_LIM}$	Configurable from 500mA to 2400mA in 100mA steps; clamped to 12W power limit	500		2400	mA
OTG Output Current Limit Setting Range (MAX77986)	$I_{CHGIN\_OTG\_LIM}$	Configurable from 500mA to 3100mA in 100mA steps; clamped to 18W power limit	500		3100	mA
OTG Output Current Limit	$I_{CHGIN\_OTG\_LIM}$	3.4V < $V_{BATT} < 4.5V$ , OTG_ILIM = 0x00	500	537	575	mA
		3.4V < $V_{BATT} < 4.5V$ , OTG_ILIM = 0x04	900	967	1035	
		3.4V < $V_{BATT} < 4.5V$ , OTG_ILIM = 0x0A	1500	1612	1725	
		3.4V < $V_{BATT} < 4.5V$ , OTG_ILIM = 0x19 (MAX77985 only)	2400	2580	2760	
		3.4V < $V_{BATT} < 4.5V$ , OTG_ILIM = 0x19 (MAX77986 only)	3000	3225	3450	
OTG Output Current Limit Alarm Time	$t_{OTG\_ALARM}$	Delay from OTG overcurrent event to BYP_I interrupt generated		20		ms
OTG Output Current Limit Fault Time	$t_{OTG\_FAULT}$	Delay from OTG overcurrent event to QCHGIN FET opening		30		ms
OTG Output Current Limit Retry Time	$t_{OTG\_RETRY}$	Delay from QCHGIN FET opening to QCHGIN FET closing again (OTG_REC_EN = 1)		300		ms
<b>SWITCH IMPEDANCES AND LEAKAGE CURRENTS</b>						
CHGIN to BYP On Resistance at Room Temp	$R_{CHGIN2BYP\_ROOM}$	CHGIN pin to BYP pin, $T_A = +25^\circ C$		14.3	18.6	$m\Omega$
CHGIN to BYP On Resistance	$R_{CHGIN2BYP}$	CHGIN pin to BYP pin, $T_A = -40^\circ C$ to $+85^\circ C$		14.3	22.0	$m\Omega$
LX High-Side On Resistance at Room Temp	$R_{HS\_ROOM}$	BYP pin to LX pin, $T_A = +25^\circ C$		31.0	43.4	$m\Omega$
LX High-Side On Resistance	$R_{HS}$	BYP pin to LX pin, $T_A = -40^\circ C$ to $+85^\circ C$		31.0	54.3	$m\Omega$
LX Low-Side On Resistance at Room Temp	$R_{LS\_ROOM}$	LX pin to PGND pin, $T_A = +25^\circ C$		16.0	22.4	$m\Omega$

### Electrical Characteristics (continued)

( $V_{SYS} = 3.8V$ ,  $V_{BATT} = 3.8V$ ,  $V_{VIO} = 1.8V$ ,  $V_{CHGIN} = 5V$ , unless otherwise specified. Limits are production tested at  $T_A = +25^\circ C$ . Limits over the operating temperature range and relevant supply voltage range are guaranteed by design and characterization.)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
LX Low-Side On Resistance	$R_{LS}$	LX pin to PGND pin, $T_A = -40^\circ C$ to $+85^\circ C$		16.0	28.0	$m\Omega$
BATT to SYS On Resistance at Room Temp	$R_{BAT2SYS\_RO}$ $OM$	BATT pin to SYS pin, $V_{BATT} = 4.4V$ , $T_A = +25^\circ C$		7.70	10.5	$m\Omega$
BATT to SYS On Resistance	$R_{BAT2SYS}$	BATT pin to SYS pin, $V_{BATT} = 4.4V$ , $T_A = -40^\circ C$ to $+85^\circ C$		7.70	12.75	$m\Omega$
LX Leakage Current	$I_{LX\_LEAK}$	$V_{LX} = V_{PGND}$ or $V_{BYP}$ , $T_A = +25^\circ C$	0.01	10		$\mu A$
		$V_{LX} = V_{PGND}$ or $V_{BYP}$ , $T_A = +85^\circ C$	1			
BST Leakage Current	$I_{BST\_LEAK}$	$V_{BST} - V_{LX} = 1.8V$ , $T_A = +25^\circ C$	0.01	10		$\mu A$
		$V_{BST} - V_{LX} = 1.8V$ , $T_A = +85^\circ C$	1			
BYP Leakage Current	$I_{BYP\_LEAK}$	$V_{BYP} = 5.5V$ , $V_{CHGIN} = 0V$ , $V_{LX} = 0V$ , charger disabled, $T_A = +25^\circ C$	0.01	10		$\mu A$
		$V_{BYP} = 5.5V$ , $V_{CHGIN} = 0V$ , $V_{LX} = 0V$ , charger disabled, $T_A = +85^\circ C$	1			
BATSP Input Current Leakage	$I_{BATSP}$	Charger disabled, $V_{BATSP} = V_{BATT}$ , $T_A = +25^\circ C$		$\pm 1$		$\mu A$
BATSN Input Current Leakage	$I_{BATSN}$	Charger disabled, $V_{BATSN} = V_{AGND}$ , $T_A = +25^\circ C$		$\pm 1$		$\mu A$

#### LOGIC AND CONTROL I/Os

Input Low Level	$V_{IL}$	SUSPND, DISQBAT, $T_A = +25^\circ C$	0.4		$V$
		EXTSM, $T_A = +25^\circ C$	0.3 x $V_{BATT}$		
Input High Level	$V_{IH}$	SUSPND, DISQBAT, $T_A = +25^\circ C$	1.4		$V$
		EXTSM, $T_A = +25^\circ C$	0.7 x $V_{BATT}$		
Input Leakage Current	$I_{LK}$	SUSPND, DISQBAT, EXTSM pin, at 5.5V (including current through pulldown resistor)	24	60	$\mu A$
Output Low Voltage QBEXT	$V_{OLQBEXT}$	Sourcing 1mA, $T_A = +25^\circ C$		0.4	$V$
Output High Leakage QBEXT	$I_{LQBEXT}$	$V_{SYS} = 5.5V$ , $T_A = +25^\circ C$	-1	0	$\mu A$
		$V_{SYS} = 5.5V$ , $T_A = +85^\circ C$	0.1		
SUSPND Internal Pulldown Resistor	$R_{SUSPND}$		235		$k\Omega$
DISQBAT Internal Pulldown Resistor	$R_{DISQBAT}$		235		$k\Omega$
EXTSM Internal Pulldown Resistor	$R_{EXTSM}$		235		$k\Omega$
EXTSM Debounce Time	$t_{EXTSM\_DEB}$	$V_{BATT}$ in 3.3V to 4.5V range, $EXTSM\_T = 0$	10		$ms$
		$V_{BATT}$ in 3.3V to 4.5V range, $EXTSM\_T = 1$	0.1		

**Electrical Characteristics (continued)**

( $V_{SYS} = 3.8V$ ,  $V_{BATT} = 3.8V$ ,  $V_{VIO} = 1.8V$ ,  $V_{CHGIN} = 5V$ , unless otherwise specified. Limits are production tested at  $T_A = +25^\circ C$ . Limits over the operating temperature range and relevant supply voltage range are guaranteed by design and characterization.)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
<b>CHARGE STATUS INDICATOR</b>						
Charge Status Current Setting Range	ISTAT_RNG	5mA to 20mA in 5mA steps; production tested at $V_{STAT} - V_{AGND} = 1.0V$ and 5.0V	5		20	mA
Charge Status Current Accuracy	ISTAT_ACC	Production tested at 5mA and 20mA	-15		+15	%
<b>THERMISTOR MONITOR</b>						
THM Threshold, COLD	THM_COLD	$V_{THM}/V_{PVDD}$ rising, 1% hysteresis (thermistor temperature falling)	73.8	75.0	76.2	%
THM Threshold, COOL	THM_COOL	$V_{THM}/V_{PVDD}$ rising, 1% hysteresis (thermistor temperature falling)	64.3	65.5	66.7	%
THM Threshold, WARM	THM_WARM	$V_{THM}/V_{PVDD}$ falling, 1% hysteresis (thermistor temperature rising)	30.8	32.0	33.2	%
THM Threshold, HOT	THM_HOT	$V_{THM}/V_{PVDD}$ falling, 1% hysteresis (thermistor temperature rising)	20.8	22.0	23.2	%
THM Threshold, Disabled	THM_DIS	$V_{THM}/V_{PVDD}$ falling, 1% hysteresis, THM function is disabled below this voltage	4.8	6.0	7.2	%
THM Threshold, Battery Removal Detection	THM_RM	$V_{THM}/V_{PVDD}$ rising, 1% hysteresis, battery removal	85	87	89	%
THM Input Leakage Current	ILKTHM	$V_{THM} = V_{AGND}$ or $V_{PVDD}$ , charger disabled, $T_A = +25^\circ C$		0.1	1	$\mu A$
		$V_{THM} = V_{AGND}$ or $V_{PVDD}$ , charger disabled, $T_A = +85^\circ C$		0.1		
<b>SUPPLIES AND MONITORING</b>						
VDD Output Voltage	V <sub>VDD_1P8</sub>	$V_{SYS}$ or $V_{BATT} = 3.8V$ , $I_{VDD} = 20mA$	1.71	1.80	1.89	V
SYS Undervoltage-Lockout Threshold (SYS Rising)	V <sub>SYS_UVLO_R</sub>		2.74	2.80	2.86	V
SYS Undervoltage-Lockout Threshold (SYS Falling)	V <sub>SYS_UVLO_F</sub>		2.55	2.60	2.65	V
SYS Undervoltage-Lockout Hysteresis	V <sub>SYS_UVLO_H</sub>			200		mV
SYS Overvoltage-Lockout Threshold (SYS Rising)	V <sub>SYS_OVLO_R</sub>	SYS rising	5.350	5.425	5.500	V
SYS Overvoltage-Lockout Threshold (SYS Falling)	V <sub>SYS_OVLO_F</sub>	SYS falling	5.200	5.275	5.350	V
SYS Overvoltage-Lockout Hysteresis	V <sub>SYS_OVLO_H</sub>			150		mV
Thermal Shutdown Threshold	T <sub>SHDN_R</sub>	T <sub>j</sub> rising		155		°C
Thermal Shutdown Threshold Hysteresis	T <sub>SHDN_H</sub>			15		°C

**Electrical Characteristics (continued)**

( $V_{SYS} = 3.8V$ ,  $V_{BATT} = 3.8V$ ,  $V_{VIO} = 1.8V$ ,  $V_{CHGIN} = 5V$ , unless otherwise specified. Limits are production tested at  $T_A = +25^\circ C$ . Limits over the operating temperature range and relevant supply voltage range are guaranteed by design and characterization.)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
PVDD Output Voltage	$V_{PVDD\_1P8}$	$V_{SYS} = 3.8V$ , $I_{PVDD} = 20mA$	1.71	1.80	1.89	V
<b>I<sup>2</sup>C-COMPATIBLE INTERFACE TIMING FOR STANDARD, FAST, AND FAST-MODE PLUS</b>						
Clock Frequency	$f_{SCL}$			1000		kHz
Hold Time (Repeated) START Condition	$t_{HD;STA}$		0.26			μs
CLK Low Period	$t_{LOW}$		0.5			μs
CLK High Period	$t_{HIGH}$		0.26			μs
Set-Up Time Repeated START Condition	$t_{SU;STA}$		0.26			μs
DATA Hold Time	$t_{HD;DAT}$		0			μs
DATA Valid Time	$t_{VD;DAT}$			0.45		μs
DATA Valid Acknowledge Time	$t_{VD;ACK}$			0.45		μs
DATA Set-Up time	$t_{SU;DAT}$		50			ns
Set-Up Time for STOP Condition	$t_{SU;STO}$		0.26			μs
Bus-Free Time Between STOP and START	$t_{BUF}$		0.5			μs
Pulse Width of Spikes that must be Suppressed by the Input Filter	$t_{SP}$			50		ns
<b>I<sup>2</sup>C-COMPATIBLE INTERFACE TIMING FOR HS-MODE (CB = 100pF)</b>						
Clock Frequency	$f_{SCL}$			3.4		MHz
Set-Up Time Repeated START Condition	$t_{SU;STA}$		160			ns
Hold Time (Repeated) START Condition	$t_{HD;STA}$		160			ns
CLK Low Period	$t_{LOW}$		160			ns
CLK High Period	$t_{HIGH}$		60			ns
DATA Set-Up time	$t_{SU;DAT}$		10			ns
DATA Hold Time	$t_{HD;DAT}$		0			ns
Set-Up Time for STOP Condition	$t_{SU;STO}$		160			ns
Pulse Width of Spikes that must be Suppressed by the Input Filter	$t_{SP}$			10		ns
<b>I<sup>2</sup>C-COMPATIBLE INTERFACE TIMING FOR HS-MODE (CB = 400pF)</b>						
Clock Frequency	$f_{SCL}$			1.7		MHz
Set-Up Time Repeated START Condition	$t_{SU;STA}$		160			ns

## Electrical Characteristics (continued)

( $V_{SYS} = 3.8V$ ,  $V_{BATT} = 3.8V$ ,  $V_{VIO} = 1.8V$ ,  $V_{CHGIN} = 5V$ , unless otherwise specified. Limits are production tested at  $T_A = +25^\circ C$ . Limits over the operating temperature range and relevant supply voltage range are guaranteed by design and characterization.)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
Hold Time (Repeated) START Condition	$t_{HD:STA}$		160			ns
CLK Low Period	$t_{LOW}$		320			ns
CLK High Period	$t_{HIGH}$		120			ns
DATA Set-Up time	$t_{SU:DAT}$		10			ns
DATA Hold Time	$t_{HD:DAT}$		0			ns
Set-Up Time for STOP Condition	$t_{SU:STO}$		160			ns
Pulse Width of Spikes that must be Suppressed by the Input Filter	$t_{SP}$			10		ns

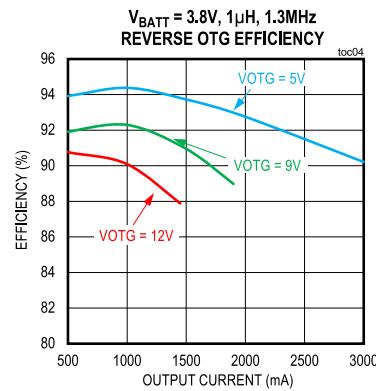
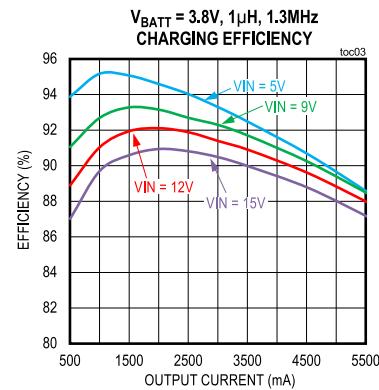
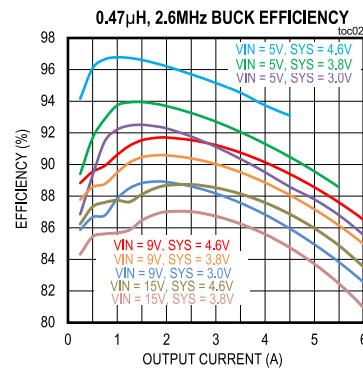
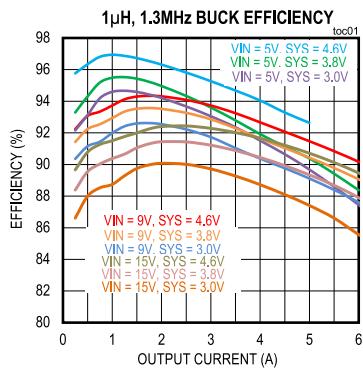
**Note 1:** See the [Unplug Detection](#) section for more information on Case1.

**Note 2:** See the [Unplug Detection](#) section for more information on Case2.

**Note 3:** For parts with chip revision 0b001 (PASS1), contact Analog Devices for Rev 0 of the data sheet. This information can be read from the CHIP\_REVISION (0x01) register.

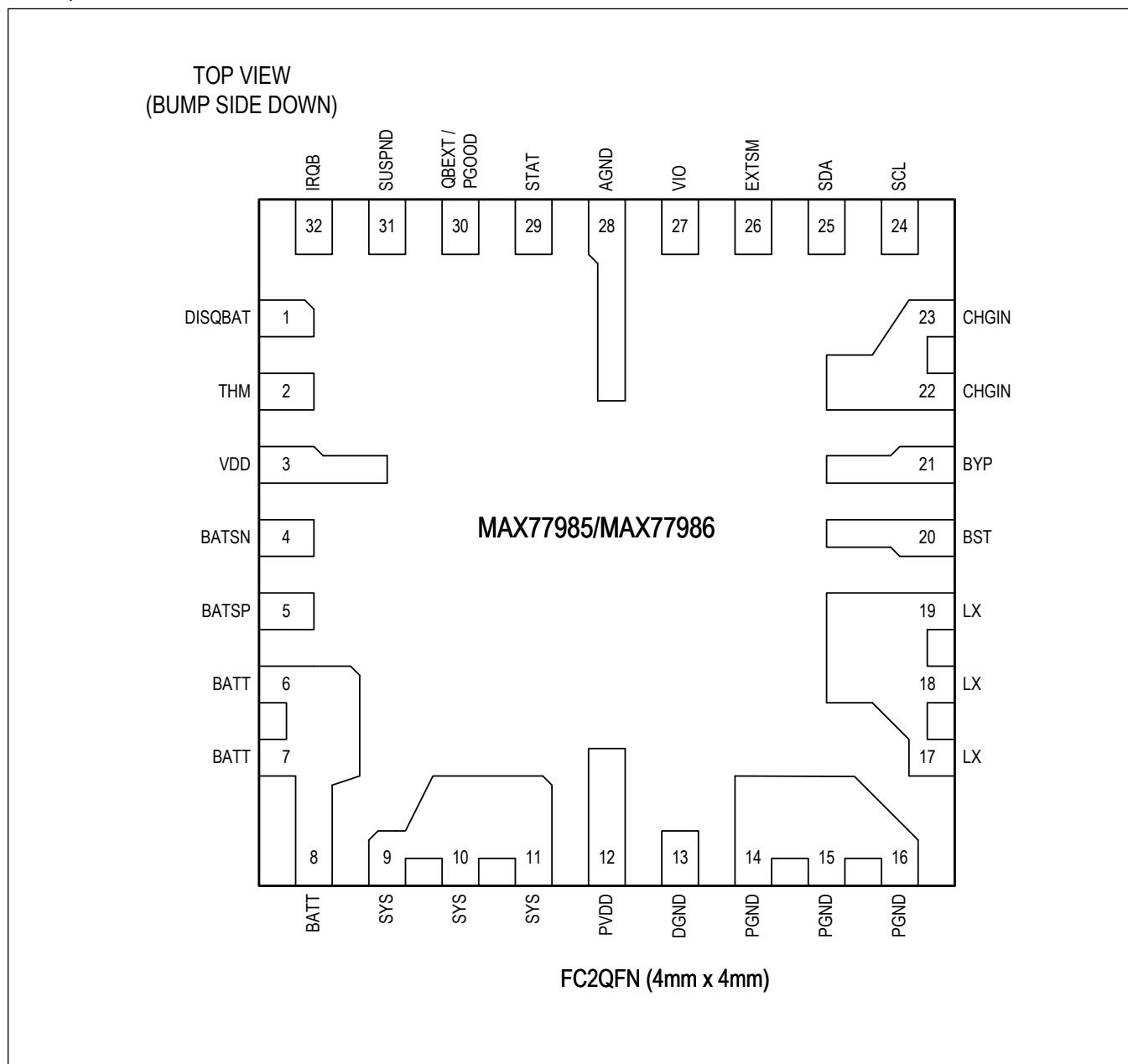
## Typical Operating Characteristics

( $T_A = +25^\circ\text{C}$ , unless otherwise noted.)



## Pin Configuration

### FC2QFN



## Pin Description

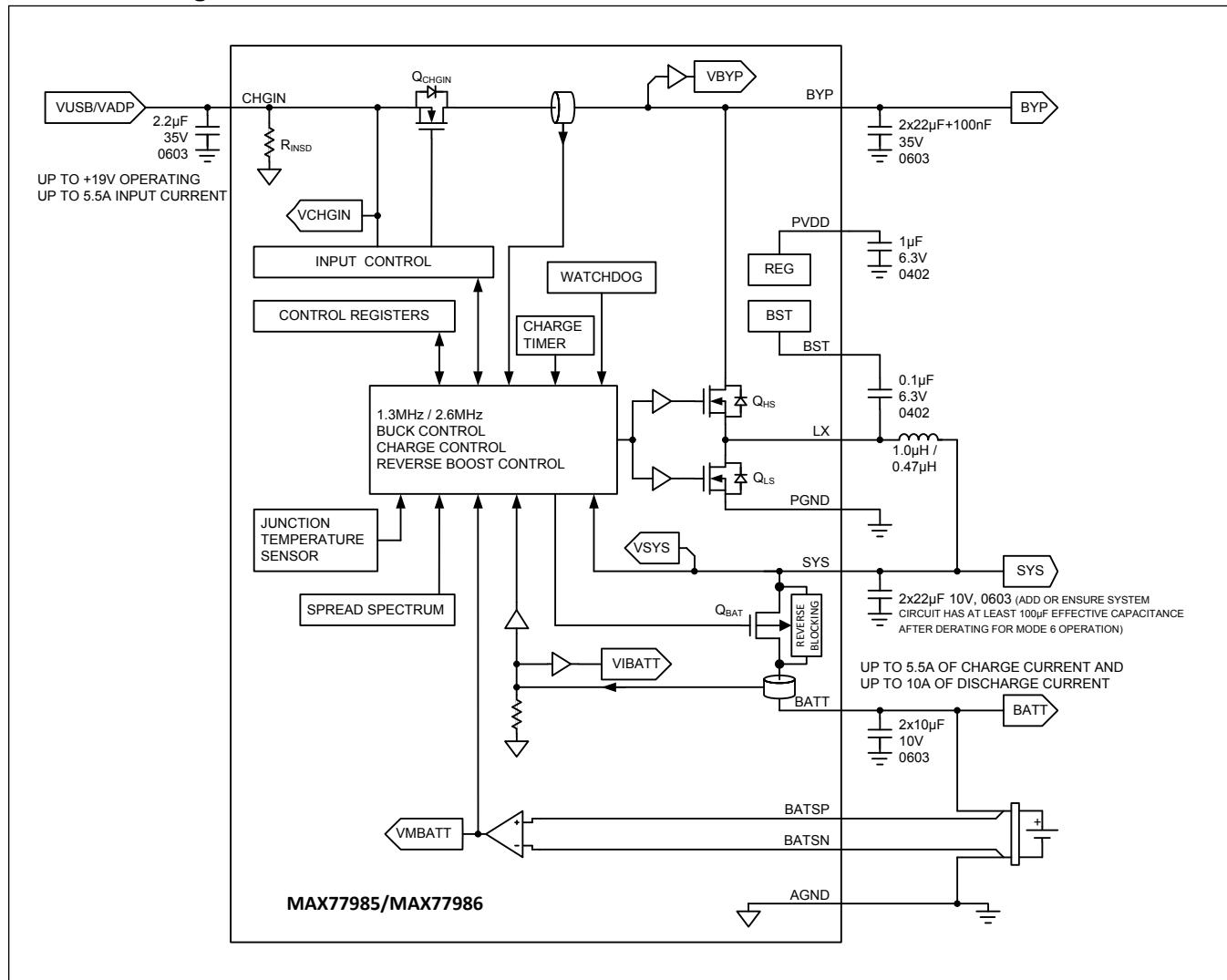
PIN	NAME	FUNCTION	TYPE
1	DISQBAT	Active-high to disable internal Q <sub>BATT</sub> FET between SYS and BATT.	DI
2	THM	Thermistor Connection. Connect an external thermistor between THM and AGND.	A

