

LTC3311-1

3.3V to 1V at 12.5A, 2MHz Automotive  
Low EMI Buck Regulator

## DESCRIPTION

Demonstration Circuit 3113A features the LTC®3311-1, fixed 1V, 12.5A synchronous step-down Silent Switcher® operating as a 2MHz, 3V to 3.6V input, 1V/12.5A output buck regulator. The LTC3311-1 supports fixed 1V output voltage with operating frequencies from 500kHz up to 5MHz. The LTC3311-1 is a compact, ultralow emission, high efficiency, and high speed synchronous monolithic step-down switching regulator. The LTC3311-1 has Active Voltage Positioning (AVP) where the output voltage is dependent on load current. At light load, the output voltage is regulated above the nominal value. At full load, the output voltage is regulated below the nominal value. The DC load regulation is adjusted to improve transient performance and reduce required output capacitance.

DC3113A is set up to run in forced continuous mode with a 2MHz switching frequency but can be configured to run at different switching frequencies, or in pulse-skipping mode. The LTC3311-1 oscillator can also synchronize to an external clock using MODE/SYNC turret, with the DC3113A default setup. Figure 7 shows the efficiency and power loss of the DC3113A with a 3.3V input in both operation modes.

The DC3113A is optimized for a 4A to 12A, 8A/μs load step transient. A fast load step transient circuit is placed on the back of the demo board to measure the load step response of the converter.

The DC3113A also has an EMI filter to reduce conducted EMI. This EMI filter can be included by applying the input voltage at the VIN EMI terminal. The EMI performance of the board is shown in the EMI TEST RESULTS section. The red lines in the EMI performance graphs illustrate the CISPR25 Class 5 peak limits for the conducted and radiated emission tests.

The LTC3311-1 data sheet gives a complete description of the part and its application information. The data sheet must be read in conjunction with this demo manual. The LTC3311-1 is assembled in a 3mm × 3mm LQFN package with exposed pads for low thermal resistance. The layout recommendations for low EMI operation and maximum thermal performance are available in the data sheet section Low EMI PCB Layout.

**Design files for this circuit board are available.**

All registered trademarks and trademarks are the property of their respective owners.

PERFORMANCE SUMMARY Specifications are at  $T_A = 25^\circ\text{C}$ 

SYMBOL	PARAMETER	CONDITIONS	MIN	TYP	MAX	UNITS
$V_{\text{IN}}/V_{\text{IN}}$ EMI	DC3113A Input Voltage Range		3.0		3.6	V
$V_{\text{OUT\_TYP}}$	DC3113A Typical Output Voltage	$I_{\text{OUT}} = 8\text{A}$	0.990	1.000	1.010	V
$V_{\text{OUT\_NO\_LOAD}}$	DC3113A Output Voltage with No Load	$I_{\text{OUT}} = 0\text{A}$	1.005	1.016	1.027	V
$V_{\text{OUT\_FULL\_LOAD}}$	DC3113A Output Voltage with Full Load	$I_{\text{OUT}} = 12.5\text{A}$	0.980	0.991	1.001	V
VAVP	$V_{\text{OUT}}$ Active Voltage Positioning			2		mV/A
$I_{\text{OUT}}$	DC3113A Output Current				12.5	A
$f_{\text{sw}}$	Switching Frequency		1.8	2	2.2	MHz
EFF	Efficiency	$V_{\text{IN}} = 3.3\text{V}$ , $I_{\text{OUT}} = 5\text{A}$	90			%