**Pin Description (continued)**

PIN	NAME	FUNCTION	TYPE
3	VDD	Analog Voltage Level. The output of on-chip low voltage LDO used to power on-chip, low-noise circuits. Bypass with a 1 $\mu$ F (6.3V) ceramic capacitor to AGND.  Powering external loads from VDD is not recommended, other than pullup resistors.	A
4	BATSN	Battery Negative Differential Sense Connection. Connect to the negative or ground terminal as close as possible.	A
5	BATSP	Battery Positive Differential Sense Pin. Connect to battery positive terminal as close as possible to eliminate errors due to trace/connector voltage drops.	A
6, 7, 8	BATT	Connection with Battery. Connect to the positive terminal of a single-cell Li-ion battery. Bypass with 2 x 10 $\mu$ F (6.3V) ceramic capacitors from BATT to PGND.	P
9, 10, 11	SYS	Connection with System. Bypass with at least 2x22 $\mu$ F (6.3V) ceramic capacitors from SYS to PGND. This ensures that the minimum effective capacitance on the SYS node is 12 $\mu$ F (effective), for stability purposes. If Mode 0x6 is used, then it additionally requires more ceramic capacitors to give a total effective capacitance of at least 100 $\mu$ F on the SYS node. The SYS node capacitance can increase up to 350 $\mu$ F total (effective).	P
12	PVDD	Internal Bias Regulator High Current Output Bypass Pin. Supplies internal noisy and high current gate drive loads. Bypass with 1 $\mu$ F (6.3V) from PVDD to PGND.  Powering external loads from PVDD is not recommended, other than pullup resistors.	P
13	DGND	Digital Ground	A
14, 15, 16	PGND	Charger Power Ground	P
17, 18, 19	LX	Charger Switching Node. Connect the inductor between LX and SYS.	P
20	BST	High-Side FET Driver Supply. Bypass BST to LX with a 1x 100nF (6.3V) ceramic capacitor.	A
21	BYP	CHGIN Bypass Pin. This pin is the input for the switching charger and the output for the boost converter when the charger is operating in 'reverse-boost' mode. Bypass with 2 x 22 $\mu$ F (35V) + 100nF (35V) ceramic capacitor from BYP to PGND.	P
22, 23	CHGIN	Charger Input. Connect a 2.2 $\mu$ F (35V) capacitor between CHGIN and PGND.	P
24	SCL	I <sup>2</sup> C Interface Clock Input	DI
25	SDA	I <sup>2</sup> C Interface Data Input	DI
26	EXTSM	Exit Ship Mode Input by Push-Button. Active-high input.	DI
27	VIO	I <sup>2</sup> C Supply Voltage Input. Bypass to AGND with a 1 $\mu$ F (6.3V) capacitor.	P
28	AGND	Analog Ground	A
29	STAT	LED Low-Side Driver Output for Indicating Charging Status	A
30	QBEXT/ PGOOD	When UNPLG_DET = 0x00, this pin is configured to QBEXT which is the external battery FET control output. Connect a pullup resistor to VIO, SYS, or BATT supply.  When UNPLG_DET ≠ 0x00, this pin is configured to PGOOD which is the unplug detection comparator output.	DO
31	SUSPND	Active-High Input to Disable the DC-DC Between CHGIN Input and SYS Output	DI
32	IRQB	Interrupt Output. Connect a 100k $\Omega$ pullup resistor between IRQB and VIO.	DO

## Functional Diagrams

### Functional Diagram



## Detailed Description

### Switching Mode Charger

#### Features

- Complete Li+/LiPoly Battery Charger
  - Prequalification, Constant Current, Constant Voltage
  - 55mA Precharge Current
  - 300mA Trickle Charge Current
  - Adjustable Constant Current Charge
    - 100mA to 5.5A in 50mA Steps (Limited to 3.5A for MAX77985)
  - Adjustable Charge-Termination Threshold
    - 150mA to 850mA in 50mA Steps
  - Adjustable Battery Regulation Voltage
    - 4.15V to 4.5375V in 12.5mV Steps for MAX7798xA
    - 3.50V to 4.275V in 25mV Steps for MAX7798xB
    - -0.8/+0.2% Accuracy from -5°C to +85°C
  - Remote Differential Sensing
- Synchronous Switch-Mode Based Design
- Smart Power Selector
  - Optimally distributes power between the charge adapter, system, and battery.
  - When powered by a charge adapter, the battery can provide supplemental current to the system.
  - The charge adapter can support the system with a dead battery or without a battery.
- No External MOSFETs Required for Switcher
- CHGIN Input
  - Adjustable Input Current Limit
    - 100mA to 5.5A in 50mA Steps (CHGIN\_ILIM)
    - Default is set to 500mA
  - Supports AC-to-DC Wall Adapters
  - VCHGIN\_OVLO = 19.5V
  - Reverse-Leakage Protection Prevents the Battery Leaking Current to the Inputs
- Charge Safety Watchdog Timer
  - Selectable: 3hr to 8hr, plus a Disable Setting
- Die Temperature Monitor with Thermal Foldback Loop
  - Selectable Die-Temperature Thresholds (°C): +85°C to +130°C in +5°C Steps
- Input Voltage Dropout Control Allows Operation from High-Impedance Sources (AICL)
- BATT to SYS Switch is 7.7mΩ Typical
  - Capable of up to 10A Steady-State Operation from BATT to SYS
- Short-Circuit Protection
  - Programmable BATT to SYS Overcurrent Threshold from 3A to 10A, plus a Disable Setting
  - DISIBS Bit Allows the Host to Disable the Battery to System Discharge Path to Protect Against a Short-Circuit
  - SYS Short to Ground
    - Buck current is limited by switcher current limit and disabling of the synchronous rectifier.
    - BATT currents above the programmed by B2SOVRC threshold generate an interrupt. The host can then disable the battery to the system discharge path by setting DISIBS or asserting the DISQBAT pin high.
- Fast Unplug Detection of 5V, 9V and 15V Sources
- Spread Spectrum Modulation for Reduced EMI

### Detailed Description

The MAX77985/MAX77986 includes a full-featured switch-mode charger for a one-cell lithium-ion (Li+) or lithium-polymer (Li-polymer) battery. The current limit for CHGIN input is independently programmable from 100mA to 3.5A/5.5A in 50mA steps allowing the flexibility for connection to either an AC-to-DC wall charger or a USB port.

The synchronous switch-mode DC-DC converter utilizes a high 1.3MHz/2.6MHz switching frequency which is ideal for portable devices because it allows the use of small components while eliminating excessive heat generation. The DC-DC has both a buck and a boost mode of operation. When charging the main battery, the converter operates as a buck. The DC-DC buck operates from a 4.3V to 19.5V source. The battery charge current is programmable from 100mA to 3.5A/5.5A in MAX77985/MAX77986.

As a boost converter, the DC-DC uses energy from the main battery to boost the voltage at BYP. The BYP supplies the USB OTG voltage (5.1V) and USB Type-C® PD Source Voltages (5V to 12V). The programmable boost output current limit range is from 0.5A to 3.1A with a 0.1A step.

The Smart Power Selector architecture makes the best use of the limited adapter power and the battery's power at all times to supply up to buck current limit from the buck to the system. (Additionally, supplement mode provides additional current from the battery to the system up to B2SOVRC.) Adapter power that is not used for the system goes to charging the battery. All power switches for charging and switching the system load between the battery and adapter power are included on-chip—no external MOSFETs are required.

A multitude of safety features ensures reliable charging. Features include a charge timer, watchdog, junction thermal regulation, over/under voltage protection, and short circuit protection.

The BATT to SYS switch has overcurrent protection (see the [Main-Battery Overcurrent Protection Due to Fault](#) section for more information).

Recommended buck output current range is as follows:

**Table 1. Recommended Buck Output Current Range**

	MINIMUM SYS CURRENT (A)	MAXIMUM SYS CURRENT (A)
MAX77985A, MAX77985B	0	4.5
MAX77986A, MAX77986B	0	7

### Smart Power Selector (SPS)

The SPS architecture is a network of internal switches and control loops that distribute energy between external power sources CHGIN, BYP, SYS, and BATT.

The [Functional Diagram](#) shows a more detailed arrangement of the Smart Power Selector switches and gives them the following names: QCHGIN, QHS, QLS, QBATT.

### Switch and Control Loop Descriptions

- CHGIN Input Switch: The input switch is either completely on or completely off. As shown in the [Functional Diagram](#), there are SPS control loops that monitor the current through the input switches as well as the input voltage.
- DC-DC Switches: Q<sub>HS</sub> and Q<sub>LS</sub> are the DC-DC switches that can operate as a buck (step-down) or a boost (step-up). When operating as a buck, energy is moved from BYP to SYS. When operating as a boost, energy is moved from SYS to BYP. SPS control loops monitor the DC-DC switch current, the SYS voltage, and the BYP voltage.
- Battery-to-System Switch: Q<sub>BATT</sub> controls the battery charging and discharging. Additionally, Q<sub>BATT</sub> allows the battery to be isolated from the system (SYS). An SPS control loop monitors the Q<sub>BATT</sub> current.

### Control Bits

- MODE configures the Smart Power Selector
- V<sub>BYPSET</sub> sets the BYP regulation voltage target
- B2SOVRC configures the main-battery overcurrent protection

### Energy Distribution Priority

- With a valid external power source:
  - The external power source is the primary source of energy

- The main battery is the secondary source of energy
- Energy delivery to BYP is the highest priority
- Energy delivery to SYS is the second priority
- Any energy that is not required by BYP or SYS is available to the main-battery charger
- With no power source available at CHGIN:
  - The main-battery is the primary source of energy
  - Energy delivery to BYP (if boost mode is selected) and SYS share the same priority
    - BYP includes CHGIN if boost OTG mode is selected, itself limited by OTG\_ILIM threshold
    - Energy delivery to BYP (if boost mode is selected) and SYS is limited by B2SOVRC threshold

#### BYP Regulation Voltage

- When the DC-DC is off or in one of its buck modes and there is a valid power source at CHGIN,  $V_{BYP} = V_{CHGIN} - I_{CHGIN} \times R_{CHGIN2BYP}$ .
- When the DC-DC is off and there is no valid power source at CHGIN, BYP is connected to LX through the high-side switch's body diode.

#### SYS Regulation Voltage

- When the DC-DC is enabled as a buck and the charger is disabled,  $Q_{BATT}$  is off and  $V_{SYS}$  is regulated to  $V_{SYSREG\_TRK\_MIN}$  when the  $V_{BATT} < V_{SYSMIN}$  or  $V_{SYSREG\_TRK}$  when the  $V_{BATT} \geq V_{SYSMIN}$ .
- When the DC-DC is enabled as a buck and the charger is enabled but in a non-charging state such as done, thermistor suspend, watchdog suspend, or timer fault,  $Q_{BATT}$  is off and  $V_{SYS}$  is regulated to  $V_{SYSREG\_TRK\_MIN}$  when the  $V_{BATT} < V_{SYSMIN}$  or  $V_{SYSREG\_TRK}$  when the  $V_{BATT} \geq V_{SYSMIN}$ .
- When the DC-DC is enabled as a buck and charging in prequalification, fast-charge, or top-off modes,  $V_{SYS}$  is regulated to  $V_{SYSMIN}$  when the  $V_{BATT} < V_{SYSMIN}$ ; in this mode, the  $Q_{BATT}$  switch acts as a linear regulator and dissipates power  $[P = (V_{SYSMIN} - V_{BATT}) \times I_{BATT}]$ . When  $V_{BATT} > V_{SYSMIN}$ , then  $V_{SYS} = V_{BATT} + I_{BATT} \times R_{BAT2SYS}$ ; in this mode, the  $Q_{BATT}$  switch is closed.
- In all of the above modes, if the combined SYS and BYP loading exceeds the input current limit, then  $V_{SYS}$  drops to  $V_{BATT} - V_{BSREG}$  and the battery provides supplemental current.
- When the DC-DC is enabled as a boost, then the  $Q_{BATT}$  switch is closed, and  $V_{SYS} = V_{BATT} - I_{BATT} \times R_{BAT2SYS}$ .

#### Input Validation

The charger input is compared with several voltage thresholds to determine if it is valid. A charger input must meet the following four characteristics to be valid:

- CHGIN must be above  $V_{CHGIN\_UVLO}$  to be valid. Once CHGIN is above the UVLO threshold, the information (together with IN2SYS, described below) is latched and can only be reset when the charger is in an adaptive input current loop (AICL) and the input current is lower than the  $I_{UL0\_DET}$  threshold.
- CHGIN must be below its overvoltage-lockout threshold ( $V_{CHGIN\_OVLO}$ ).
- CHGIN must be above the system voltage by IN2SYS drop out.
- CHGIN input generates a CHGIN\_I interrupt when its status changes. The input status can be read with CHGIN\_OK and CHGIN\_DTLS. Interrupts can be masked with CHGIN\_M.

#### Input Current Limit

The default settings of the CHGIN\_ILIM and MODE control bits are such that when a charge source is applied to CHGIN, the IC turns its DC-DC converter on in BUCK mode, limits  $V_{SYS}$  to  $V_{SYSREG\_TRK}$ , and limits the charge source current to  $I_{INLIMIT}$ . All control bits are reset on global shutdown.

### Input Voltage Regulation Loop

An input voltage regulation loop allows the charger to be well behaved when it is attached to a poor-quality charge source. The loop improves performance with relatively high resistance charge sources that exist when long cables are used or devices are charged with non-compliant USB hub configurations. Additionally, this input voltage regulation loop improves performance with current limited adapters. If the ICs input current limit is programmed above the current-limit threshold of a given adapter, the input voltage loop allows the IC to regulate at the current limit of the adapter. Finally, the input-voltage regulation loop allows the IC to perform well with adapters that have poor transient load response times.

The input voltage regulation loop automatically reduces the inductor average current to keep the input voltage at  $V_{CHGIN\_REG}$ . If the input current is reduced to  $I_{ULO\_DET}$  and the input voltage is below  $V_{CHGIN\_REG}$ , then the charger input is turned off.  $V_{CHGIN\_REG}$  is programmable with  $V_{CHGIN\_REG}[1:0]$ .

After operating with the input voltage regulation loop active, an  $AICL\_I$  interrupt is generated, and  $AICL\_OK$  sets to 0. To optimize input power when working with a current limited charge source, monitor the  $AICL\_OK$  status while decreasing the input current limit. When the input current limit is set below the limit of the adapter, the input voltage rises. Although the input current limit is lowered, more power can be extracted from the input source when the input voltage is allowed to rise.

#### Example 1. Optimum use of the Input Voltage Regulation Loop Along with a Current Limited Adapter.

Sequence of Events:

1.  $V_{BATT} = 3.2V$ , the system is operating normally.
2.  $MODE = 0x04$ ,  $CHGIN\_ILIM = 100mA$ ,  $CHG\_CV\_PRM = 4.2V$ ,  $V_{CHGIN\_REG} = 4.5V$ ,  $CHG\_CC\_TOT = 2.0A$ .
3. A 5.0V 1.2A current limited dedicated USB charger is applied to  $CHGIN$ .
4. The DC-DC buck regulator turns on,  $V_{SYS}$  is regulated to  $V_{BATTREG}$  (4.2V) and the input is allowed to provide 100mA to the system.
5. The system detects that the charge source is a dedicated USB charger and enables the battery charger ( $MODE = 0x05$ ) and programs an input current limit to 1.8A ( $CHGIN\_ILIM = 1.8A$ ).
6. The input current limit starts to ramp up from 100mA to 1.8A, but at the input current limit of the adapter (1.2A), the adapter voltage collapses. The ICs input voltage regulation loop prevents the adapter voltage from falling below 4.5V ( $V_{CHGIN\_REG} = 4.5V$ ). An  $AICL\_I$  interrupt is generated and  $AICL\_OK$  sets to 0.
7. With the input-voltage regulation loop active, the adapter provides 1.2A at 4.5V which is a total of 5.4W being delivered to the system.
8. The system software detects that the input voltage regulation loop is active and it begins to ramp down the programmed input current limit. When the current limit ramps down to 1.175A, the adapter is no longer in the current limit, and the adapter voltage increases from 4.5V to 5.0V.
9. With the adapter operating just below its current limit, it provides 1.175A at 5.0V which is a total of 5.88W to the system. This is 440mW more than when the adapter was in the current limit.

### System Self-Discharge with No Power

To ensure a timely, complete, repeatable, and reliable reset behavior when the system has no power, the ICs actively discharge the  $SYS$  nodes when  $Q_{BATT}$  and switcher are disabled and  $V_{SYS}$  is less than  $V_{SYSUVLO}$ . As shown in [Figure 1](#), the  $SYS$  discharge resistor is  $600\Omega$ .

#### Example 1. Basic System Self-Discharge

Initial Conditions: No charger adapter is present at  $CHGIN$ , the BAT-to-SYS switch is closed,  $C_{BAT} = 100\mu F$ ,  $C_{SYS} = 200\mu F$ ,  $V_{BATT} = 3.6V$ , and  $V_{SYSUVLO}$  falling is  $SYS\_UVLOB\_F$ .

Sequence of Events:

1. With the system in its normal operating mode it is drawing 1A.
2. The main battery is removed.
3. The system continues to draw 1A until  $V_{SYS}$  falls below  $V_{SYSUVLO}$ . This takes  $480\mu s$  ( $(3.6V-2.0V)/1A \times 300\mu F$ ).
4. When the system voltage falls below  $V_{SYSUVLO}$ , the system turns off the leakage current. To facilitate discharging  $C_{BAT}$  and  $C_{SYS}$  the IC engages its  $600\Omega$  discharge resistors.

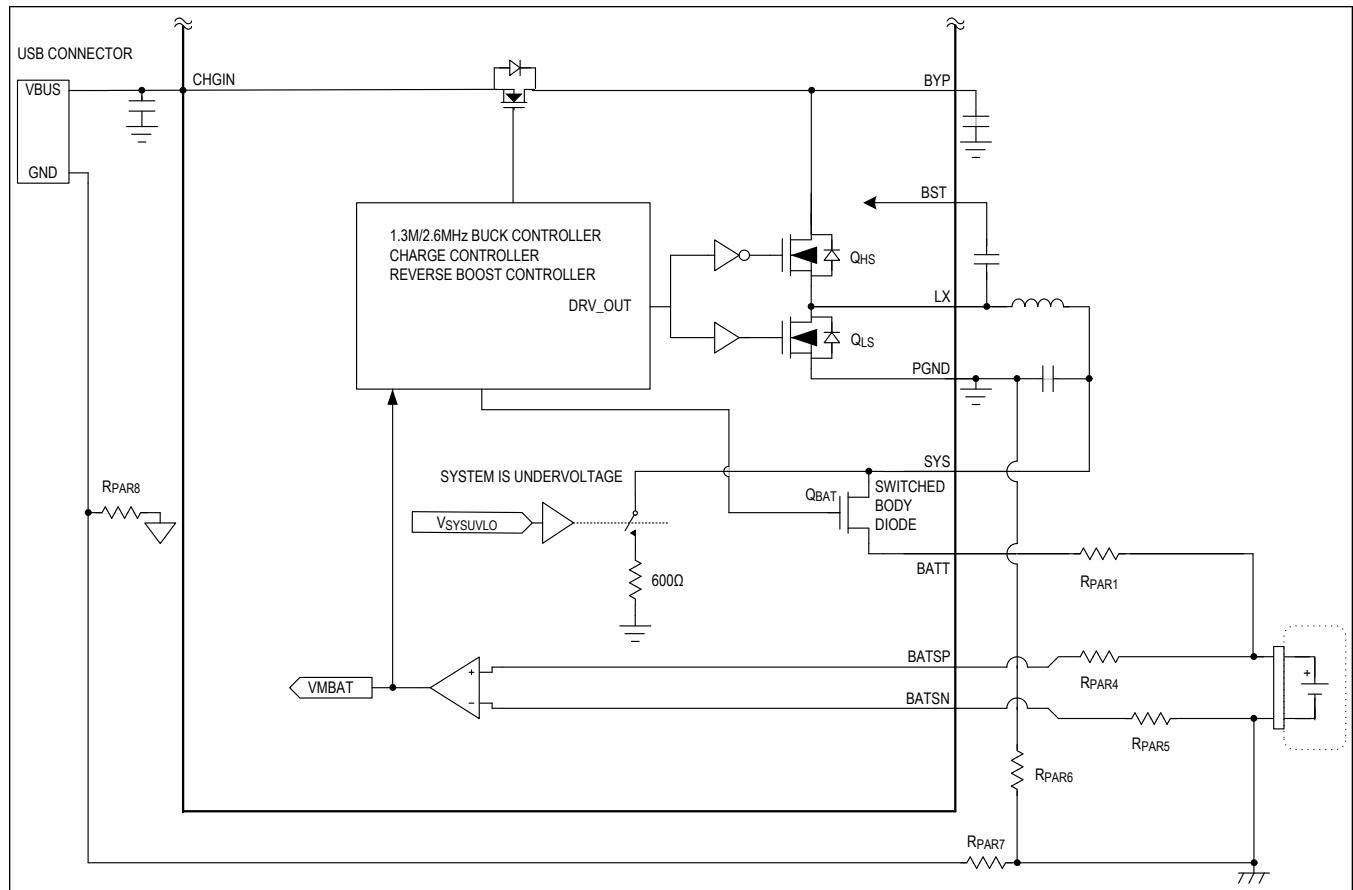


Figure 1. System Self-Discharge Circuit

**Power States**

The MAX77985/MAX77986 transitions between power states as input/battery and load conditions dictate; see [Figure 2](#).

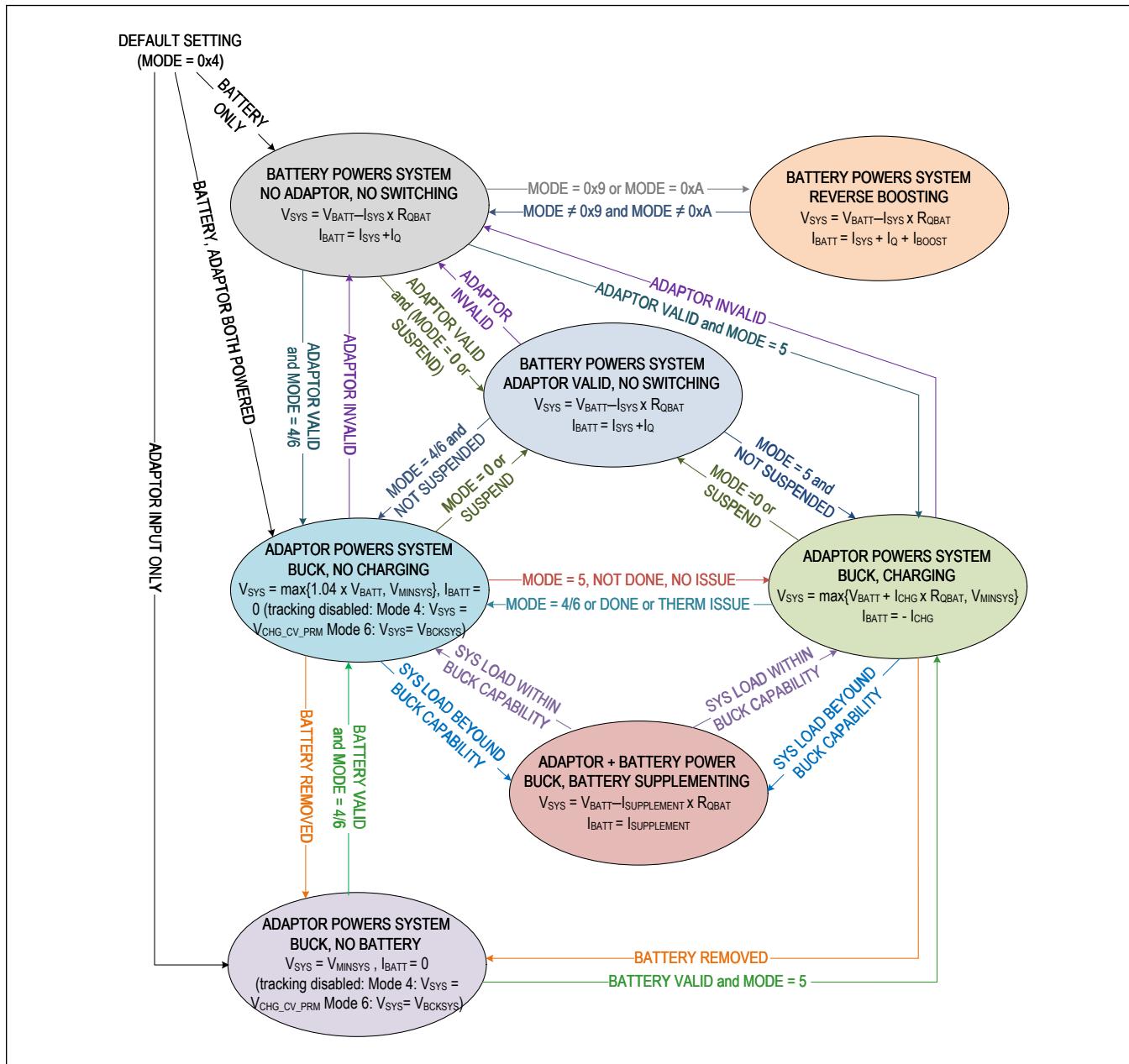


Figure 2. Power State Diagram

The IC provides five (5) power modes and one (1) no power mode (MODE detailed description is at register CHG\_CNFG\_00 [3:0]). Under power limited conditions, the PowerPath™ feature maintains SYS load at the expense of battery charge current. Also, the battery supplements the input power when required. As shown, transitions between power states are initiated by the detection/removal of valid power sources, OTG events, and under-voltage conditions. Details of the SYS voltage and BATT current are provided for each state. There are six main usage modes:

1. NO INPUT POWER, MODE = undefined: No input adapter or battery is detected. The charger and system are off. The battery is disconnected and the charger is off.
2. BATTERY-ONLY, MODE = any modes: The adapter is invalid and outside the input voltage operating range ( $Q_{CHGIN} = OFF$ ). The battery is connected to power the SYS load ( $Q_{BATT} = ON$ ).
3. NO CHARGE-BUCK, MODE = 0x04: The adapter is valid, buck supplies power to SYS. The battery is disconnected ( $Q_{BATT} = OFF$ ) when SYS load is less than the power that the buck can supply.  
When the SYS load is larger than the power that the buck can supply, the battery is reconnected ( $Q_{BATT} = ON$ ) and supplements extra SYS load.
4. CHARGE-BUCK, MODE = 0x05: The adapter is valid, the buck supplies power to SYS, and charges the battery with  $I_{BATT}$ .
5. HIGH-VOLTAGE-BUCK, MODE = 0x06: Power path same as MODE = 0x04, except for the SYS regulation voltage when SYS tracking disabled is controlled by BCKSYS instead of CHG\_CV\_PRM.
6. BATTERY-BOOST (FLASH), MODE = 0x09: OTG is inactive ( $Q_{CHGIN} = OFF$ ). The battery is connected to support SYS and BYP loads ( $Q_{BATT} = ON$ ), and the charger is operating in boost mode (Boost = ON).
7. BATTERY-BOOST (OTG), MODE = 0x0A: OTG is active ( $Q_{CHGIN} = ON$ ). The battery is connected to support SYS and OTG loads ( $Q_{BATT} = ON$ ), and the charger is operating in boost mode (Boost = ON).

### Charger States

The ICs utilize several charging states to safely and quickly charge batteries as shown in [Figure 3](#). The figure shows an exaggerated view of a Li+/Li-Poly battery progressing through the following charge states when there is no system load and the die and battery are close to room temperature. It shows a complete charging state transition process with four states: prequalification, fast-charge, top-off, and done.

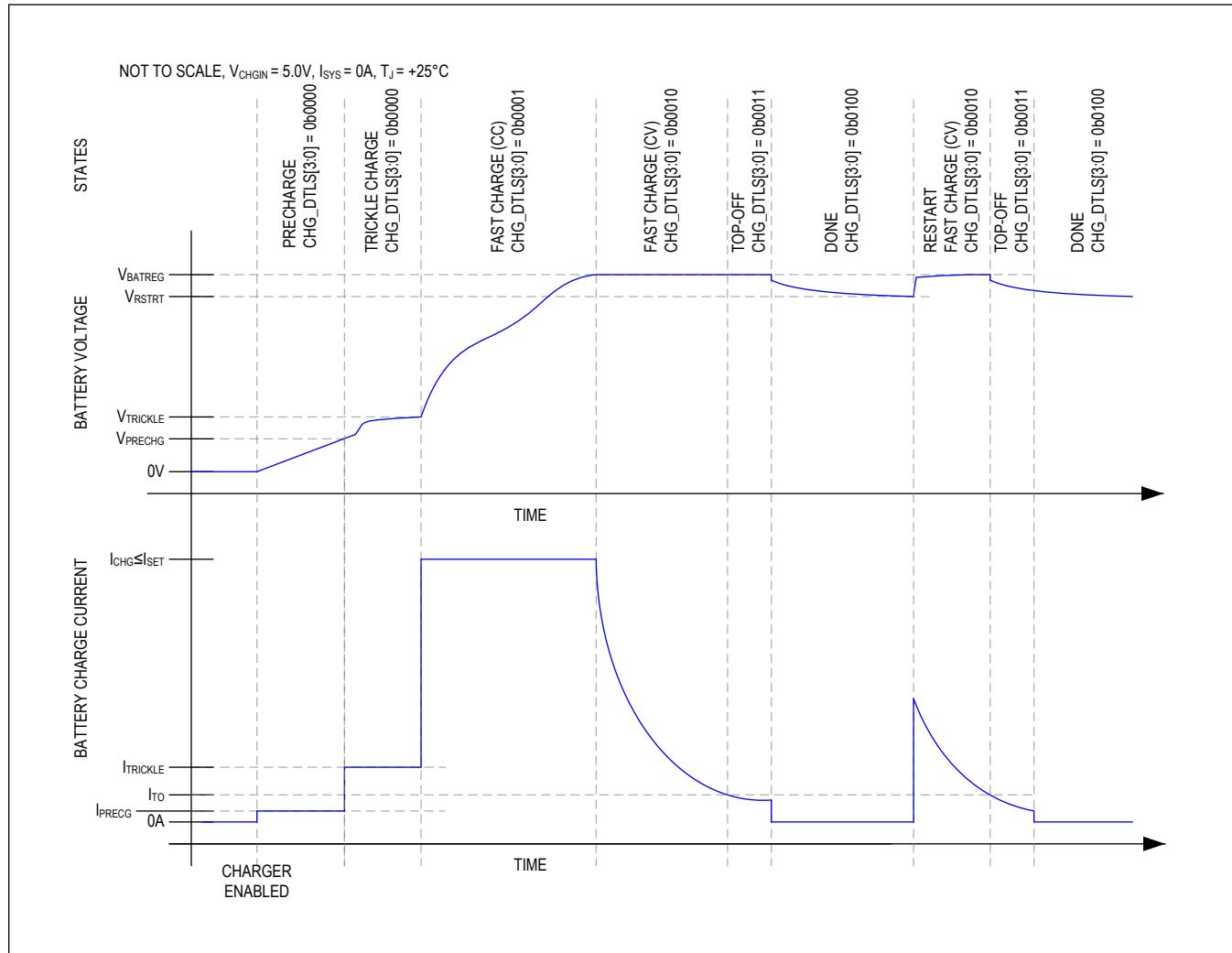


Figure 3. Li+/Li-Poly Charge Profile

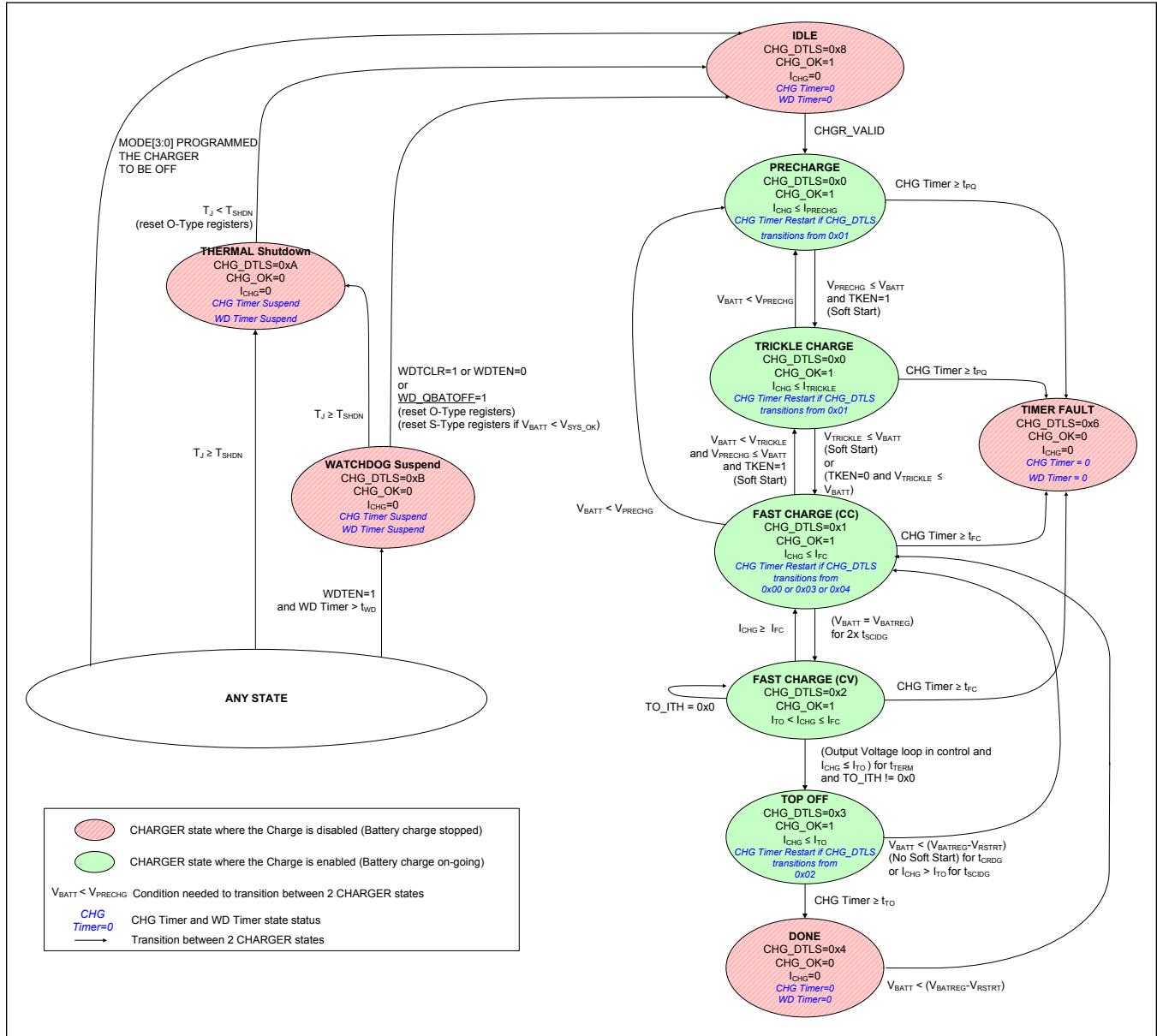


Figure 4. Charger State Diagram

### No Input Power or Charge Idle State

While in the “no input power or charger idle” state, the charge current is 0mA, the watchdog and charge timers are forced to 0, and the power to the system is provided by either the battery or the adapter. When both battery and adapter power is available, the adapter provides primary power to the system and the battery contributes supplemental energy to the system if necessary.

To exit the “no input power or charger idle” state, the charger input must be valid and the charger has to be enabled.

### Precharge State

As shown in [Figure 3](#), the precharge state occurs when the main-battery voltage is less than  $V_{PRECHG}$ . After being in this state for  $t_{SCIDG}$ , a CHG\_I interrupt is generated only if CHG\_OK was 0 previously, CHG\_OK is set to 1, and CHG\_DTLS is set to 0x00. In the precharge state, the charge current in the battery is  $I_{PRECHG}$ .

The following events cause the state machine to exit this state:

- Main battery voltage rises above  $V_{PRECHG}$  and the charger enters the next state in the charging cycle: “Trickle Charge”.
- If the battery charger remains in this state for longer than  $t_{PQ}$ , the charger state machine transitions to the “Timer Fault” state.
- If the watchdog timer is not serviced (see the [Watchdog Timer](#) section), the charger state machine transitions to the “Watchdog Suspend” state.

Note that the precharge state works with battery voltages down to 0V. The low 0V operation typically allows this battery charger to recover batteries that have an “open” internal pack protector. Typically a pack internal protection circuit opens if the battery has seen an overcurrent, undervoltage, or overvoltage. When a battery with an “open” internal pack protector is used with this charger, the precharge mode current flows into the 0V battery—this current raises the pack’s terminal voltage to the point where the internal pack protection switch closes.

Note that a normal battery typically stays in the precharge state for several minutes or less. Therefore a battery that stays in the precharge for longer than  $t_{PQ}$  may be experiencing a problem.

### Trickle Charge State

As shown in [Figure 3](#), the trickle charge state occurs when  $V_{BATT} > V_{PRECHG}$  and  $V_{BATT} < V_{TRICKLE}$ . After being in this state for  $t_{SCIDG}$ , a CHG\_I interrupt is generated only if CHG\_OK was 0 previously, CHG\_OK is set to 1, and CHG\_DTLS = 0x00.

With  $TKEN = 1$  and the IC is in its trickle charge state, the current in the battery is less than or equal to  $I_{TRICKLE}$ . When  $TKEN = 0$ , the battery current is less than or equal to  $I_{FC}$ .

Charge current may be less than  $I_{TRICKLE}/I_{FC}$  for any of the following reasons:

- The charger input is in input current limit
- The charger input voltage is low
- The charger is in thermal foldback
- The system load is consuming adapter current. Note that the system load always gets priority over the battery charge current.

Typical systems operate with  $TKEN = 1$ . When operating with  $TKEN = 0$ , the system’s software usually sets  $I_{FC}$  to a low value such as 450mA and then monitors the battery voltage. When the battery exceeds a relatively low voltage such as 3.1V, then the system’s software usually increases  $I_{FC}$ .

The following events cause the state machine to exit this state:

- When the main battery voltage rises above  $V_{TRICKLE}$  or the  $TKEN$  bit is cleared, the charger enters the next state in the charging cycle: “Fast Charge (CC)”.
- If the battery charger remains in this state for longer than  $t_{PQ}$ , the charger state machine transitions to the “Timer Fault” state.
- If the watchdog timer is not serviced, the charger state machine transitions to the “Watchdog Suspend” state.

Note that a normal battery typically stays in the trickle charge state for several minutes or less. Therefore a battery that stays in trickle charge for longer than  $t_{PQ}$  may be experiencing a problem.

### Fast-Charge Constant Current (CC) State

As shown in [Figure 3](#), the fast-charge CC state occurs when the main-battery voltage is greater than the low-battery prequalification threshold and less than the battery regulation threshold ( $V_{TRICKLE} < V_{BATT} < V_{BATTREG}$ ). After being in the fast-charge CC state for  $t_{SCIDG}$ , a CHG\_I interrupt is generated only if CHG\_OK was 0 previously, CHG\_OK is set to 1, and CHG\_DTLS = 0x01.

In the fast-charge CC state, the current in the battery is less than or equal to  $I_{FC}$ . Charge current may be less than  $I_{FC}$  for any of the following reasons:

- The charger input is in input current limit
- The charger input voltage is low
- The charger is in thermal foldback
- The system load is consuming adapter current. Note that the system load always gets priority over the battery charge current.

The following events cause the state machine to exit this state:

- When the main battery voltage rises above  $V_{BATTREG}$ , the charger enters the next state in the charging cycle: “Fast Charge (CV)”.
- If the battery charger remains in this state for longer than  $t_{FC}$ , the charger state machine transitions to the “Timer Fault” state.
- If the watchdog timer is not serviced, the charger state machine transitions to the “Watchdog Suspend” state.

The battery charger dissipates the most power in the fast-charge constant current state. This power dissipation causes the internal die temperature to rise. If the die temperature exceeds  $T_{REG}$ ,  $I_{FC}$  is reduced. See the [Thermal Foldback](#) section for more information.

### Fast-Charge Constant Voltage (CV) State

As shown in [Figure 3](#), the fast-charge CV state occurs when the battery voltage rises to  $V_{BATTREG}$  from the fast-charge CC state. After being in the fast-charge CV state for  $t_{SCIDG}$ , a CHG\_I interrupt is generated only if CHG\_OK was 0 previously, CHG\_OK is set to 1, and CHG\_DTLS = 0x02.

In the fast-charge CV state, the battery charger maintains  $V_{BATTREG}$  across the battery and the charge current is less than or equal to  $I_{FC}$ . As shown in [Figure 3](#), the charger current decreases exponentially in this state as the battery becomes fully charged.

The smart power selector control circuitry may reduce the charge current lower than the battery may otherwise consume for any of the following reasons:

- The charger input is in input current limit
- The charger input voltage is low
- The charger is in thermal foldback
- The system load is consuming adapter current. Note that the system load always gets priority over the battery charge current.

The following events causes the state machine to exit this state:

- When the charger current is below  $I_{TO}$  for  $t_{TERM}$ , the charger enters the next state in the charging cycle: the “TOP OFF” state.
- If the battery charger remains in this state for longer than  $t_{FC}$ , the charger state machine transitions to the “Timer Fault” state.
- If the watchdog timer is not serviced, the charger state machine transitions to the “Watchdog Suspend” state.

### Top-Off State

As shown in [Figure 3](#), the top-off state can only be entered from the fast-charge CV state when the charger current decreases below  $I_{TO}$  for  $t_{TERM}$ . After being in the top-off state for  $t_{SCIDG}$ , a CHG\_I interrupt is generated only if CHG\_OK was 0 previously, CHG\_OK is set to 1, and CHG\_DTLS = 0x03. In the top-off state, the battery charger tries to maintain  $V_{BATTREG}$  across the battery and typically the charge current is less than or equal to  $I_{TO}$ .

The smart power selector control circuitry may reduce the charge current lower than the battery may otherwise consume

for any of the following reasons:

- The charger input is in the input current limit
- The charger input voltage is low
- The charger is in thermal foldback
- The system load is consuming adapter current. Note that the system load always gets priority over the battery charge current.

The following events cause the state machine to exit this state:

- After being in this state for the top-off time ( $t_{TO}$ ), the charger enters the next state in the charging cycle: the “DONE” state.
- If  $V_{BATT} < V_{BATTREG} - V_{RSTART}$ , the charger goes back to the “FAST CHARGE (CC)” state.
- If the watchdog timer is not serviced, the charger state machine transitions to the “Watchdog Suspend” state.

### Top-Off Current Shift in Skip Mode

When  $DISKIP = 0$ , the buck charger operates in skip mode during light loads. The skip mode operation makes a long-term ripple in the battery charging current and usually results in the  $I_{TO}$  shifting down than the actual target. Approximately, if  $(I_{SYS} + I_{TO}) < I_{ULO\_DET} \times V_{CHGIN}/V_{BATTREG}$  (let's call this "condition T") then the IC may exhibit lower  $I_{TO}$  accuracy..

To maintain  $I_{TO}$  accuracy, one of the following two methods may be employed:

1. Use higher  $I_{TO}$  and longer  $t_{TO}$ : For example,  $V_{CHGIN} = 15V$ ,  $V_{BATTREG} = 4.2$ ,  $I_{SYS} = 0A$ ,  $I_{TO} = 200mA$ . It may show  $I_{TO}$  shifted down to around 100mA because it meets condition T. In the case where the selected  $I_{TO}$  is 250mA or 300mA, then  $I_{TO}$  accuracy is maintained since the buck charger is still operating in non-skip mode.
2. Set  $DISKIP = 1$  only when  $V_{CHGIN} \geq 15V \ \&\& \ CHG\_DTLS = 0x02$  (CV mode), and set  $DISKIP = 0$  when  $CHG\_DTLS = 0x03$  (Top-Off mode). The alternative condition could be  $V_{CHGIN} \geq 15V \ \&\& \ I_{BAT} = 50mA + I_{TO}$  for setting  $DISKIP = 1$  assuming the system processor can read the battery current through the fuel gauge.

### Done State

As shown in [Figure 3](#), the battery charger enters its done state after the charger has been in the top-off state for  $t_{TO}$ . After being in this state for  $t_{SCIDG}$ , a CHG\_I interrupt is generated only if CHG\_OK was 0 previously, CHG\_OK is set to 0, and CHG\_DTLS = 0x04.

The following events cause the state machine to exit this state:

- If  $V_{BATT} < V_{BATTREG} - V_{RSTART}$ , the charger goes back to the “FAST-CHARGE CC” state.
- If the watchdog timer is not serviced, the charger state machine transitions to the “Watchdog Suspend” state.

In the done state, the charge current into the battery ( $I_{CHG}$ ) is 0A. In the done state, the charger presents a very low load ( $I_{MBDN}$ ) to the battery. If the system load presented to the battery is low, then a typical system can remain in the done state for many days. If left in the done state long enough, the battery voltage decays below the restart threshold ( $V_{RSTART}$ ), and the charger state machine transitions back into the fast-charge CV state. There is no soft-start (di/dt limiting) during the done to a fast-charge state transition.

### Timer Fault State

The battery charger provides both a charge timer and a watchdog timer to ensure safe charging. The charge timer prevents the battery from charging indefinitely. The time that the charger is allowed to remain in each of its prequalification states is  $t_{PQ}$ . The time that the charger is allowed to remain in the fast-charge CC & CV states is  $t_{FC}$  which is programmable with FCHGTIME. Finally, the time that the charger is in the top-off state is  $t_{TO}$  which is programmable with TO\_TIME. Upon entering the timer fault state a CHG\_I interrupt is generated without a delay, CHG\_OK is cleared, and CHG\_DTLS = 0x06.

In the timer fault state, the charger is off. The charger can exit the timer fault state by programming the charger to be off and then programming it to be on again through the MODE bits. Alternatively, the charger input can be removed and re-inserted to exit the timer fault state.

### Watchdog Timer

The battery charger provides both a charge timer and a watchdog timer to ensure safe charging. The watchdog timer protects the battery from charging indefinitely if the host hangs or otherwise cannot communicate correctly. The watchdog timer is disabled by default with WDTEN = 0. To use the watchdog timer feature enable the feature by setting WDTEN. While enabled, the system controller must reset the watchdog timer within the timer period ( $t_{WD}$ ) for the charger to operate normally. Reset the watchdog timer by programming WDTCLR = 0x01.

If WD\_QBATTOFF bit is set to 0 and the watchdog timer expires while the charger is in dead-battery prequalification, low-battery prequalification, fast-charge CC or CV, top-off, done, or timer fault, the charging stops, a CHG\_I interrupt is generated only if CHG\_OK was 1 previously, CHG\_OK is cleared, and CHG\_DTLS indicates that the charger is off because the watchdog timer expired. Once the watchdog timer has expired, the charger may be restarted by programming WDTCLR = 0x01. The SYS node can be supported by the battery and/or the adapter through the DC-DC buck while the watchdog timer is expired.

If the WD\_QBATTOFF bit is set to 1 and the watchdog timer expires, MAX77986 turns off the buck, charger, and Q\_BATT switch for 150ms. And then VSYS voltage collapses and it resets all I<sup>2</sup>C registers. The IC restarts as the initial power-up condition.

### Thermal Shutdown State

The thermal shutdown state occurs when the battery charger is in any state and the junction temperature ( $T_J$ ) exceeds the device's thermal-shutdown threshold ( $T_{SHDN}$ ). When  $T_J$  is close to  $T_{SHDN}$  the charger folds back the charge current to 0A (see the [Thermal Foldback](#) section). Upon entering this state, CHG\_I interrupt is generated if CHG\_OK was 1 previously, CHG\_OK is cleared, and CHG\_DTLS = 0x0A.

In the thermal shutdown state, the charger is off. MODE register (CHG\_CNF0[3:0]) is reset to its default value as well as all O-type registers.

### Charger Interrupt Debounce Time

**Table 2. Charger Interrupt Debounce Time**

INTERRUPT	DEBOUNCE TIME RISING	DEBOUNCE TIME FALLING
	Typ (ms)	Typ (ms)
AICL_I	30	30
CHGIN_I	7.5	—
INLIM_I	30	30
BAT_I (Overvoltage $T_{BATOV}$ )	7.5	—
BYP_I ( $T_{OTG\_I}$ )	20	—
BYP_I ( $BST\_I_{LIM}$ )	30	—
BYP_I (Buck Neg $I_{LIM}$ )	0.5	—

Accuracy of the timer is defined by  $T_{ACC}$ .

### Main-Battery Differential Voltage Sense

BATSP and BATSN are differential remote sense lines for the main battery. To improve accuracy and decrease charging times, the battery charger voltage sense is based on the differential voltage between BATSP and BATSN. Similarly, the thermistor voltage is interpreted with respect to BATSN.

A Maxim battery charger without the remote sensing function would typically measure the battery voltage between BATT and GND. In case a charge current of 1A measuring from BATT to GND leads to a  $V_{BATT}$  that is 40mV higher than the real voltage because of RPAR1 and RPAR7 ( $|I_{CHG} \times (R_{PAR1} + R_{PAR7})| = 1A \times 40m\Omega = 40mV$ ). Since the charger thinks the battery voltage is higher than it actually is, it enters its fast-charge CV state sooner and the effective charge time may be extended by 10 minutes (based on real lab measurements). This charger with differential remote sensing does not experience this type of problem because BATSP and BATSN sense the battery voltage directly. To get the maximum benefit from these sense lines, connect them as close as possible to the main-battery connector.

### Reverse Boost Mode

The DC-DC converter topology of the IC allows it to operate as a forward buck converter or as a reverse boost converter. The modes of the DC-DC converter are controlled with MODE. When MODE = 0x09 or 0x0A, the DC-DC converter operates in reverse boost mode allowing it to source current to BYP. To allow current flow to CHGIN, set MODE = 0x0A. This mode allows current to be sourced from CHGIN and is commonly referred to as OTG mode.

When MODE = 0x0A, the DC-DC converter operates in reverse boost mode and regulates V<sub>BYP</sub> to V<sub>BYP\_OTG</sub> and the low ohmic (R<sub>CHGIN2BYP</sub>) switch from BYP to CHGIN is closed. The current through the BYP to CHGIN switch is limited to the value programmed by OTG\_ILIM. The programmable OTG\_ILIM options allow for supplying from 500mA to 3100mA to an external load. When the OTG mode is selected, the unipolar CHGIN transfer function measures the current going out of CHGIN. When OTG mode is not selected, the unipolar CHGIN transfer function measures the current going into CHGIN.

If the external OTG load at CHGIN exceeds ICHGIN.OTG.ILIM current during a minimum time of T<sub>OTG\_I</sub> ms, then a BYP\_I interrupt is generated. BYP\_OK = 0 and BYP\_DTLS[0] = 1. In response to an overload at CHGIN during OTG mode operation, the BYP to CHGIN switch is latched off T<sub>OTG</sub> fault after entering the OTG\_ILIM condition. If the overload at CHGIN persists, BYP\_DTLS keeps continuing to report OTG\_ILIM fault through BYP\_DTLS[0] = 1.

If OTG\_REC\_EN bit = '1': other functions remain unaffected, i.e., BYP is supplied by reverse boost, and the BYP to CHGIN switch automatically retries after T<sub>OTG\_retry</sub>. If the overload at CHGIN persists, then the CHGIN switch toggles ON and OFF with T<sub>OTG\_fault</sub> ON time and T<sub>OTG\_retry</sub> OFF time.

If OTG\_REC\_EN bit = '0': the BYP to CHGIN switch remains off and the switcher is turned off until MODE is toggled.

BYP\_I exit interrupt is only generated on OTG load release such as IOTG < ICHGIN.OTG.ILIM or FET opening. At that time, the BYP\_I interrupt is generated. BYP\_OK = 1 and BYP\_DTLS[0] = 0.

**Note:** On OTG\_ILIM debounce time out, BYP\_DTLS[0] is latched until the BYP\_DTLS register is read by AP. BYP\_OK is matching BYP\_DTLS[0] behavior.

### Main-Battery Overcurrent Protection During System Power-Up

The “main-battery overcurrent protection during system power-up” feature limits the main battery to system current to I<sub>SYS</sub> as long as V<sub>SYS</sub> is less than V<sub>SYS</sub><sub>BAT</sub>. This feature limits the surge current that typically flows from the main battery to the device’s low-impedance system-bypass capacitors during a system power-up. System power-up is anytime that energy from the battery is supplied to SYS when V<sub>SYS</sub> < V<sub>SYS</sub><sub>BAT</sub>. This “system power-up” condition typically occurs when a battery is hot-inserted into an otherwise unpowered device. Similarly, the “system power-up” condition can occur when the DISIBS bit is driven low.

When “system power-up” occurs due to hot-insertion into an otherwise unpowered device, a small delay is required for this feature’s control circuits to activate. A current spike over I<sub>SYS</sub><sub>BAT</sub> can occur during this time.

### Main-Battery Overcurrent Protection Due to Fault

The IC protects itself, the battery, and the system from potential damage due to excessive battery discharge current. Excessive battery discharge current can occur in a smartphone for several reasons such as exposure to moisture, a software problem, an IC failure, a component failure, or a mechanical failure that causes a short circuit. The main-battery overcurrent protection feature is enabled with B2SOVRC; disabling this feature reduces the main-battery current consumption by 2 $\mu$ A.

When the main battery (BATT) to system (SYS) discharge current (I<sub>BATT</sub>) exceeds the programmed overcurrent threshold (I<sub>BOVRC</sub>) for at least t<sub>BOVRC</sub>, a BAT\_I interrupt is generated, BAT\_OK is cleared, and BAT\_DTLS reports an overcurrent condition. Typically when the system’s processor detects this overcurrent interrupt it executes a housekeeping routine that tries to mitigate the overcurrent situation. If the processor cannot correct the overcurrent within t<sub>OCP</sub>, then the IC disables the BATT to SYS discharge path (Q<sub>BATT</sub> switch) and turns off the Buck.

Under OCP fault condition, when SYS is low (V<sub>SYS</sub> < V<sub>SYS</sub><sub>UP</sub>) for t<sub>OCP\_RETRY</sub>, the IC restarts on its own and attempts to pullup SYS again. If the fault condition remains, the whole cycle repeats until this fault condition is removed.

AP can also turn off the Q<sub>BATT</sub> switch by driving the DISIBS bit to a logic-high or pulling the DISQBAT pin high.

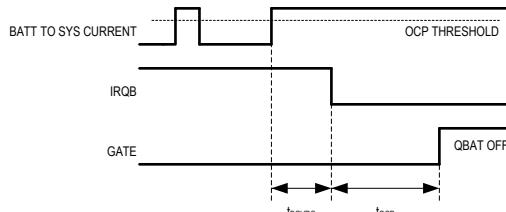


Figure 5. BATT to SYS Overcurrent Protection

There are different scenarios of how the ICs respond to the OCP event depending on the available power source and the state of the charger:

1) The IC is only powered by a battery, then the OCP event occurs:

- a. QBATT switch opens.
- b. SYS collapses and is allowed to go to 0V.
- c. If RECYCLE\_EN = 1: After SYS is low ( $V_{SYS} < V_{SYSUP}$ ) for  $t_{OCP\_RETRY}$ , the IC retries to bring up  $V_{SYS}$  above  $V_{SYSUP}$ . If the fault condition is not removed by SYS collapsing to 0V, the IC may fail to bring up  $V_{SYS}$  above  $V_{SYSUP}$  (see the [V<sub>SYS</sub> Power-Up Failure \(P<sub>WRUPFAIL</sub>\)](#) section). If a  $V_{SYS}$  power-up failure happened during a retry, then a valid charger input has to be inserted to turn on the buck and QBATT switch. If RECYCLE\_EN = 0: The QBATT switch remains open. When a valid charger input is inserted, the buck and QBATT switch turns on.

2) The IC is powered from BATT and CHGIN, buck is switching, charge is ON, then an OCP event occurs:

- a. Buck is off and the QBATT switch opens.
- b. SYS collapses and is allowed to go to 0V.
- c. Regardless of the RECYCLE\_EN setting, the IC retries to bring up  $V_{SYS}$  above  $V_{SYSUP}$ . If the fault condition is not removed by SYS collapsing to 0V, the IC may fail to bring up  $V_{SYS}$  above  $V_{SYSUP}$  (see the [V<sub>SYS</sub> Power-Up Failure \(P<sub>WRUPFAIL</sub>\)](#) section). If a  $V_{SYS}$  power-up failure happened during a retry, then a valid charger input has to be inserted to turn on the buck and QBATT switch.

3) The IC is powered from CHGIN, buck is switching, charge is OFF, then an OCP event occurs:

- a. Buck is off and the QBATT switch opens.
- b. SYS collapses and is allowed to go to 0V.
- c. Regardless of the RECYCLE\_EN setting, the IC retries to bring up  $V_{SYS}$  above  $V_{SYSUP}$ . If the fault condition is not removed by SYS collapsing to 0V, the IC may fail to bring up  $V_{SYS}$  above  $V_{SYSUP}$  (see the [V<sub>SYS</sub> Power-Up Failure \(P<sub>WRUPFAIL</sub>\)](#) section). If a  $V_{SYS}$  power-up failure happened during a retry, then a valid charger input has to be inserted to turn on the buck and QBATT switch.

**Note:** For parts with chip revision 0b001 (PASS1), contact Analog Devices for Rev 0 of the data sheet. This information can be read from the CHIP\_REVISION (0x01) register.

#### Battery to SYS QBATT Switch Control (DISIBS)

To protect the system from unexpected and critical events (e.g., excessive battery discharge current), the AP can control the MAX77985/MAX77986 QBATT switch by driving the DISIBS bit to a logic-high.

There are different scenarios of how the IC responds to setting the DISIBS bit high depending on the available power source and the state of the charger:

1) The IC is only powered from BATT and DISIBS bit is set:

- a. QBATT switch opens

- b. SYS collapses and is allowed to go to 0V
- c. If RECYCLE\_EN = 1, the IC self-recovers and restarts after  $t_{OCP\_RETRY}$ . If the fault condition is not removed by SYS collapsing to 0V, the IC may fail to bring up  $V_{SYS}$  above  $V_{SYSUP}$  (see the [V<sub>SYS</sub> Power-Up Failure \(PWRUPFAIL\) section](#)). If RECYCLE\_EN = 0, after  $t_{OCP\_RETRY}$ , the IC does not recycle until a valid charger input is inserted.

2) The IC is powered from BATT, CHGIN is present, the charger buck is not switching, and the DISIBS bit is set:

- a.  $Q_{BATT}$  switch opens
- b. SYS collapses and is allowed to go to 0V
  - a. Regardless of RECYCLE bit setting, the IC self-recovers and restarts after  $t_{OCP\_RETRY}$ . If the fault condition is not removed by SYS collapsing to 0V, the IC may fail to bring up  $V_{SYS}$  above  $V_{SYSUP}$  (see the [V<sub>SYS</sub> Power-Up Failure \(PWRUPFAIL\) section](#)).

3) The IC is powered from CHGIN, the buck is switching, the charge is OFF, and the DISIBS bit is set:

- a.  $Q_{BATT}$  stays OFF (opened)
- b. Turn off Buck
- c. SYS collapses and is allowed to go to 0V
- d. Regardless of RECYCLE bit setting, the IC self-recovers and restarts after  $t_{OCP\_RETRY}$ . If the fault condition is not removed by SYS collapsing to 0V, the IC may fail to bring up  $V_{SYS}$  above  $V_{SYSUP}$  (see the [V<sub>SYS</sub> Power-Up Failure \(PWRUPFAIL\) section](#)).

4) The IC is powered from CHGIN, the buck is switching, the charge is ON, and the DISIBS bit is set:

- a. Charge is disabled
- b.  $Q_{BATT}$  turns off (opened)
- c. Turn off Buck
- d. SYS collapses and is allowed to go to 0V
- e. Regardless of the RECYCLE bit setting, the IC self-recovers and restarts after  $t_{OCP\_RETRY}$ . If the fault condition is not removed by SYS collapsing to 0V, the IC may fail to bring up  $V_{SYS}$  above  $V_{SYSUP}$  (see the [V<sub>SYS</sub> Power-Up Failure \(PWRUPFAIL\) section](#)).

**Note:** For parts with chip revision 0b001 (PASS1), contact Analog Devices for Rev 0 of the data sheet. This information can be read from the CHIP\_REVISION (0x01) register.

### HW Control of Battery to SYS $Q_{BATT}$ Switch—DISQBAT

To protect the system from unexpected and critical events (e.g., excessive battery discharge current), the AP can control the IC's  $Q_{BATT}$  switch by driving the DISQBAT hardware pin. This pin can also be driven during factory test modes.

On DISQBAT low-to-high assertion,  $Q_{BATT}$  FET opens and any ongoing charge is disabled but buck keeps switching (if allowed by MODE setting).

The IC supports factory-boost mode to enter in boost mode (through CHG\_CNFG\_00.MODE setting) and keep  $Q_{BATT}$  OFF even if boost mode is set.

This functionality is only enabled once functional register CHG\_CNFG\_07.QBEXT\_CTRL\_EN bit is set 1.

DISQBAT is an input control signal for  $Q_{BATT}$  FET with an external logic signal. If DISQBAT is driven by high,  $Q_{BATT}$  FET is truly disconnected. It has an internal 470k $\Omega$  pulldown resistor.

### Thermal Management

The ICs charger uses several thermal management techniques to prevent excessive battery and die temperatures.

### Thermal Foldback

Thermal foldback maximizes the battery charge current while regulating the ICs junction temperature. As shown in [Figure 6](#), when the die temperature exceeds the value programmed by REGTEMP ( $T_{REG}$ ), a thermal limiting circuit reduces the battery charger's target current by  $A_{TJREG}$ . The target charge current reduction is achieved with an analog control

loop (i.e., not a digital reduction in the input current). When the thermal foldback loop changes state, a CHG\_I interrupt is generated and the system's microprocessor may read the status of the thermal regulation loop through the  $T_{REG}$  status bit. Note that the thermal foldback loop being active is not considered to be an abnormal operation and the thermal foldback loop status does not affect the CHG\_OK bit (only information contained within CHG\_DTLS affects CHG\_OK).

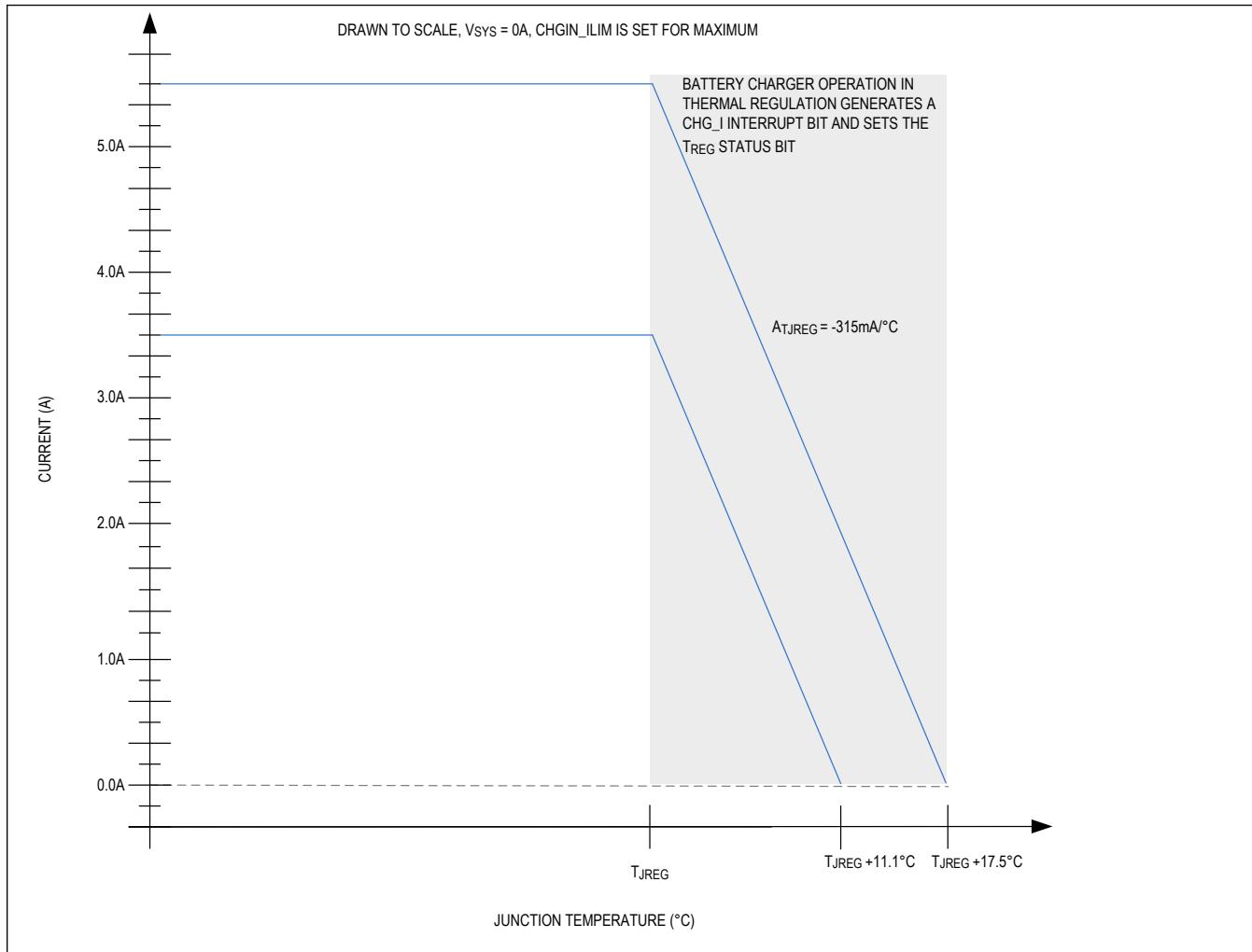


Figure 6. Charge Currents vs. Junction Temperature

### Thermistor Input (THM)

The thermistor input can be utilized to achieve functions such as charge suspension, JEITA compliant charging, and battery removal detection. The thermistor monitoring feature can be disabled by connecting the THM pin to ground.

The THM input connects to an external negative temperature coefficient (NTC) thermistor to monitor battery or system temperature.

### JEITA Compliant Charging

JEITA compliant charging is available with JEITA\_EN = 1.

Charging stops when the thermistor temperature is out of range ( $T < T_{COLD}$  or  $T > T_{HOT}$ ). The charge timers are reset and the CHG\_DTLS[3:0], CHG\_OK register bits report the charging suspension status, and CHG\_I interrupt bit is set. When the thermistor comes back into range ( $T_{COLD} < T < T_{HOT}$ ), charging resumes, and the charge timer restarts.

**Battery Removal Detection**

With pullup connected between PVDD and THM, if the battery is removed, the thermistor is disconnected from THM; this event is detected as THM is pulled up to PVDD. Battery removal event prevents charging.

**Disable Thermistor Monitoring**

Connecting THM to GND disables the thermistor monitoring function, and JEITA controlled charging is unavailable in this configuration. The IC detects an always-connected battery when THM is grounded, and charging starts automatically when a valid adapter is plugged in. In applications with removable batteries, do not connect THM to GND because the IC is not able to detect battery removal when THM is grounded. Instead, connecting THM to the thermistor pin in the battery pack is recommended.

Since the thermistor monitoring circuit employs an external bias resistor from THM to PVDD, the thermistor is not limited only to 10kΩ (at +25°C). Any resistance thermistor can be used as long as the value is equivalent to the thermistor +25°C resistance. For example, with a 10kΩ at RTB resistor, the charger enters a temperature suspend state when the thermistor resistance falls below 4.67kΩ (too hot) or rises above 30.3kΩ (too cold). This corresponds to 0°C to +45°C range when using a 10kΩ NTC thermistor with a beta of 3610. The general relation of thermistor resistance to temperature is defined by the following equation:

$$R_T = R_{25} \times e^{\left\{ \beta \left( \frac{1}{T+273^\circ C} - \frac{1}{298^\circ C} \right) \right\}}$$

where:

$R_T$  = The resistance in Ω of the thermistor at temperature T in Celsius

$R_{25}$  = The resistance in Ω of the thermistor at +25°C

$\beta$  = The material constant of the thermistor, which typically ranges from 3000k to 5000k

T = The temperature of the thermistor in °C

Some designs might prefer other thermistor temperature limits. Threshold adjustment can be accommodated by changing RTB, connecting a resistor in series and/or in parallel with the thermistor, or using a thermistor with different  $\beta$ . For example, a +45°C hot threshold and 0°C cold threshold can be realized by using a thermistor with a  $\beta$  to 4250 and connecting 120kΩ in parallel. Since the thermistor resistance near 0°C is much higher than it is near +50°C, a large parallel resistance lowers the cold threshold while only slightly lowering the hot threshold. Conversely, a small series resistance raises the hot threshold, while only slightly raising the cold threshold. Raising RTB, lowers both the hot and cold threshold, while lowering RTB raises both thresholds.

Thermistor bias current flows whenever PVDD is enabled (CHGIN valid or BOOST enabled). When using a 10kΩ thermistor and a 10kΩ pullup to THM, this results in an additional 90μA load. This load can be reduced to 9μA by instead using a 100kΩ thermistor and 100kΩ pullup resistor.

**Table 3. Trip Temperatures for Different Thermistors**

Thermistor	R25 (Ω)	10000	10000	47000	100000
	Thermistor Beta (β)	3380	3610	4050	4250
	RTB (Ω)	10000	10000	47000	100000
	R15 (Ω)	14826	15223	75342	164083
	R45 (Ω)	4900	4671	19993	40781
Trip Temperatures	T <sub>COLD</sub> (°C)	-1.3	0.2	2.7	3.7
	T <sub>COOL</sub> (°C)	9.0	10.0	11.6	12.2
	T <sub>WARM</sub> (°C)	46.2	44.8	42.5	41.7
	T <sub>HOT</sub> (°C)	62.5	59.8	55.6	54.1

**JEITA Controlled Charging**

The IC safely charges Li+ batteries in accordance with JEITA specifications. The IC monitors the battery temperature with an NTC thermistor connected at the THM pin and automatically adjusts the fast-charge current and/or charge termination voltage as the battery temperature varies. JEITA controlled charging can be disabled by setting JEITA\_EN to '0'; if JEITA\_EN = '0', thermistor input is not taken into account to determine charge state or charge current and voltage levels.

CHG\_DTLS and THM\_DTLS registers report JEITA controlled charging status.

The JEITA controlled fast-charging current (I<sub>CHGCC\_JEITA</sub>) for T<sub>WARM</sub> < T < T<sub>HOT</sub> is programmable with I<sup>2</sup>C bit CHG\_CC\_WARM.

The JEITA controlled charge termination voltage (V<sub>CHGCV\_JEITA</sub>) for T<sub>COLD</sub> < T < T<sub>COOL</sub> is programmable with I<sup>2</sup>C bit CHG\_CV\_COOL.

The JEITA controlled fast-charging current for T<sub>COLD</sub> < T < T<sub>COOL</sub> is halved (to CHG\_CC x 0.5) and the charge termination voltage for T<sub>WARM</sub> < T < T<sub>HOT</sub> is reduced (CHG\_CV\_PRM - 187.5mV for MAX77985A/MAX77986A, CHG\_CV\_PRM - 375mV for MAX77985B/MAX77986B) as shown in [Figure 7](#).

The JEITA controlled charging is suspended when the battery temperature is too cold or too hot (T < T<sub>COLD</sub> or T<sub>HOT</sub> < T).

Temperature thresholds T<sub>COLD</sub>, T<sub>COOL</sub>, T<sub>WARM</sub>, and T<sub>HOT</sub> depend on the thermistor selection. When JEITA controlled battery charge current is reduced by 50%, the charger timer is doubled.

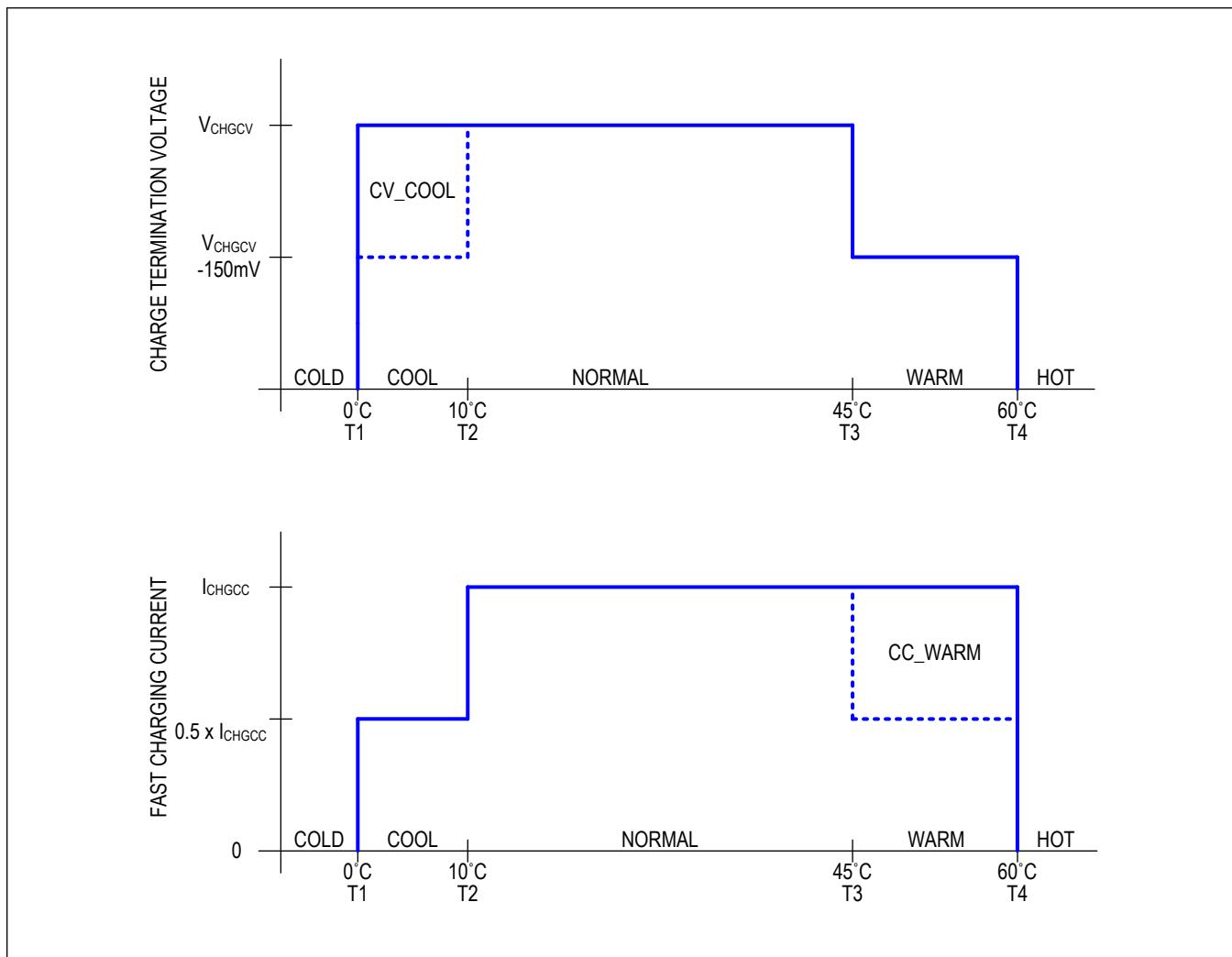


Figure 7. JEITA Controlled Charging

### Analog Low-Noise Power PVDD and VDD

VDD is the 1.8V LDO output for the charger's analog circuitry. VDD takes its power from the higher voltage of CHGIN, BATT, and SYS. VDD has a bypass capacitance of 1 $\mu$ F.

PVDD is the 1.8V LDO output for internal power circuitry. PVDD has a bypass capacitance of 1 $\mu$ F.

### Factory-Ship Mode

The ICs support factory-ship mode.

Charger's CHG\_CNFG\_07 bit 0: FSHIP\_MODE bit controls this mode.

When this bit is set to 1, the IC goes into factory-ship mode.

This mode can be exited by battery removal or on a valid charger input plug or by pulling EXTSM high longer than t<sub>EXTSM\_DEB</sub> (programmable with EXTSM\_T bit).

Factory-ship mode can not be entered when a valid charger is present.

This feature minimizes battery leakage current when the factory ships battery-connected devices.

**External QBATT Control I/O**

QBEXT is an open-drain output that is driven low in Battery mode and high-impedance (pulled-up externally) in non-battery mode.

The QBATT in MAX77986 has a very low  $R_{DSON}$  that equals  $7.7\text{m}\Omega$ . If the application requires a lower resistive discharging path then this output can be utilized to drive an external QBATT FET driver in parallel with an internal QBATT. This output can be enabled or disabled by the QBEXT\_CTRL bit.

Once this function is enabled, the BAT2SYS OCP detection is not valid and can be disabled by setting bits B2SOVRC = 0x0.

**Table 4. QBEXT Output in Different System Modes**

SYSTEM MODE	USE CASE DETAILS	QBEXT OUTPUT
Battery Mode	All use cases except non-battery mode	Low
Non-Battery Mode	Valid adapter is present, and buck is switching (whatever charge status is) or MODE = 0x09 (Boost) or MODE = 0x0A (Boost + OTG)	Hi-Z (pulled-up)

**Unplug Detection**

The QBEXT pin can also be configured to an alternative function called PGOOD. If a user writes UNPLUG\_TH[1:0] with 0b10 or 0b11, then the QBEXT pin is working as a PGOOD which is the comparator output of VBUS unplug detection.

For the 9V and 15V unplug detection, the output goes high when  $V_{CHGIN}$  voltage comes across the threshold of  $V_{CHGIN\_UNPLUG}$  with a selectable debounce by the UNPLUG\_DB bit.

For the 5V unplug case, the following are two cases:

Case1: AICL function is enabled—the output goes Hi-Z when ( $V_{CHGIN} = V_{CHGIN\_REG}$ ) AND ( $I_{CHGIN} < I_{IULO\_DET}$  with 1ms debounce time)

Case2: AICL functions is disabled—the output goes Hi-Z when ( $V_{CHGIN2SYS} < V_{CHGIN2SYS\_F}$ ) OR ( $V_{CHGIN} < V_{CHGIN\_UVLO\_F}$  (80% of  $V_{CHGIN\_UVLO}$ ))

The unplug detection does not apply to Mode 9 or Mode A.

**Note:** For parts with chip revision 0b001 (PASS1), contact Analog Devices for Rev 0 of the data sheet. This information can be read from the CHIP\_REVISION (0x01) register.

The PGOOD output state follows the below table.

**Table 5. PGOOD States Under Different Situations**

VBUS	BATTERY	MODE	PGOOD
Plugged, > programmed threshold	Yes/No	Normal	Low
Plugged, < programmed threshold	Yes/No	Normal	Hi-Z
Not plugged	Yes	Normal	Hi-Z
Not plugged	Yes	Factory-ship mode	Hi-Z
Not plugged	No	No Power	Hi-Z
Plugged, > programmed threshold	Yes	Suspend	Low
Plugged, < programmed threshold	Yes	Suspend	Hi-Z
Don't care	Yes	Deep Suspend	Hi-Z
Don't care	Yes	OTG (reverse boost)	Hi-Z

**Charge Status LED Indication**

STAT is the LED current sink shown in the following tables based on the STAT\_MODE bit.

The LED driving current can be programmed through I<sup>2</sup>C STAT\_CURR from 5mA to 20mA with a 5mA step.

**Table 6. STAT\_MODE = 0x0**

CHG STATUS	LED	DUTY (%)
No DC input or Suspend or Buck operation	Off	0
Any Charging Timeout, Off by JEITA feature, Off by thermal shutdown	Blink in 2Hz	50
DBAT, Pre-Q, CC, CV	Blink in 1Hz	50
Top-off, Done, Restart	Solid on	100

**Table 7. STAT\_MODE = 0x1**

CHG STATUS	LED	DUTY (%)
No DC input or Suspend or Buck operation	Off	0
Any Charging Timeout, Off by JEITA feature, Off by thermal shutdown	Off	0
DBAT, Pre-Q, CC, CV	Blink in 1Hz	50
Top-off, Done, Restart	Solid on	100

### Audio Mode

When Audio Mode is enabled by writing AUDIO\_MODE = 0b1, the switching frequency is limited to at least 64kHz so that the switching frequency is always higher than the audible frequency. Audio mode affects bootstrap refresh frequency and spread spectrum.

### Spread Spectrum

The buck converter is capable of dithering its switching frequency for noise-sensitive applications. The dithering allows the buck converter to be operated not at a single fixed frequency but over a varying band of frequencies. This helps reduce the peak value of EMI emission. The spread-spectrum function is activated only in CCM (Continuous Conduction Mode) and it is automatically deactivated when the buck converter enters DCM (Discontinuous Conduction Mode). The spread-spectrum function can be disabled by the SS\_ENV[1:0] bits.

Modulation envelope (ΔFSS) determines the maximum difference between the modulated switching frequency and the nominal switching frequency. The modulation envelope is programmable ( $\pm 4\%$ ,  $\pm 8\%$ , or  $\pm 16\%$ ) by the SS\_ENV[1:0] bits and it controls 'how wide' the switching frequency dithers.

Two modulation patterns are determined by the SS\_PAT bit. One is triangular and the other is pseudo-random.

The modulation frequency determines how often the switching frequency dithering cycle repeats. In a triangular pattern, the modulation frequency is fixed at 2.133kHz. In pseudo-random, it is 4.267kHz. When audio mode is enabled, then the modulation frequency is automatically changed to 133Hz for a triangular pattern and to 267Hz for a pseudo-random pattern.

### Programmable Minimum System Voltage

MAX77985/MAX77986 supports programmable MINSYS (minimum system voltage) by MINSYS[2:1] value. It is recommended to position the Min SYS greater than the minimum system supply voltage for proper operation.

**Table 8. Minimum System Voltage Values for Different Part Variants**

VALUE	MAX77985A/MAX77986A (V)	MAX77985B/MAX77986B (V)
0b00	3.4	3.0
0b01	3.5	3.1
0b10	3.6	3.5
0b11	3.7	3.6

## Top System Management

### Overview

This section discusses the top system of the MAX77985/MAX77986 and how the IC manages its bias, system faults, and turn-on and off events.

### Main Bias

The main bias includes voltage and current references for all circuitry that runs from the  $V_{SYS}$  node.

### System Faults

#### $V_{SYS}$ Fault

The system monitors the  $V_{SYS}$  node for undervoltage and overvoltage events. The following describes the IC behavior if any of these events are to occur.

#### $V_{SYS}$ Undervoltage Lockout ( $V_{SYSUVLO}$ )

$V_{SYS}$  undervoltage lockout prevents the regulators from being used when the input voltage is below the operating range. When the voltage from SYS to GND ( $V_{SYS}$ ) is less than the undervoltage-lockout threshold ( $V_{SYSUVLO}$ ), MAX77985/MAX77986 shuts down and resets the "O" Type I<sup>2</sup>C registers.

#### $V_{SYS}$ Overvoltage Lockout ( $V_{SYSOVLO}$ )

$V_{SYS}$  overvoltage lockout is a fail-safe mechanism and prevents the regulators from being used when the input voltage is above the operating range. The absolute maximum ratings state that the SYS node withstands up to 6V. The SYS OVLO threshold is set to 5.35V (typ)—ideally,  $V_{SYS}$  should not exceed the battery charge termination threshold. Systems must be designed such that  $V_{SYS}$  never exceeds 5.2V (transient and steady-state). If the  $V_{SYS}$  exceeds  $V_{SYS\_OVLO\_R}$ , the ICs shuts down and resets the "O" Type I<sup>2</sup>C registers.

#### $V_{SYS}$ Power-Up Failure (PWRUPFAIL)

$V_{SYS}$  power-up failure is a hardware diagnostic mechanism to detect failures affecting the system and preventing the platform from powering up. When a **valid** power source (battery  $V_{BATT} > V_{SYS\_UVLOB\_R}$  or charger with  $V_{CHGIN} > V_{CHGIN\_UVLO}$ ) is plugged, the MAX77985/MAX77986 is expected to pull the SYS node up by utilizing one of the system power-up current sources ( $I_{SYSBU\_BAT}$  or  $I_{SYSBU\_BYP}$  respectively). If  $V_{SYS}$  does not rise above  $V_{SYSBU}$  due to a fault in the application (external to MAX77986), after a time-out elapses ( $t_{SYSBU\_BAT}$  or  $t_{SYSBU\_BYP}$  respectively) a power-up fault is asserted and an interrupt (PWRUP\_FAIL\_INT) is generated. Because the SYS node is down, the application software may not be able to service the interrupt; the interrupt can only be observed by pulling VIO up externally and serviced by taking control of the I<sup>2</sup>C interface.

#### Thermal Fault

The ICs have one centralized thermal circuit which senses the temperature on the die. If the temperature increases  $>155^{\circ}\text{C}$  ( $T_{SHDN}$ ) this constitutes a thermal shutdown event and the MAX77986 shuts down and resets the "O" Type I<sup>2</sup>C registers. There is a  $15^{\circ}\text{C}$  thermal hysteresis. After thermal shutdown, if the die temperature is reduced by  $15^{\circ}\text{C}$ , the thermal shutdown bus is deasserted and the IC can be enabled again. The main battery charger has an independent thermal control loop which does not cause a thermal shutdown event. If a charger thermal overload occurs, only the charger turns off.

#### S-Type and O-Type Registers

S-type registers include: TOP\_FUNC registers 0x00, 0x01, 0x02, 0x03, 0x04, 0x05, 0x51; I2C\_FUNC register 0x40; CHARGER\_FUNC registers 0x10, 0x12, 0x13, 0x14, 0x15.

O-type registers include: TOP\_FUNC register 0x50, CHARGER\_FUNC registers 0x11, and all registers from 0x16 to 0x24.

#### Charger Register Write Protection

CHG\_CNFG register 1, 2, 3, 4, 5, 7, 8 (CHARGER\_FUNC register address 0x17, 0x18, 0x19, 0x1A, 0x1B, 0x1D, 0x1E) are protected by CHG\_CNFG\_06.CHGPROT bitfield. By default, these configurations are not writable, and need unlocking by writing bitfield CHGPROT = 0b11 first.

**System Faults Debounce Time**

Applicable in charge or buck mode.

**Table 9. System Faults Debounce Time Summary**

	EDGE TO I/T		I/T TO FAULT		ACTION ON FAULT
	t <sub>DEB</sub> (Rising)	t <sub>DEB</sub> (Falling)	t <sub>DEB</sub> (Rising)	t <sub>DEB</sub> (Falling)	
SYS UVLO	—	—	8ms	—	O-Type reset
SYS OVLO	*-/5ms by I <sup>2</sup> C	—	—	—	O-Type reset
TSHDN	175µs	—	—	—	O-Type reset
BATT OCP	t <sub>BOVRC</sub>	—	t <sub>OCP</sub>	—	Q <sub>BATT</sub> opens
OTG OCP	t <sub>OTG_ALARM</sub>	—	t <sub>OTG_FAULT</sub> - t <sub>OTG_ALARM</sub>	—	RBFET opens

(\*) Depending on I<sup>2</sup>C bit SYSOVLO\_DEB\_EN**I<sup>2</sup>C Interface Description****Main I<sup>2</sup>C Interface**

The IC acts as a Slave Transmitter/Receiver and has the following slave addresses:

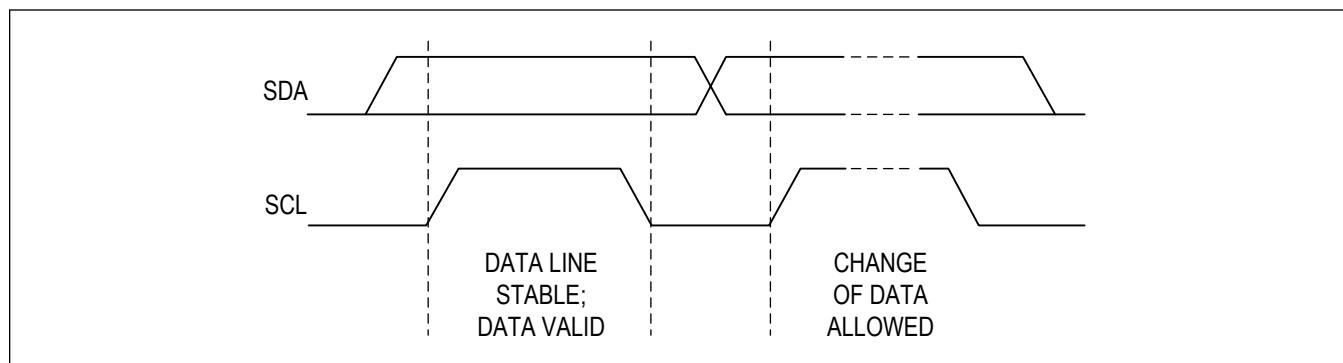
Slave Address (7 bit) 0x6B 110 1011

Slave Address (Write) 0xD6 1101 0110

Slave Address (Read) 0xD7 1101 0111

**I<sup>2</sup>C Bit Transfer**

One data bit is transferred for each clock pulse. The data on SDA must remain stable during the high portion of the clock pulse as changes in data during this time are interpreted as a control signal.

Figure 8. I<sup>2</sup>C Bit Transfer

### I<sup>2</sup>C Start and Stop Conditions

Both SDA and SCL remain High when the bus is not busy. The Start (S) condition is defined as a high-to-low transition of the SDA while the SCL is high. The Stop (P) condition is defined as a low-to-high transition of the SDA while the SCL is high.

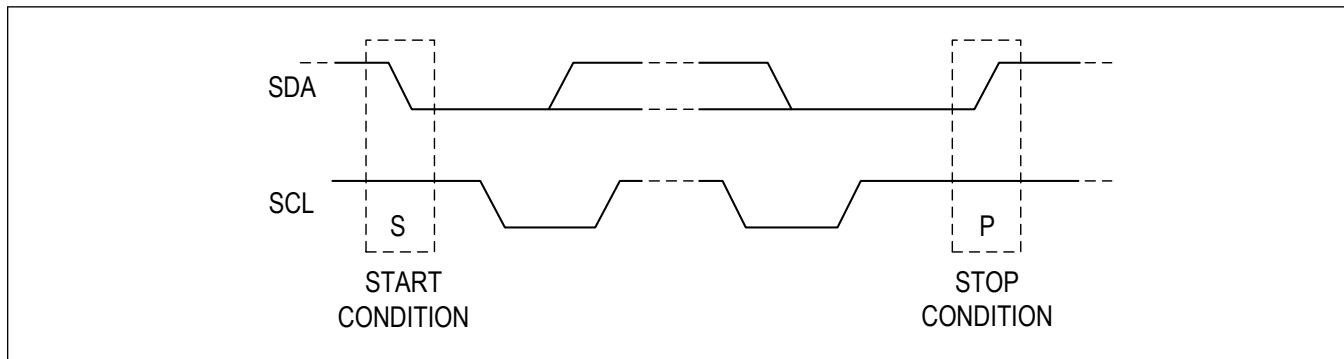


Figure 9. I<sup>2</sup>C Start and Stop

### I<sup>2</sup>C System Configuration

A device on the I<sup>2</sup>C bus that generates a “message” is called a “Transmitter” and a device that receives the message is a “Receiver”. The device that controls the message is the “Master” and the devices that are controlled by the “Master” are called “Slaves”.

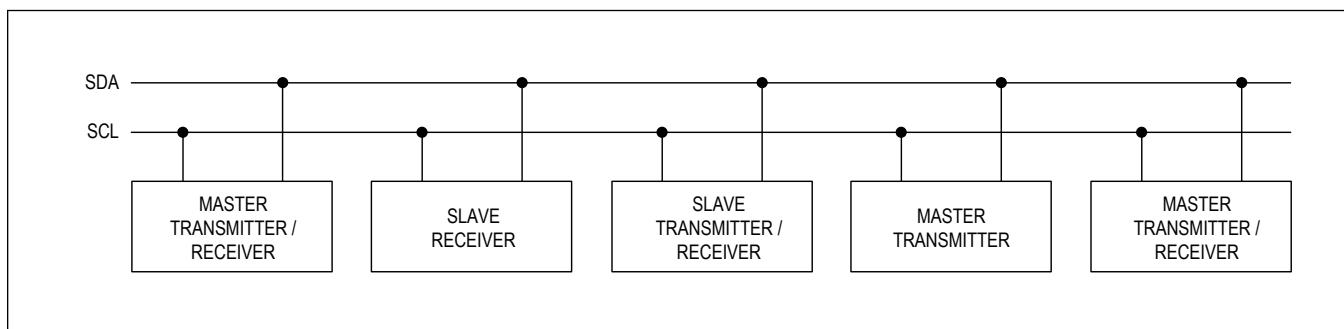


Figure 10. System Configurations

## I<sup>2</sup>C Acknowledge

The number of data bytes between the start and stop conditions for the Transmitter and Receiver are unlimited.

Each 8-bit byte is followed by an Acknowledge bit. The Acknowledge bit is a high-level signal put on SDA by the transmitter during which time the master generates an extra acknowledge-related clock pulse. A slave receiver that is addressed must generate an acknowledge after each byte it receives. Also a master receiver must generate an acknowledge after each byte it receives that has been clocked out of the slave transmitter.

The device that acknowledges must pulldown the SDA line during the acknowledge-clock pulse so that the SDA line is stable and low during the high period of the acknowledge-clock pulse (setup and hold times must also be met). A master receiver must signal the end of data to the transmitter by not generating an acknowledge on the last byte that has been clocked out of the slave. In this case, the transmitter must leave SDA high to enable the master to generate a stop condition.

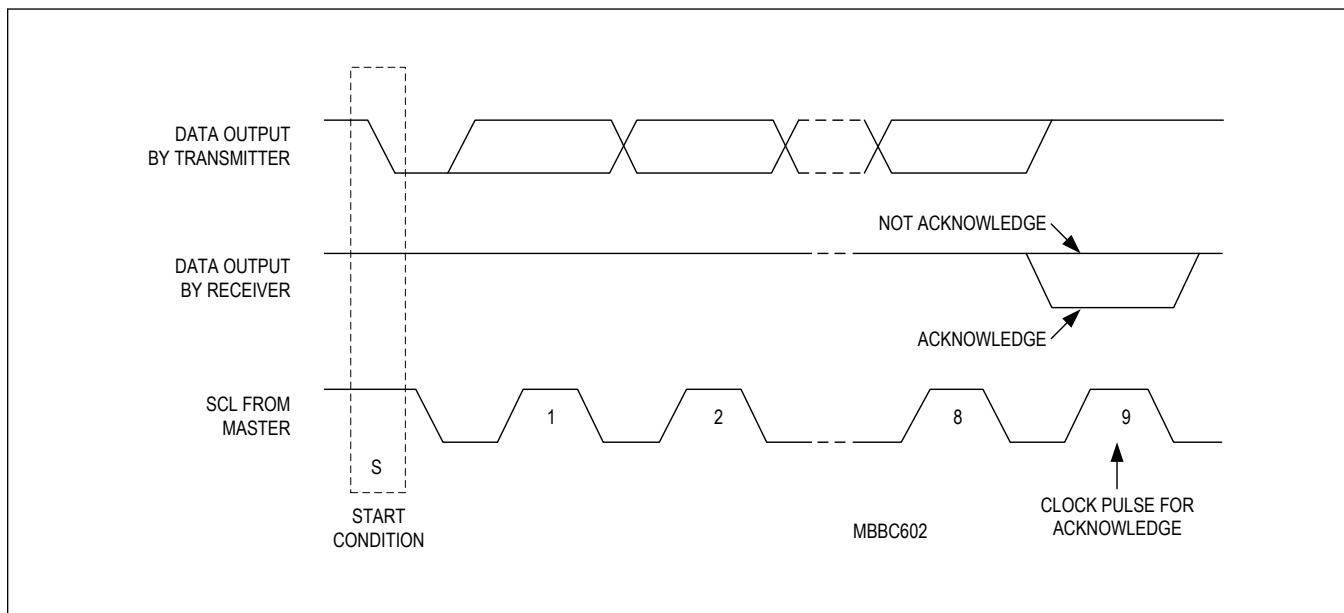


Figure 11. I<sup>2</sup>C Acknowledge

## Master Transmits (Write Mode)

Use the following format when the master writes to the slave.

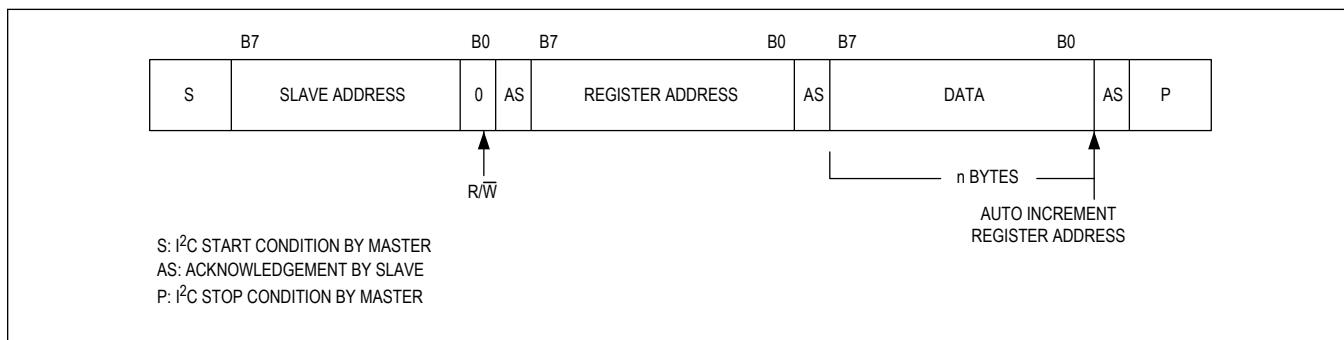


Figure 12. I<sup>2</sup>C Master Transmits

**Master Reads after Setting Register Address (Write Register Address and Read Data)**

Use the following format to read a specific register.

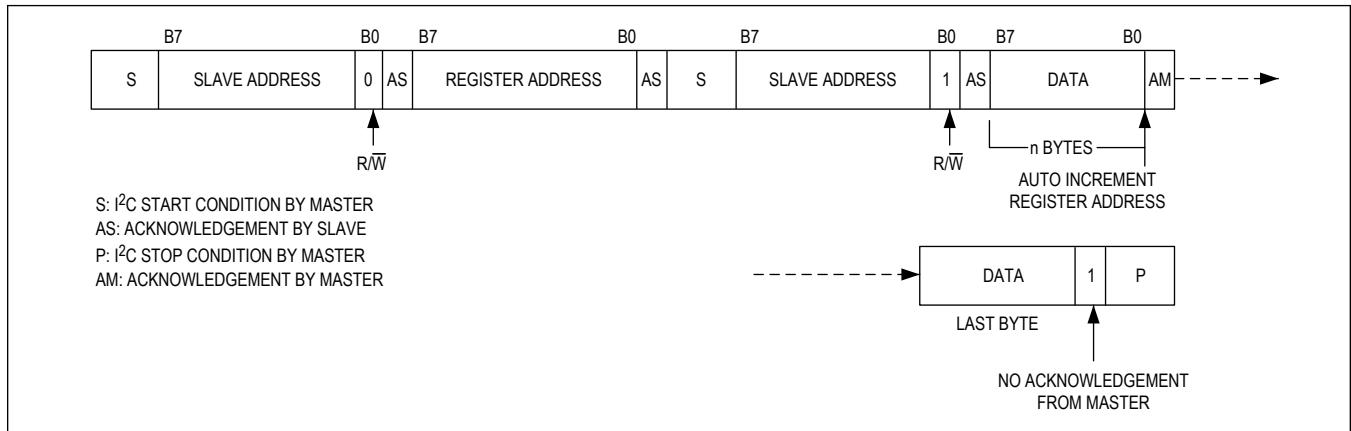


Figure 13. I<sup>2</sup>C Master Reads After Setting Register Address

**Master Reads Register Data Without Setting Register Address (Read Mode)**

Use the following format to read registers continuously starting from first address.

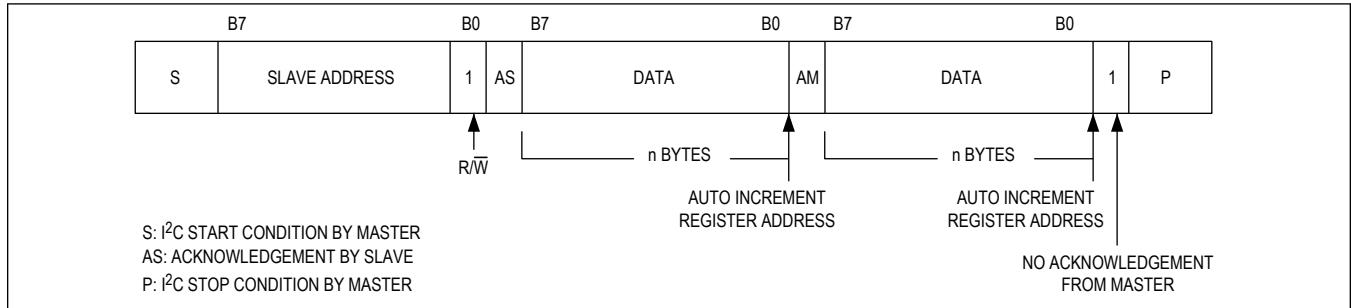


Figure 14. I<sup>2</sup>C Master Block Read

## Register Map

### TOP

#### I<sup>2</sup>C Slave Address

Slave Address (7 bit) 0x6B (7'b110 1011)  
 Slave Address (Write) 0xD6 (8'b1101 0110)  
 Slave Address (Read) 0xD7 (8'b1101 0111)

#### Functional Reset Conditions

The chip has different levels of reset as defined below:

- Type S: Registers are reset each time when: SYS < VDD (1.8V)
- Type O: Registers are reset each time when: SYS < VDD or SYS < SYS UVLO or SYS > SYS OVLO or die temp > TSHDN or software reset (SW\_RST)

ADDRESS	NAME	MSB							LSB			
<b>TOP_FUNC</b>												
0x00	<a href="#">CHIP_ID[7:0]</a>	ID[7:0]										
0x01	<a href="#">CHIP_REVISION[7:0]</a>	VERSION[3:0]				REVISION[3:0]						
0x02	<a href="#">OTP_REVISION[7:0]</a>	SPR_7_4[3:0]				OTP_REV[3:0]						
0x03	<a href="#">TOP_INT[7:0]</a>	SPR_7	TSHDN_INT	SYSOVL_O_INT	SYSUVL_O_INT	SPR_3_1[2:0]			PWRUP_FAIL_INT			
0x04	<a href="#">TOP_INT_MASK[7:0]</a>	SPR_7	TSHDN_INT_M	SYSOVL_O_INT_M	SYSUVL_O_INT_M	SPR_3_1[2:0]			PWRUP_FAIL_INT_M			
0x05	<a href="#">TOP_CTRL[7:0]</a>	–	SPR_6_4[2:0]			LPM	SYSOVL_O_DIS	SYSOVL_O_DEB_EN	TSHDN_DIS			
0x50	<a href="#">SW_RESET[7:0]</a>	SWR_RST[7:0]										
0x51	<a href="#">SM_CTRL[7:0]</a>	SPR_7_1[6:0]							EXTSM_T			
<b>I<sup>2</sup>C_FUNC</b>												
0x40	<a href="#">I2C_CNFG[7:0]</a>	SPR_7	RSVD[1:0]		PAIR	SPR_3_1[2:0]			HS_EXT_EN			

## Register Details

### [CHIP\\_ID \(0x0\)](#)

#### PMIC ID

BIT	7	6	5	4	3	2	1	0
Field	ID[7:0]							
Reset	0x86							
Access Type	Read Only							
BITFIELD	BITS	DESCRIPTION			DECODE			
ID	7:0	ID of MAX77986/MAX77985			0x86: MAX77986 0x85: MAX77985			

[CHIP\\_REVISION \(0x1\)](#)

PMIC revision

BIT	7	6	5	4	3	2	1	0
<b>Field</b>	VERSION[3:0]					REVISION[3:0]		
<b>Reset</b>	0xA or 0xB					0b010		
<b>Access Type</b>	Read Only					Read Only		

BITFIELD	BITS	DESCRIPTION	DECODE
VERSION	7:4	Version ( <b>Note:</b> For parts with chip revision 0b001 (PASS1), contact Analog Devices for Rev 0 of the data sheet. This information can be read from the CHIP_REVISION (0x01) register.)	0b1010: MAX77985/MAX77986 A variation 0b1011: MAX77985/MAX77986 B variation
REVISION	3:0	Revision ( <b>Note:</b> For parts with chip revision 0b001 (PASS1), contact Analog Devices for Rev 0 of the data sheet.)	0b001: PASS1 0b010: PASS2 0b011: Reserved 0b100: Reserved

[OTP\\_REVISION \(0x2\)](#)

BIT	7	6	5	4	3	2	1	0
<b>Field</b>	SPR_7_4[3:0]					OTP_REV[3:0]		
<b>Reset</b>	0x0					0x1		
<b>Access Type</b>	Read Only					Read Only		

BITFIELD	BITS	DESCRIPTION
SPR_7_4	7:4	
OTP_REV	3:0	Revision

[TOP\\_INT \(0x3\)](#)

Top SYS Interrupts

BIT	7	6	5	4	3	2	1	0
<b>Field</b>	SPR_7	TSHDN_IN_T	SYSOVLO_INT	SYSUVLO_INT	SPR_3_1[2:0]			PWRUP_FAIL_INT
<b>Reset</b>	0b0	0b0	0b0	0b0	0x0			0b0
<b>Access Type</b>	Read Clears All			Read Clears All				

BITFIELD	BITS	DESCRIPTION	DECODE
SPR_7	7		
TSHDN_INT	6	Thermal Shutdown Interrupt (entering fault condition)	0b0: No interrupt. 0b1: Interrupt is detected.
SYSOVLO_INT	5	SYSOVLO Interrupt (entering fault condition)	0b0: No interrupt. 0b1: Interrupt is detected.
SYSUVLO_INT	4	SYSUVLO Interrupt (entering fault condition)	0b0: No interrupt. 0b1: Interrupt is detected.
SPR_3_1	3:1		

BITFIELD	BITS	DESCRIPTION	DECODE
PWRUP_FAI_L_INT	0	Power-Up Fail Interrupt (entering fault condition)	0b0: No interrupt. 0b1: Interrupt is detected.

### TOP\_INT\_MASK (0x4)

Top SYS Interrupt Mask

BIT	7	6	5	4	3	2	1	0
Field	SPR_7	TSHDN_IN_T_M	SYSOVLO_INT_M	SYSUVLO_INT_M	SPR_3_1[2:0]			PWRUP_FAIR_INT_M
Reset	0b1	0b1	0b1	0b1	0x7			0b0
Access Type	Write, Read	Write, Read	Write, Read	Write, Read	Write, Read			Write, Read

BITFIELD	BITS	DESCRIPTION	DECODE
SPR_7	7		
TSHDN_INT_M	6	Thermal Shutdown Interrupt Mask	0b0: Unmasked 0b1: Masked
SYSOVLO_INT_M	5	SYSOVLO Interrupt Mask	0b0: Unmasked 0b1: Masked
SYSUVLO_INT_M	4	SYSUVLO Interrupt Mask	0b0: Unmasked 0b1: Masked
SPR_3_1	3:1		
PWRUP_FAIR_INT_M	0	Powe-Up Fail Interrupt Mask	0b0: Unmasked 0b1: Masked

### TOP\_CTRL (0x5)

Main Control1

BIT	7	6	5	4	3	2	1	0
Field	—	SPR_6_4[2:0]		LPM	SYSOVLO_DIS	SYSOVLO_DEB_EN	TSHDN_DIS	
Reset	—	0b000		0b0	0b0	0b0	0b0	0b1
Access Type	—	Write, Read		Write, Read	Write, Read	Write, Read	Write, Read	Write, Read

BITFIELD	BITS	DESCRIPTION	DECODE
SPR_6_4	6:4		
LPM	3	Low-Power Mode Cycling mode is allowed for SYS UVLO, SYS OVLO, and THERM comparators.	0: Low-power mode is disabled. SYUVLO comparator is always ON. SYOVLO comparator is controlled by SYOVLO_DIS. THERM comparator is controlled by TSHDN_DIS. 1: Low-power mode is allowed. Comparators are periodically enabled (depending on SYOVLO_DIS/TSHDN_DIS control)/disabled and cycling every 3ms.
SYOVLO_DIS	2	SYOVLO Disable	0: SYOVLO comparator is enabled. 1: SYOVLO comparator is disabled.
SYOVLO_DEB_EN	1	SYOVLO Debounce (rising 5ms)	0: SYOVLO debounce is disabled. 1: SYOVLO debounce is enabled.

BITFIELD	BITS	DESCRIPTION	DECODE
TSHDN_DIS	0	Internal Die Temperature Shutdown Disable Bit. If a valid CHGIN is detected, this bit has no effect and the Internal Die Temperature Comparator is permanently enabled.	1: TSHDN comparator is disabled unless a valid CHGIN is detected. 0: TSHDN comparator is enabled.

**SW\_RESET (0x50)**

SW-reset register

BIT	7	6	5	4	3	2	1	0
<b>Field</b>	SWR_RST[7:0]							
<b>Reset</b>	0x00							
<b>Access Type</b>	Write, Read							
BITFIELD	BITS	DESCRIPTION			DECODE			
SWR_RST	7:0	Software Reset			0xA5: O-Type registers are reset.			

**SM\_CTRL (0x51)**

SW-reset register

BIT	7	6	5	4	3	2	1	0
<b>Field</b>	SPR_7_1[6:0]							
<b>Reset</b>	0x00							
<b>Access Type</b>	Write, Read							
BITFIELD	BITS	DESCRIPTION			DECODE			
SPR_7_1	7:1							
EXTSM_T	0	External Ship Mode Timer			0b0: 10ms 0b1: 0.1ms			

**I2C\_CNFG (0x40)**

BIT	7	6	5	4	3	2	1	0		
<b>Field</b>	SPR_7	RSVD[1:0]		PAIR	SPR_3_1[2:0]		HS_EXT_E_N			
<b>Reset</b>	0b0	0b0		0b000	0b000		0b0			
<b>Access Type</b>	Write, Read	Write, Read		Write, Read	Write, Read		Write, Read			
BITFIELD	BITS	DESCRIPTION			DECODE					
SPR_7	7	Reserved								
RSVD	6:5	Reserved								
PAIR	4	Pair address mode option for register write burst operation.			1 = Pair address mode is enabled for the channel. 0 = Pair address mode is disabled and sequential mode is used.					
SPR_3_1	3:1									

BITFIELD	BITS	DESCRIPTION	DECODE					
HS_EXT_EN	0	Enable HS-Mode Extension	0b0: HS-mode extension is disabled. (I <sup>2</sup> C Rev. 4 Compliant) 0b1: HS-mode extension is enabled. HS-mode is enabled without HS-mode entrance code and keeps HS-mode during STOP condition.					

**CHARGER**

ADDRESS	NAME	MSB							LSB					
<b>CHARGER_FUNC</b>														
0x10	<a href="#">CHG_INT[7:0]</a>	AICL_I	CHGIN_I	INLIM_I	CHG_I	BAT_I	RSVD_2	DISQBA_T_I	BYP_I					
0x11	<a href="#">CHG_INT_MASK[7:0]</a>	AICL_M	CHGIN_M	INLIM_M	CHG_M	BAT_M	SPR_2	DISQBA_T_M	BYP_M					
0x12	<a href="#">CHG_INT_OK[7:0]</a>	AICL_O_K	CHGIN_OK	INLIM_O_K	CHG_O_K	BAT_OK	RSVD_2	DISQBA_T_OK	BYP_OK					
0x13	<a href="#">CHG_DETAILS_00[7:0]</a>	RSVD_7	CHGIN_DTLS[1:0]	RSVD_4_3[1:0]		SPSN_DTLS[1:0]	RSVD_0							
0x14	<a href="#">CHG_DETAILS_01[7:0]</a>	TREG	BAT_DTLS[2:0]			CHG_DTLS[3:0]								
0x15	<a href="#">CHG_DETAILS_02[7:0]</a>	RSVD_7	THM_DTLS[2:0]			BYP_DTLS[3:0]								
0x16	<a href="#">CHG_CNFG_00[7:0]</a>	SS_ENV[1:0]		SS_PAT	BATRMV_MSK	MODE[3:0]								
0x17	<a href="#">CHG_CNFG_01[7:0]</a>	TKEN	WDTEN	CHG_RSTRT[1:0]		SPR_3	FCHGTIME[2:0]							
0x18	<a href="#">CHG_CNFG_02[7:0]</a>	SPR_7	CHG_CC[6:0]											
0x19	<a href="#">CHG_CNFG_03[7:0]</a>	SPR_7	TO_TIME[2:0]			TOITH[3:0]								
0x1A	<a href="#">CHG_CNFG_04[7:0]</a>	SYS_TR ACK_DI S	MINSYS[1:0]		CHG_CV_PRM[4:0]									
0x1B	<a href="#">CHG_CNFG_05[7:0]</a>	B2SOVR C_DTC	B2SOVR C_ALAR M_ONLY	B2SOVR C_CTRL	RECYCL E_EN	B2SOVRC[3:0]								
0x1C	<a href="#">CHG_CNFG_06[7:0]</a>	SPR_7_4[3:0]				CHGPROT[1:0]		WDTCLR[1:0]						
0x1D	<a href="#">CHG_CNFG_07[7:0]</a>	WD_QB ATOFF	SPR_6	DISIBS	SPSN_D ET_EN	QBEXT CTRL_E N	SPR_2_1[1:0]		FSHIP MODE					
0x1E	<a href="#">CHG_CNFG_08[7:0]</a>	AUDIO_MODE	BCKSYS[2:0]			FMBST	SLOWLX	FSW	DISKIP					
0x1F	<a href="#">CHG_CNFG_09[7:0]</a>	SPR_7	CHGIN_ILIM[6:0]											
0x20	<a href="#">CHG_CNFG_10[7:0]</a>	OTG_RE C_EN	INLIM_CLK[1:0]		OTG_ILIM[4:0]									
0x21	<a href="#">CHG_CNFG_11[7:0]</a>	SPR_7	VBYPSET[6:0]											
0x22	<a href="#">CHG_CNFG_12[7:0]</a>	BYPDIS CHG_EN	DEEP_S USP_DI S	VCHGIN_REG[1:0]		UNPLUG_TH[1:0]	UNPLUG _DB	DIS_AIC L						
0x23	<a href="#">CHG_CNFG_13[7:0]</a>	JEITA_E N	SPR_6	CHG_C _COOL	CHG_C _WAR M	REGTEMP[3:0]								
0x24	<a href="#">STAT_CNFG[7:0]</a>	STAT_E N	SPR_6_4[2:0]			STAT_CURR[1:0]	SPR_1	STAT_M ODE						

**Register Details****CHG\_INT (0x10)**

Interrupt status register for the charger block.

BIT	7	6	5	4	3	2	1	0
<b>Field</b>	AICL_I	CHGIN_I	INLIM_I	CHG_I	BAT_I	RSVD_2	DISQBAT_I	BYP_I
<b>Reset</b>						0x0		
<b>Access Type</b>	Read Clears All	Read Clears All	Read Clears All	Read Clears All	Read Clears All	Read Clears All	Read Clears All	Read Clears All
BITFIELD	BITS	<b>DESCRIPTION</b>				<b>DECODE</b>		
AICL_I	7	AICL Interrupt				0b0: The AICL_OK bit has not changed since the last time this bit was read. 0b1: The AICL_OK bit has changed since the last time this bit was read.		
CHGIN_I	6	CHGIN Interrupt				0b0: The CHGIN_OK bit has not changed since the last time this bit was read. 0b1: The CHGIN_OK bit has changed since the last time this bit was read.		
INLIM_I	5	Input Current Limit Interrupt				0b0: The INLIM_OK bit has not changed since the last time this bit was read. 0b1: The INLIM_OK bit has changed since the last time this bit was read.		
CHG_I	4	Charger Interrupt				0b0: The CHG_OK bit has not changed since the last time this bit was read. 0b1: The CHG_OK bit has changed since the last time this bit was read.		
BAT_I	3	Battery Interrupt				0b0: The BAT_OK bit has not changed since the last time this bit was read. 0b1: The BAT_OK bit has changed since the last time this bit was read.		
RSVD_2	2							
DISQBAT_I	1	DISQBAT Interrupt				0b0: The DISQBAT_OK bit has not changed since the last time this was read. 0b1: The DISQBAT_OK bit has changed since the last time this was read.		
BYP_I	0	Bypass Node Interrupt				0b0: The BYP_OK bit has not changed since the last time this bit was read. 0b1: The BYP_OK bit has changed since the last time this bit was read.		

**CHG\_INT\_MASK (0x11)**

Mask register to mask the corresponding charger interrupts.

BIT	7	6	5	4	3	2	1	0
<b>Field</b>	AICL_M	CHGIN_M	INLIM_M	CHG_M	BAT_M	SPR_2	DISQBAT_M	BYP_M
<b>Reset</b>	0b1	0b1	0b1	0b1	0b1	0x1	0b1	0b1
<b>Access Type</b>	Write, Read							

BITFIELD	BITS	DESCRIPTION	DECODE
AICL_M	7	AICL Interrupt Mask	0b0: Unmasked 0b1: Masked
CHGIN_M	6	CHGIN Interrupt Mask	0b0: Unmasked 0b1: Masked
INLIM_M	5	Input Current Limit Interrupt Mask	0b0: Unmasked 0b1: Masked
CHG_M	4	Charger Interrupt Mask	0b0: Unmasked 0b1: Masked
BAT_M	3	Battery Interrupt Mask	0b0: Unmasked 0b1: Masked
SPR_2	2		
DISQBAT_M	1	DISQBAT Interrupt Mask	0b0: Unmasked 0b1: Masked
BYP_M	0	Bypass Interrupt Mask	0b0: Unmasked 0b1: Masked

[CHG\\_INT\\_OK \(0x12\)](#)

BIT	7	6	5	4	3	2	1	0
<b>Field</b>	AICL_OK	CHGIN_OK	INLIM_OK	CHG_OK	BAT_OK	RSVD_2	DISQBAT_OK	BYP_OK
<b>Reset</b>	0x1	0x0	0x1	0x1	0x1	0x0	0x1	0x1
<b>Access Type</b>	Read Only	Read Only						

BITFIELD	BITS	DESCRIPTION	DECODE
AICL_OK	7	AICL_OK Status	0b0: AICL mode. 0b1: Not in AICL mode.
CHGIN_OK	6	CHGIN Input Status Indicator	0b0: The CHGIN input is invalid. CHGIN_DTLS≠0x03 0b1: The CHGIN input is valid. CHGIN_DTLS=0x03
INLIM_OK	5	Input Current Limit Status Indicator	0b0: The CHGIN input current has been reaching the current limit for at least 30ms. 0b1: The CHGIN input current has not reached the current limit.
CHG_OK	4	Charger Status Indicator	0b0: The charger has suspended charging or TREG = 1. 0b1: The charger is okay or the charger is off.
BAT_OK	3	Battery Status Indicator	0b0: The battery has an issue or the charger has been suspended. BAT_DTLS≠0x03, ≠0x04 and ≠0x07. 0b1: The battery is okay. BAT_DTLS = 0x03,0x04 or 0x07.
RSVD_2	2		
DISQBAT_OK	1	DISQBAT Status Indicator	0b0: DISQBAT is high and QBATT is disabled. 0b1: DISQBAT is low and QBATT is not disabled.
BYP_OK	0	Bypass Status Indicator	0b0: Something powered by the bypass node has hit current limit. BYP_DTLS≠0x00. 0b1: The bypass node is okay. BYP_DTLS=0x00.

[CHG\\_DETAILS\\_00 \(0x13\)](#)

BIT	7	6	5	4	3	2	1	0
<b>Field</b>	RSVD_7	CHGIN_DTLS[1:0]			RSVD_4_3[1:0]			SPSN_DTLS[1:0]
<b>Reset</b>	0x0				0x0			0b0
<b>Access Type</b>	Read Only	Read Only			Read Only			Read Only
BITFIELD	BITS	DESCRIPTION			DECODE			
RSVD_7	7							
CHGIN_DTLS	6:5	CHGIN Details			0b00: VBUS is invalid. $V_{CHGIN} \leq V_{CHGIN\_REG}$ AND $I_{CHGIN} < I_{UL0\_DET}$ for 1ms AND $DIS\_AICL=0$ OR $[V_{CHGIN} < V_{CHGIN\_UVLO\_F}]$ 0b01: VBUS is invalid. $V_{CHGIN} < V_{BATT} + V_{CHGIN2SYS\_F}$ and $V_{CHGIN} > V_{CHGIN\_UVLO}$ (80% of $V_{CHGIN\_UVLO}$ ) 0b10: VBUS is invalid. $V_{CHGIN} > V_{CHGIN\_OVLO}$ 0b11: VBUS is valid. $V_{CHGIN} > V_{CHGIN\_UVLO}$ and $V_{CHGIN} > V_{BATT} + V_{CHGIN2SYS\_TH}$ and $V_{CHGIN} < V_{CHGIN\_OVLO}$			
RSVD_4_3	4:3							
SPSN_DTLS	2:1	SP/SN Remote Sense Battery Line Connection Status			0b00: SPSN remote sense line is connected. 0b01: SP remote sense line detected as opened. 0b10: SN remote sense line detected as opened. 0b11: SP and SN remote sense lines are both detected as opened.			
RSVD_0	0	Spare Bit						

[CHG\\_DETAILS\\_01 \(0x14\)](#)

BIT	7	6	5	4	3	2	1	0
<b>Field</b>	TREG	BAT_DTLS[2:0]			CHG_DTLS[3:0]			
<b>Reset</b>								
<b>Access Type</b>	Read Only	Read Only			Read Only			
BITFIELD	BITS	DESCRIPTION			DECODE			
TREG	7	Temperature Regulation Status			0b0: The junction temperature is less than the threshold set by REGTEMP and the full charge current limit is available. 0b1: The junction temperature is greater than the threshold set by REGTEMP and the charge current limit may be folding back to reduce power dissipation.			

BITFIELD	BITS	DESCRIPTION	DECODE
BAT_DTLS	6:4	Battery Details	<p>0b000: Battery Removal A valid adapter is present and the battery is detached, detected on the THM pin.</p> <p>0b001: Battery Prequalification Voltage A valid adapter is present and the battery voltage is low: <math>V_{BATT} &lt; V_{TRICKLE}</math>. <b>Note:</b> This condition is also reported in the CHG_DTLS as 0x00.</p> <p>0b010: Battery Timer Fault A valid adapter is present and the battery has taken longer than expected to charge (exceeded <math>t_{FC}</math>). This could be due to high system currents, an old battery, a damaged battery, or something else. Charging has been suspended and the charger is in timer-fault mode. <b>Note:</b> This condition is also reported in the CHG_DTLS as 0x06.</p> <p>0b011: Battery Regular Voltage A valid adapter is present and the battery voltage is greater than the minimum system regulation level but lower than the overvoltage level: <math>V_{SYSMIN} &lt; V_{BATT} &lt; V_{BATTREG} + V_{COV}</math> <math>V_{SYS}</math> is approximately equal to <math>V_{BATT}</math>.</p> <p>0b100: Battery Low Voltage A valid adapter is present and the battery voltage is lower than the minimum system regulation level but higher than the prequalification voltage: <math>V_{TRICKLE} &lt; V_{BATT} &lt; V_{SYSMIN}</math> <math>V_{SYS}</math> is regulated at least equal to <math>V_{SYSMIN}</math>.</p> <p>0b101: Battery Overvoltage A valid adapter is present and the battery voltage is greater than the battery-overvoltage threshold (<math>V_{BATTREG} + V_{COV}</math>) for the last 30ms. <b>Note:</b> This flag is only generated when there is a valid input.</p> <p>0b110: Battery Overcurrent The battery current has exceeded the battery-overcurrent threshold (<math>I_{BOVRC}</math>) for at least 3ms since the last time this register was read.</p> <p>0b111: Battery Only, No Overcurrent No valid adapter is present and the battery current is lower than the battery-overcurrent threshold (<math>I_{BOVRC}</math>). The battery voltage and battery removal monitoring are not available. <b>Note:</b> In the case of deep suspend, it is considered that no valid adapter is present.</p>

BITFIELD	BITS	DESCRIPTION	DECODE
CHG_DTLS	3:0	Charger Details	<p>0x00: Charger is in dead-battery prequalification or low-battery prequalification mode. CHG_OK = 1 and <math>V_{BATT} &lt; V_{PQLB}</math> and <math>T_J &lt; T_{SHDN}</math></p> <p>0x01: Charger is in fast-charge constant current mode. CHG_OK = 1 and <math>V_{BATT} &lt; V_{BATTREG}</math> and <math>T_J &lt; T_{SHDN}</math></p> <p>0x02: Charger is in fast-charge constant voltage mode. CHG_OK = 1 and <math>V_{BATT} = V_{BATTREG}</math> and <math>T_J &lt; T_{SHDN}</math></p> <p>0x03: Charger is in top-off mode. CHG_OK = 1 and <math>V_{BATT} = V_{BATTREG}</math> and <math>T_J &lt; T_{SHDN}</math></p> <p>0x04: Charger is in done mode. CHG_OK = 0 and <math>V_{BATT} &gt; V_{BATTREG} - V_{RSTRT}</math> and <math>T_J &lt; T_{SHDN}</math></p> <p>0x05: Reserved</p> <p>0x06: Charger is in timer-fault mode. CHG_OK = 0 and if BAT_DTLS = 0b001 then <math>V_{BATT} &lt; V_{PQLB}</math> or <math>V_{BATT} &lt; V_{PQDB}</math> and <math>T_J &lt; T_{SHDN}</math></p> <p>0x07: Charger is suspended because QBATT is disabled (DISQBAT = H or DISIBS = 1). CHG_OK = 0</p> <p>0x08: Charger is off, charger input invalid and/or charger is disabled. CHG_OK = 1</p> <p>0x09: Reserved</p> <p>0x0A: Charger is off and the junction temperature is <math>&gt; T_{SHDN}</math>. CHG_OK = 0</p> <p>0x0B: Charger is off because the watchdog timer expired. CHG_OK = 0</p> <p>0x0C: Charger is suspended or charge current or voltage is reduced based on JEITA control. This condition is also reported in THM_DTLS. CHG_OK = 0</p> <p>0x0D: Charger is suspended because battery removal is detected on THM pin. This condition is also reported in THM_DTLS. CHG_OK = 0</p> <p>0x0E: Charger is suspended because SUSPEND pin is high. CHG_OK = 0</p> <p>0x0F: Reserved</p>

### CHG\_DETAILS\_02 (0x15)

BIT	7	6	5	4	3	2	1	0
Field	RSVD_7		THM_DTLS[2:0]			BYP_DTLS[3:0]		
Reset								
Access Type	Read Only		Read Only			Read Only		

BITFIELD	BITS	DESCRIPTION	DECODE
RSVD_7	7		
THM_DTLS	6:4	Thermistor Details	0b000: Low temperature and charging suspended (COLD). 0b001: Low temperature charging (COOL). 0b010: Normal temperature charging (NORMAL). 0b011: High temperature charging (WARM). 0b100: High temperature and charging suspended (HOT). 0b101: Battery removal detected on THM pin. 0b110: Thermistor monitoring is disabled. 0b111: RSVD
BYP_DTLS	3:0	Bypass Node Details	0x0: The bypass node is okay. 0x1: OTG_ILIM when CHG_CNFG_00.MODE = 0xA or 0xE or 0xF The BYP to CHGIN switch (OTG switch) current limit was reached within the last 37.5ms. BYP_DTLS[0] status bit is latched until CHG_DETAILS_02 register read access is performed by AP. 0x2: BSTILIM The BYP reverse boost converter has hit its current limit and condition persisted for 30ms. 0x4: BCKNegILIM The BYP buck converter has hit the max negative demand current limit. BYP_DTLS[2] status bit is latched until CHG_DETAILS_02 register read access is performed by AP. 0x8: BST_SWON_DONE (This status bit is only available in CHG_CNFG_00.MODE = 0x9) The BYP reverse boost converter switch-on is done and V <sub>BYP</sub> reached the V <sub>BYPSET</sub> target.

### CHG\_CNFG\_00 (0x16)

Charger configuration 0

BIT	7	6	5	4	3	2	1	0	
<b>Field</b>	SS_ENV[1:0]		SS_PAT	BATRMV_MSK	MODE[3:0]				
<b>Reset</b>	0x0		0b0	0b0	0x4				
<b>Access Type</b>	Write, Read		Write, Read	Write, Read	Write, Read				

BITFIELD	BITS	DESCRIPTION	DECODE
SS_ENV	7:6	Spread Spectrum Control Register Bits	0x0: Disabled 0x1: 4% modulation envelope 0x2: 8% modulation envelope 0x3: 16% modulation envelope
SS_PAT	5	Spread Spectrum Pattern Type	0b0: Pseudo-random 0b1: Linear
BATRMV_MSK	4	Battery Removal Detection Masking When masked, battery removal detection is ignored.	0b0: Unmasked 0b1: Masked

BITFIELD	BITS	DESCRIPTION	DECODE
MODE	3:0	Smart Power Selector Configuration	<p>0x0: Charger = off, OTG = off, buck = off, boost = off. The Q<sub>BATT</sub> switch is on to allow the battery to support the system. BYP may or may not be biased based on the CHGIN availability.</p> <p>0x1: Same as 0x0</p> <p>0x2: Same as 0x0</p> <p>0x3: Same as 0x0</p> <p>0x4: Charger = off, OTG = off, buck = on, boost = off. When there is a valid input, the buck converter regulates the system voltage to be the maximum of (V<sub>MINSYS</sub> and V<sub>SYSREG_TRK</sub>) if SYS_TRACK_DIS = 0b0. V<sub>BYP</sub> is equal to V<sub>CHGIN</sub> minus the resistive drops.</p> <p>0x5: Charger = on, OTG = off, buck = on, boost = off. When there is a valid input, the battery is charging. V<sub>SYS</sub> is the larger of V<sub>SYSMIN</sub> and ~V<sub>BATT</sub> + I<sub>BATT</sub> × R<sub>BAT2SYS</sub>. V<sub>BYP</sub> is equal to V<sub>CHGIN</sub> minus the resistive drops.</p> <p>0x6: Same as 0x4 but V<sub>SYS</sub> regulated to be the maximum of (V<sub>BCKSYS</sub> and V<sub>SYSREG_TRK</sub>).if SYS_TRACK_DIS = 0b0</p> <p>0x7: Same as 0x4</p> <p>0x8: Reserved</p> <p>0x9: Charger = off, OTG = off, buck = off, boost = on.</p> <p>The Q<sub>BATT</sub> switch is on to allow the battery to support the system, the charger's DC-DC operates as a boost converter.</p> <p>BYP voltage is regulated to V<sub>BYPSET</sub>. Q<sub>CHGIN</sub> is off.</p> <p>0xA: Charger = off, OTG = on, buck = off, boost = on. The Q<sub>BATT</sub> switch is on to allow the battery to support the system, the charger's DC-DC operates as a boost converter.</p> <p>BYP voltage is regulated to V<sub>BYPSET</sub>. Q<sub>CHGIN</sub> is on allowing it to source current up to I<sub>CHGIN_OTG_LIM</sub>.</p> <p>0xB: Reserved</p> <p>0xC: Reserved</p> <p>0xD: Reserved</p> <p>0xE: Reserved</p> <p>0xF: Reserved</p>

### CHG\_CNFG\_01 (0x17)

Charger configuration 1

BIT	7	6	5	4	3	2	1	0
<b>Field</b>	TKEN	WDTEN	CHG_RSTRT[1:0]			SPR_3	FCHGTIME[2:0]	
<b>Reset</b>	0b1	0b0	0b01			0b0	0b011	
<b>Access Type</b>	Write, Read	Write, Read	Write, Read			Write, Read	Write, Read	

BITFIELD	BITS	DESCRIPTION	DECODE
TKEN	7	Trickle Charge Enable	0b0: Trickle charge is disabled: When $V_{BATT}$ is in trickle charge voltage range, charge current target level is $I_{FC}$ . 0b1: Trickle charge is enabled: When $V_{BATT}$ is in trickle charge voltage range, charge current target level is $I_{TRICKLE}$ .
WDTEN	6	Watchdog Timer Enable Bit	0b0: Watchdog timer disabled. 0b1: Watchdog timer enabled.
CHG_RSTR_T	5:4	Charger-Restart Threshold	0b00: 100mV below the value programmed by CHG_CV_PRM. 0b01: 150mV below the value programmed by CHG_CV_PRM. 10: 200mV below the value programmed by CHG_CV_PRM. 11: Disabled
SPR_3	3	Spare Bit	
FCHGTIME	2:0	Fast-Charge Timer Setting ( $t_{FC}$ , hrs)	0b000: Disable 0b001: 3 0b010: 4 0b011: 5 0b100: 6 0b101: 7 0b110: 8 0b111: Reserved

### CHG CNFG 02 (0x18)

#### Charger configuration 2

BIT	7	6	5	4	3	2	1	0
<b>Field</b>	SPR_7	CHG_CC[6:0]						
<b>Reset</b>	0b0	0x09						
<b>Access Type</b>	Write, Read	Write, Read						
BITFIELD	BITS	DESCRIPTION			DECODE			
SPR_7	7	Spare Bit						

BITFIELD	BITS	DESCRIPTION	DECODE
CHG_CC	6:0	<p>Fast-Charge Current Selection (mA). When the charger is enabled, the charge current limit is set by these bits. These bits range from 0.10A (0x00) to 5.5A (0x6E) in 50mA step.</p> <p>Note that the first three codes are all 100mA. Note that the thermal-foldback loop can reduce the battery charger's target current by ATJREG.</p> <p>Note that the fast-charge current is clamped at 3.5A from 0x46 to 0x7F in MAX77985.</p>	Value: Decode 0x00: 100 0x01: 100 0x02: 100 0x03: 150 0x04: 200 0x05: 250 0x06: 300 0x07: 350 0x08: 400 0x09: 450 0x0A: 500 0x0B: 550 0x0C: 600 0x0D: 650 0x0E: 700 0x0F: 750 0x10: 800 0x11: 850 0x12: 900 0x13: 950 0x14: 1000 0x15: 1050 0x16: 1100 0x17: 1150 0x18: 1200 0x19: 1250 0x1A: 1300 0x1B: 1350 0x1C: 1400 0x1D: 1450 0x1E: 1500 0x1F: 1550 0x20: 1600 0x21: 1650 0x22: 1700 0x23: 1750 0x24: 1800 0x25: 1850 0x26: 1900 0x27: 1950 0x28: 2000 0x29: 2050 0x2A: 2100 0x2B: 2150 0x2C: 2200 0x2D: 2250 0x2E: 2300 0x2F: 2350 0x30: 2400 0x31: 2450 0x32: 2500 0x33: 2550 0x34: 2600 0x35: 2650 0x36: 2700 0x37: 2750 0x38: 2800 0x39: 2850

BITFIELD	BITS	DESCRIPTION	DECODE
			0x3A: 2900 0x3B: 2950 0x3C: 3000 0x3D: 3050 0x3E: 3100 0x3F: 3150 0x40: 3200 0x41: 3250 0x42: 3300 0x43: 3350 0x44: 3400 0x45: 3450 0x46: 3500 0x47: 3550 0x48: 3600 0x49: 3650 0x4A: 3700 0x4B: 3750 0x4C: 3800 0x4D: 3850 0x4E: 3900 0x4F: 3950 0x50: 4000 0x51: 4050 0x52: 4100 0x53: 4150 0x54: 4200 0x55: 4250 0x56: 4300 0x57: 4350 0x58: 4400 0x59: 4450 0x5A: 4500 0x5B: 4550 0x5C: 4600 0x5D: 4650 0x5E: 4700 0x5F: 4750 0x60: 4800 0x61: 4850 0x62: 4900 0x63: 4950 0x64: 5000 0x65: 5050 0x66: 5100 0x67: 5150 0x68: 5200 0x69: 5250 0x6A: 5300 0x6B: 5350 0x6C: 5400 0x6D: 5450 0x6E: 5500 0x6F: 5500 0x70: 5500 0x71: 5500 0x72: 5500 0x73: 5500 0x74: 5500

BITFIELD	BITS	DESCRIPTION	DECODE
			0x75: 5500 0x76: 5500 0x77: 5500 0x78: 5500 0x79: 5500 0x7A: 5500 0x7B: 5500 0x7C: 5500 0x7D: 5500 0x7E: 5500 0x7F: 5500

[CHG\\_CNFG\\_03 \(0x19\)](#)

Charger configuration 3

BIT	7	6	5	4	3	2	1	0
<b>Field</b>	SPR_7	TO_TIME[2:0]					TO_I TH[3:0]	
<b>Reset</b>	0b0	0b011					0b0010	
<b>Access Type</b>	Write, Read	Write, Read					Write, Read	

BITFIELD	BITS	DESCRIPTION	DECODE
SPR_7	7	Spare Bit	
TO_TIME	6:4	Top-Off Timer Setting (min)	0b000: 30sec 0b001: 10 0b010: 20 0b011: 30 0b100: 40 0b101: 50 0b110: 60 0b111: 70
TO_I TH	3:0	Top-Off Current Threshold (mA). The charger transitions from its fast-charge constant voltage mode to its top-off mode when the charger current decays to the value programmed by this register. This transition generates a CHG_I interrupt and causes the CHG_DTLS register to report top-off mode. This transition also starts the top-off time as programmed by TO_TIME.	0b0000: Disable 0b0001: 150mA 0b0010: 200mA 0b0011: 250mA 0b0100: 300mA 0b0101: 350mA 0b0110: 400mA 0b0111: 450mA 0b1000: 500mA 0b1001: 550mA 0b1010: 600mA 0b1011: 650mA 0b1100: 700mA 0b1101: 750mA 0b1110: 800mA 0b1111: 850mA

[CHG\\_CNFG\\_04 \(0x1A\)](#)

Charger configuration 4

BIT	7	6	5	4	3	2	1	0	
<b>Field</b>	SYS_TRAC_K_DIS	MINSYS[1:0]			CHG_CV_PRM[4:0]				
<b>Reset</b>	0b0	0b01			0x04				
<b>Access Type</b>	Write, Read	Write, Read			Write, Read				
BITFIELD	BITS	DESCRIPTION			DECODE				
SYS_TRACK_DIS	7	BUCK SYS Tracking Disable Control			0x0: SYS tracking is enabled. In Buck mode, SYS is regulated to MAX of (VBATT +4%, VMINSYS). This is also valid in charge Done state. 0x1: SYS tracking is disabled. In Buck mode, SYS is regulated to VBATTERM.				
MINSYS	6:5	Minimum System Regulation Voltage			VALUE	MAX77985A/ MAX77986A	MAX77985B/ MAX77986B		
					0b00	3.4V	3.0V		
					0b01	3.5V	3.1V		
					0b10	3.6V	3.5V		
					0b11	3.7V	3.6V		
CHG_CV_PRM	4:0	Charge Termination Voltage Setting			VALUE	MAX77985A/ MAX77986A	MAX77985B/ MAX77986B		
					0x00	4.1500	3.5000		
					0x01	4.1625	3.5250		
					0x02	4.1750	3.5500		
					0x03	4.1875	3.5750		
					0x04	4.2000	3.6000		
					0x05	4.2125	3.6250		
					0x06	4.2250	3.6500		
					0x07	4.2375	3.6750		
					0x08	4.2500	3.7000		
					0x09	4.2625	3.7250		
					0x0A	4.2750	3.7500		
					0x0B	4.2875	3.7750		
					0x0C	4.3000	3.8000		
					0x0D	4.3125	3.8250		
					0x0E	4.3250	3.8500		
					0x0F	4.3375	3.8750		
					0x10	4.3500	3.9000		
					0x11	4.3625	3.9250		
					0x12	4.3750	3.9500		
					0x13	4.3875	3.9750		
					0x14	4.4000	4.0000		
					0x15	4.4125	4.0250		
					0x16	4.4250	4.0500		
					0x17	4.4375	4.0750		
					0x18	4.4500	4.1000		
					0x19	4.4625	4.1250		
					0x1A	4.4700	4.1500		
					0x1B	4.4875	4.1750		
					0x1C	4.5000	4.2000		
					0x1D	4.5125	4.2250		
					0x1E	4.5250	4.2500		
					0x1F	4.5375	4.2750		

[CHG\\_CNFG\\_05 \(0x1B\)](#)

Charger configuration 5

BIT	7	6	5	4	3	2	1	0	
<b>Field</b>	B2SOVRC_DTC	B2SOVRC_ALARM_ON_LY	B2SOVRC_CTRL	RECYCLE_EN	B2SOVRC[3:0]				
<b>Reset</b>	0b0	0b0	0b0	0b0	0x6				
<b>Access Type</b>	Write, Read	Write, Read	Write, Read	Write, Read	Write, Read				
BITFIELD	BITS	DESCRIPTION				DECODE			
B2SOVRC_DTC	7	BATT to SYS Overcurrent Debounce to Q <sub>BATT</sub> Clear Control				0x0: 105µs 0x1: 10ms			
B2SOVRC_A_LARM_ONLY	6	B2SOVRC Alarm Only Control				0x0: Alarm only is disabled: when tripping B2SOVRC, I/T is triggered and Q <sub>BATT</sub> opens after T <sub>OCP</sub> . 0x1: Alarm only is enabled: when tripping B2SOVRC, I/T is triggered but Q <sub>BATT</sub> remains closed even after T <sub>OCP</sub> .			
B2SOVRC_CTRL	5	Battery Mode B2SOVRC Monitoring Control				0x0: Automatic mode 0x1: Continuous mode			
RECYCLE_EN	4	B2S OCP or DISIBS Event Recycle Option				0b0: In the case of B2S OCP or DISIBS events, the buck is disabled (OFF) and Q <sub>BATT</sub> FET is opened. System recycles after 150ms (min) only if a valid charger is present. 0b1: In the case of B2S OCP or DISIBS events, buck is disabled (OFF) and Q <sub>BATT</sub> FET is opened. System recycles after 150ms (min).			
B2SOVRC	3:0	BATT to SYS Overcurrent Threshold (A)				0x0: Disabled 0x1: 3.0 0x2: 3.5 0x3: 4.0 0x4: 4.5 0x5: 5.0 0x6: 5.5 0x7: 6.0 0x8: 6.5 0x9: 7.0 0xA: 7.5 0xB: 8.0 0xC: 8.5 0xD: 9.0 0xE: 9.5 0xF: 10.0			

[CHG\\_CNFG\\_06 \(0x1C\)](#)

Charger configuration 6

BIT	7	6	5	4	3	2	1	0
<b>Field</b>	SPR_7_4[3:0]				CHGPROT[1:0]	WDTCLR[1:0]		
<b>Reset</b>	0x0				0b00	0b00		
<b>Access Type</b>	Write, Read				Write, Read	Write, Read		
BITFIELD	BITS	DESCRIPTION				DECODE		
SPR_7_4	7:4	Spare Bit						
CHGPROT	3:2	Charger Settings Protection Bit Writing "11" to these bits unlocks the write capability for the registers that are "Protected with CHGPROT." Writing any value besides "11" locks these registers.				0b00: Write capability locked. 0b01: Write capability locked. 0b10: Write capability locked. 0b11: Write capability unlocked.		
WDTCLR	1:0	Watchdog Timer Clear Bit. Writing "01" to these bits clears the watchdog timer when the watchdog timer is enabled.				0b00: The watchdog timer is not cleared. 0b01: The watchdog timer is cleared. 0b10: The watchdog timer is not cleared. 0b11: The watchdog timer is not cleared.		

**CHG\_CNFG\_07 (0x1D)**

Charger configuration 7

BIT	7	6	5	4	3	2	1	0
<b>Field</b>	WD_QBAT OFF	SPR_6	DISIBS	SPSN_DET_EN	QBEXT_CTR_L_EN	SPR_2_1[1:0]		FSHIP_MODE
<b>Reset</b>	0b0	0b0	0b0	0b0	0b0	0b00		0b0
<b>Access Type</b>	Write, Read	Write, Read	Write, Read	Write, Read	Write, Read	Write, Read		Write, Read
BITFIELD	BITS	DESCRIPTION				DECODE		
WD_QBATO FF	7	QBATT FET Control Under Watchdog Condition				0b0: When watchdog timer expires, turn off only the charger. 0b1: When watchdog timer expires, turn off buck, charger, and QBATT switch for 150ms.		
SPR_6	6	Spare Bit						
DISIBS	5	BATT to SYS FET Disable Control				0b0: BATT to SYS FET is controlled by the power-path state machine. 0b1: BATT to SYS FET is forced off.		
SPSN_DET_EN	4	SPSN Remote Sense Line Detection Enable. Enable SPSN remote sense line detection only when MODE = 0x0 ( <i>detection is discarded if not</i> ). End of SPSN detection triggers a BAT_I interrupt. Detection result available in dedicated status bit field SPSN_DTLS[1:0].				0b0: SPSN remote sense line detection disabled. 0b1: SPSN remote sense line detection enabled.		
QBEXT_CTR_L_EN	3					0b0: External QBATT control is disabled. 0b1: External QBATT control is enabled.		
SPR_2_1	2:1	Spare Bit						
FSHIP_MODE	0	Factory-Ship Mode. When asserted to "1", system enters into factory-ship mode. This bit can be reset by battery removal or on a valid charger input plug.				0b0: Not factory-ship mode. 0b1: Factory-ship mode.		

[CHG\\_CNFG\\_08 \(0x1E\)](#)

Charger configuration 8

BIT	7	6	5	4	3	2	1	0
<b>Field</b>	AUDIO_MO DE		BCKSYS[2:0]		FMBST	SLOWLX	FSW	DISKIP
<b>Reset</b>	0b0		0b000		0b0	0b0	0b1	0b0
<b>Access Type</b>	Write, Read		Write, Read		Write, Read	Write, Read	Write, Read	Write, Read

BITFIELD	BITS	DESCRIPTION	DECODE		
AUDIO_MOD_E	7	Audio Mode Control Bit	0x0: No audio mode enabled. 0x1: Audio mode enabled.		
BCKSYS	6:4	BCKSYS Settings	VALUE	MAX77985A/ MAX77986A	MAX77985B/ MAX77986B
			0b000	4.65	4.5
			0b001	4.70	4.6
			0b010	4.75	4.7
			0b011	4.80	4.8
			0b100	4.85	4.9
			0b101	4.90	5.0
			0b110	4.95	5.0
			0b111	5.00	5.0
			<b>Note:</b> Do not set BCKSYS > CHGIN -0.3V.		
FMBST	3	Factory Mode Boost	0b0: When DISQBAT = high, any mode change is not possible. 0b1: When DISQBAT = high, this bit makes mode change (Boost mode) possible.		
SLOWLX	2	LX Slope Control Options	0b0: Fastest LX slope without control. 0b1: Slowest LX slope.		
FSW	1	Switching Frequency Options (MHz)	0b0: 2.6 0b1: 1.3		
DISKIP	0	Charger Skip Mode Disable	0b0: Auto skip mode. 0b1: Disable skip mode.		

[CHG\\_CNFG\\_09 \(0x1F\)](#)

Charger configuration 9

BIT	7	6	5	4	3	2	1	0	
<b>Field</b>	SPR_7	CHGIN_ILIM[6:0]							
<b>Reset</b>	0b0	0x09							
<b>Access Type</b>	Write, Read	Write, Read							
BITFIELD	BITS	DESCRIPTION			DECODE				
SPR_7	7	Spare Bit							

BITFIELD	BITS	DESCRIPTION	DECODE
CHGIN_ILIM	6:0	<p>CHGIN Input Current Limit (mA)</p> <p>7 Bit adjustment :</p> <p>from 0.1A to 3.5A in 50mA steps (MAX77985A/B)</p> <p>from 0.1A to 5.5A in 50mA steps (MAX77986A/B)</p> <p>Note that the first two codes are both 0.1A.</p>	0x00: 100 0x01: 100 0x02: 150 0x03: 200 0x04: 250 0x05: 300 0x06: 350 0x07: 400 0x08: 450 0x09: 500 0x0A: 550 0x0B: 600 0x0C: 650 0x0D: 700 0x0E: 750 0x0F: 800 0x10: 850 0x11: 900 0x12: 950 0x13: 1000 0x14: 1050 0x15: 1100 0x16: 1150 0x17: 1200 0x18: 1250 0x19: 1300 0x1A: 1350 0x1B: 1400 0x1C: 1450 0x1D: 1500 0x1E: 1550 0x1F: 1600 0x20: 1650 0x21: 1700 0x22: 1750 0x23: 1800 0x24: 1850 0x25: 1900 0x26: 1950 0x27: 2000 0x28: 2050 0x29: 2100 0x2A: 2150 0x2B: 2200 0x2C: 2250 0x2D: 2300 0x2E: 2350 0x2F: 2400 0x30: 2450 0x31: 2500 0x32: 2550 0x33: 2600 0x34: 2650 0x35: 2700 0x36: 2750 0x37: 2800 0x38: 2850 0x39: 2900 0x3A: 2950

BITFIELD	BITS	DESCRIPTION	DECODE
			0x3B: 3000 0x3C: 3050 0x3D: 3100 0x3E: 3150 0x3F: 3200 0x40: 3250 0x41: 3300 0x42: 3350 0x43: 3400 0x44: 3450 0x45: 3500 0x46: 3550 0x47: 3600 0x48: 3650 0x49: 3700 0x4A: 3750 0x4B: 3800 0x4C: 3850 0x4D: 3900 0x4E: 3950 0x4F: 4000 0x50: 4050 0x51: 4100 0x52: 4150 0x53: 4200 0x54: 4250 0x55: 4300 0x56: 4350 0x57: 4400 0x58: 4450 0x59: 4500 0x5A: 4550 0x5B: 4600 0x5C: 4650 0x5D: 4700 0x5E: 4750 0x5F: 4800 0x60: 4850 0x61: 4900 0x62: 4950 0x63: 5000 0x64: 5050 0x65: 5100 0x66: 5150 0x67: 5200 0x68: 5250 0x69: 5300 0x6A: 5350 0x6B: 5400 0x6C: 5450 0x6D: 5500

[CHG CNFG 10 \(0x20\)](#)

Charger configuration 10

BIT	7	6	5	4	3	2	1	0	
<b>Field</b>	OTG_REC_EN	INLIM_CLK[1:0]			OTG_ILIM[4:0]				
<b>Reset</b>	0b0	0b10			0x00				
<b>Access Type</b>	Write, Read	Write, Read			Write, Read				
BITFIELD	BITS	DESCRIPTION				DECODE			
OTG_REC_EN	7	OTG OCP Event Recycle Option				1b0: In case of OTG OCP, OTG FET is disabled (OFF = opened). System does not recycle OTG output. 1b1: In case of OTG OCP, OTG FET is disabled (OFF = opened). OTG recycles after $T_{OTG}$ retry.			
INLIM_CLK	6:5	Input Current Limit Soft-Start Clock ( $\mu$ sec per 50mA step)				0b00: 16 0b01: 512 0b10: 2048 0b11: 8192			
OTG_ILIM	4:0	CHGIN OTG Output Current Limit (mA). When the boost-OTG mode (MODE = 0xA) is enabled, the OTG output current limit is set by these bits. These bits range from 0.50A (0x00) to 3.1A (0x1A) in 100mA steps. Note that the OTG output current limit is clamped at 2.4A from 0x13 to 0x1F in MAX77985.				Value: Decode 0x00: 500 0x01: 600 0x02: 700 0x03: 800 0x04: 900 0x05: 1000 0x06: 1100 0x07: 1200 0x08: 1300 0x09: 1400 0x0A: 1500 0x0B: 1600 0x0C: 1700 0x0D: 1800 0x0E: 1900 0x0F: 2000 0x10: 2100 0x11: 2200 0x12: 2300 0x13: 2400 0x14: 2500 0x15: 2600 0x16: 2700 0x17: 2800 0x18: 2900 0x19: 3000 0x1A: 3100 0x1B: 3100 0x1C: 3100 0x1D: 3100 0x1E: 3100 0x1F: 3100			

[CHG\\_CNFG\\_11 \(0x21\)](#)

Charger configuration 11

BIT	7	6	5	4	3	2	1	0
<b>Field</b>	SPR_7	VBYPSET[6:0]						
<b>Reset</b>	0b0	0x1						
<b>Access Type</b>	Write, Read	Write, Read						
BITFIELD	BITS	DESCRIPTION						
SPR_7	7	Spare Bit						
VBYPSET	6:0	VBYP Target Output Voltage (V). Bypass target output voltage in boost mode. MODE = 0x9/0xA. 5.0V to 12.0V with 100mV step.						

### CHG CNFG\_12 (0x22)

Charger configuration 12

BIT	7	6	5	4	3	2	1	0
<b>Field</b>	BYPDISCH G_EN	DEEP_SUS P_DIS	VCHGIN_REG[1:0]	UNPLUG_TH[1:0]	UNPLUG_D B	DIS_AICL		
<b>Reset</b>	0b0	0b0	0b01	0b01	0b0	0b0		
<b>Access Type</b>	Write, Read	Write, Read	Write, Read	Write, Read	Write, Read	Write, Read	Write, Read	
BITFIELD	BITS	DESCRIPTION			DECODE			
BYPDISCHG _EN	7	Boost BYP Discharge after Overshoot. When enabled, if BYP is seen to be above target, a soft pulldown is activated to discharge BYP back to target, even if auto-skip mode is active.			0b0: Disabled 0b1: Enabled			
DEEP_SUSP _DIS	6	When SUSPND pin pulls high or in MODE 0, input FET is enabled or disabled by this bit.			0b0: Disabled 0b1: Enabled			
VCHGIN_REG	5:4	CHGIN Voltage Regulation Threshold ( $V_{CHGIN\_REG}$ ) Adjustment. The CHGIN to GND minimum turn-on threshold ( $V_{CHGIN\_UVLO}$ ) also scales with this adjustment.			0b00: $V_{CHGIN\_REG} = 4.5V$ and $V_{CHGIN\_UVLO} = 4.7V$ 0b01: $V_{CHGIN\_REG} = 4.6V$ and $V_{CHGIN\_UVLO} = 4.8V$ 0b10: $V_{CHGIN\_REG} = 4.7V$ and $V_{CHGIN\_UVLO} = 4.9V$ 0b11: $V_{CHGIN\_REG} = 4.85V$ and $V_{CHGIN\_UVLO} = 5.05V$			
UNPLUG_TH	3:2	VBUS Unplug Detection Enabling and Threshold. If written as 0b01, 0b10, or 0b11, QBEXT pin works as unplug detection active-low output (so called PGOOD)			0b00: Disabled (functions as QBEXT) 0b01: 5V unplug detection 0b10: 9V unplug detection 0b11: 15V unplug detection  (Note: For parts with chip revision 0b001 (PASS1), contact Analog Devices for Rev 0 of the data sheet. This information can be read from the CHIP_REVISION (0x01) register.)			
UNPLUG_D B	1	VBUS Unplug Detection Debounce Selection			0b0: No debounce 0b1: 2μs (only valid when SUSPEND = 0)			
DIS_AICL	0	AICL Disable Feature			0b0: AICL feature is not disabled. 0b1: AICL feature is disabled.			

[CHG\\_CNGF\\_13 \(0x23\)](#)

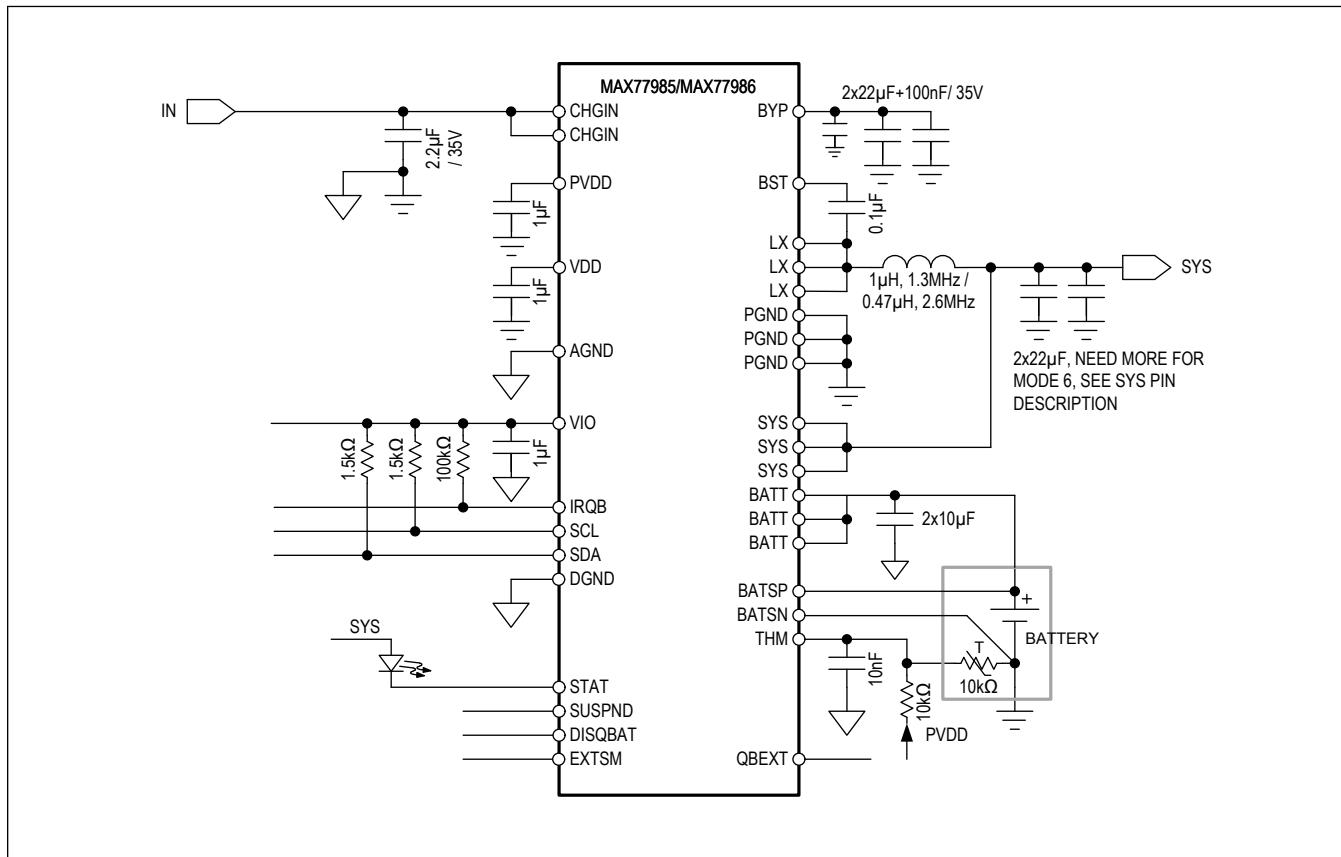
BIT	7	6	5	4	3	2	1	0	
<b>Field</b>	JEITA_EN	SPR_6	CHG_CV_C_OOL	CHG_CC_WARM	REGTEMP[3:0]				
<b>Reset</b>	0b0	0b0	0b0	0b0	0x6				
<b>Access Type</b>	Write, Read	Write, Read	Write, Read	Write, Read	Write, Read				
BITFIELD	BITS	DESCRIPTION				DECODE			
JEITA_EN	7	JEITA Enable				0x0: JEITA disabled Fast-charge current and charge termination voltage do not change based on thermistor temperature. 0x1: JEITA enabled Fast-charge current and charge termination voltage change based on thermistor temperature.			
SPR_6	6	Spare Bit							
CHG_CV_C_OOL	5	JEITA controlled battery termination voltage when thermistor temperature is between $T_{COLD}$ and $T_{COOL}$ .				0x0: Battery termination voltage is set by CHG_CV_PRM. 0x1: Battery termination voltage is set by (CHG_CV_PRM - 187.5mV) [MAX77985A/MAX77986A] by (CHG_CV_PRM - 375mV) [MAX77985B/MAX77986B]			
CHG_CC_WARM	4	JEITA controlled battery fast-charge current when thermistor temperature is between $T_{WARM}$ and $T_{HOT}$ .				0x0: Battery fast-charge current is set by CHG_CC. 0x1: Battery fast-charge current is to 50% of CHG_CC.			
REGTEMP	3:0	Junction Temperature Thermal Regulation (°C). The charger's target current limit starts to foldback and the TREG bit is set if the junction temperature is greater than the REGTEMP setpoint.				0x0: 85 0x1: 90 0x2: 95 0x3: 100 0x4: 105 0x5: 110 0x6: 115 0x7: 120 0x8: 125 0x9: 130			

[STAT\\_CNGF \(0x24\)](#)

BIT	7	6	5	4	3	2	1	0
<b>Field</b>	STAT_EN	SPR_6_4[2:0]				STAT_CURR[1:0]	SPR_1	STAT_MOD_E
<b>Reset</b>	0b1	0b0				0x00	0b0	0b0
<b>Access Type</b>	Write, Read	Write, Read				Write, Read	Write, Read	Write, Read
BITFIELD	BITS	DESCRIPTION				DECODE		
STAT_EN	7	STAT Charging Status Indication LED Enable Bit				0x0: Disable 0x1: Enable		
SPR_6_4	6:4	Spare Bit						

BITFIELD	BITS	DESCRIPTION	DECODE
STAT_CURR	3:2	STAT LED Driving Current (mA)	0b00: 5 0b01: 10 0b10: 15 0b11: 20
SPR_1	1	Spare Bit	
STAT_MODE	0	STAT LED Behaviour Selection Bit	0b0: LED mode 1 0b1: LED mode 2

## Typical Application Circuits



### Layout Guidelines

1. Do not connect AGND/DGND directly with the top layer PGND. Allocate ground isolation between AGND/DGND and PGND. This can be done by connecting AGND/DGND to the system ground plane near the input capacitor. Usually, the input source's ground plane is more stable.
2. Place CHGIN, BYP, SYS, and BATT bypass capacitors as close as possible to the IC pins and connect them to the power ground plane on the PCB top layer. Especially, make sure the connection between BYP cap ground and SYS cap ground is as direct, short, and wide as possible. If possible, reinforce the connection between BYP cap ground and SYS cap ground in other layers.
3. Place PVDD and AVDD bypass cap as close to IC as possible. Reduce the loop area between the bypass cap and the IC.
4. Use wide and short traces for high current connections such as CHGIN, BYP, SYS, and BATT.
5. BATSP and BATSN should have direct Kelvin sensing connections to the interface of the battery. Connect as close to the battery terminal as possible, so that the battery voltage is sensed correctly.
6. The kelvin sensing traces should not be shared with other circuits and via on kelvin traces are not recommended.

**Ordering Information**

PART NUMBER	TEMP RANGE	MAX FAST-CHARGE CURRENT (A)	SUPPORTED BATTTERY TYPE	INDUCTOR CURRENT LIMIT (A)	REVERSE BOOST POWER CEILING (W)	PIN-PACKAGE	DEFAULT MODE[3:0]
MAX77985AEFD+	-40°C to +85°C	3.5	General Li-ion	8.3	12	32-FC2QFN	0x4
MAX77985AEFD+T	-40°C to +85°C	3.5	General Li-ion	8.3	12	32-FC2QFN	0x4
MAX77985BEFD+	-40°C to +85°C	3.5	LiFePO <sub>4</sub>	8.3	12	32-FC2QFN	0x4
MAX77985BEFD+T	-40°C to +85°C	3.5	LiFePO <sub>4</sub>	8.3	12	32-FC2QFN	0x4
MAX77986AEFD+	-40°C to +85°C	5.5	General Li-ion	11.1	18	32-FC2QFN	0x4
MAX77986AEFD+T	-40°C to +85°C	5.5	General Li-ion	11.1	18	32-FC2QFN	0x4
MAX77986BEFD+	-40°C to +85°C	5.5	LiFePO <sub>4</sub>	11.1	18	32-FC2QFN	0x4
MAX77986BEFD+T	-40°C to +85°C	5.5	LiFePO <sub>4</sub>	11.1	18	32-FC2QFN	0x4

+Denotes a lead(Pb)-free/RoHS-compliant package.

T = Tape and reel.

**Revision History**

REVISION NUMBER	REVISION DATE	DESCRIPTION	PAGES CHANGED
0	4/22	Release for Market Intro	—
1	10/23	Updated <i>Electrical Characteristics</i> table, <i>Pin Description</i> table, <i>Functional Diagram</i> , <i>Detailed Description</i> , <i>Switching Mode Charger</i> , <i>JEITA Controlled Charging</i> , <i>Unplug Detection</i> , <i>Table 8</i> , <i>Top-Off State</i> , <i>Main-Battery Overcurrent Protection Due to Fault</i> , <i>Battery to SYS QBATT Switch Control (DISIBS)</i> , <i>Register Details</i> table, <i>Typical Application Circuit</i> , <i>Layout Guidelines</i> , and <i>Ordering Information</i> table	10, 12–14, 16, 17, 25–28, 37, 41, 42, 44–48, 55, 64, 70, 73, 74, 77, 78, 80
2	11/23	Updated <i>General Description</i> , <i>Electrical Characteristics</i> table, <i>Main-Battery Overcurrent Protection Due to Fault</i> , <i>Battery to SYS QBATT Switch Control (DISIBS)</i> , <i>Unplug Detection</i> , and <i>Register Details</i>	1, 10, 12, 13, 15, 17, 22, 41, 42, 47, 55, 82
3	2/25	Updated <i>Register Details</i> section	56

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