# DEMO MANUAL DC3113A

---

## BOARD PHOTO

Part marking is either ink mark or laser mark

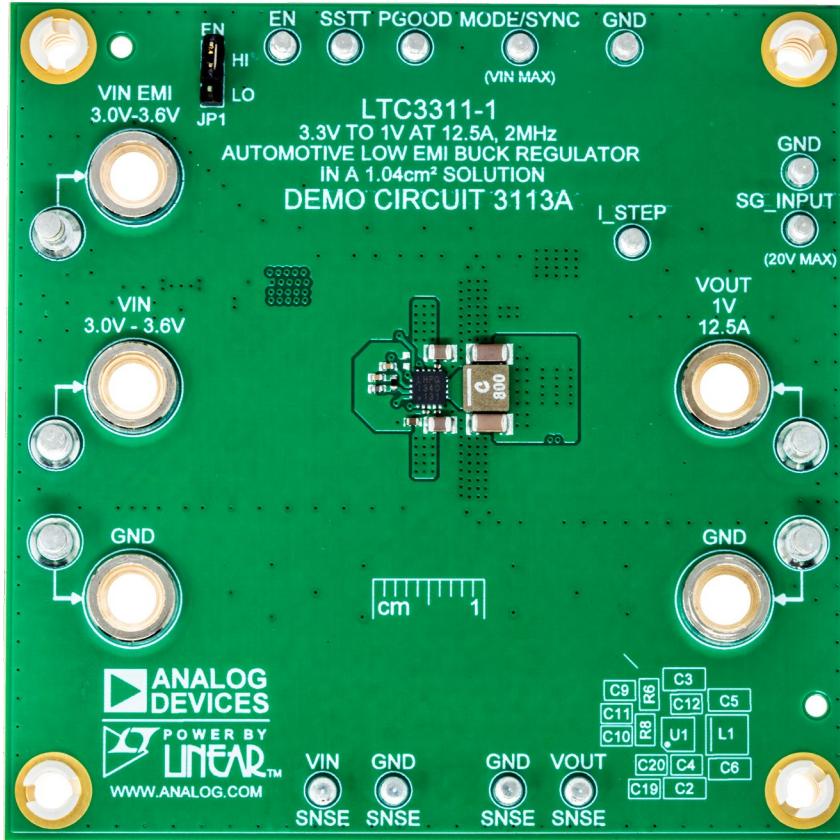


Figure 1. DC3113A Board Photo

## QUICK START PROCEDURE

Refer to Figure 2 for the proper measurement equipment setup and follow the procedure below:

NOTE: For accurate  $V_{IN}$ ,  $V_{OUT}$  and efficiency measurements, measure  $V_{IN}$  at the VIN SNSE and GND SNSN turrets, and measure  $V_{OUT}$  at the VOUT SNSE and GND SNSE turrets as illustrated as VM1 and VM2 in Figure 2. When measuring the input or output ripple, care must be taken to avoid a long ground lead on the oscilloscope probe.

1. Set the JP1 Jumper to the HI position.
2. With power off, connect the input power supply to VIN and GND. If the input EMI filter is desired, connect the input power supply to VIN EMI and GND.
3. Set power supply PS1 current limit to 10A. Set the electronic load LD1 to CC mode and 0A current. Slowly increase PS1 to 1V. If PS1 output current reads less than 20mA, increase PS1 to 3.3V. Verify that VM1

## QUICK START PROCEDURE

reads 3.3V and VM2 reads around 764mV. Check VM1, VM2, VM3, PS1 output current and LD1 input current. Connect an oscilloscope voltage probe as shown in Figure 3 or Figure 4. Set Channel to AC-coupled, voltage scale to 20mV and time base to 10μs. Check  $V_{OUT}$  ripple voltage. Verify that PGOOD voltage is above 3V.

4. Increase the load by 1A intervals up to 12.5A and observe the voltage output regulation, ripple voltage, and the voltage on the SSTT turret. Calculate Die temperature using the formula below:

$$T_J(\text{°C}) = \frac{V_{SSTT}}{4\text{mV}} - 273 \quad (1)$$

5. If pulse-skipping mode is desired, set PS1 to 0V. Install a 0Ω resistor in the R5 location and remove R3. Repeat steps 1 through 4. In step 4, observe that the switching waveform is now in pulse-skipping mode at light load.
6. Optional: To change the frequency, remove R4 and R5, if installed. Install the desired  $R_T$  resistor in the R6 location. Note that the MODE/SYNC pin should have high impedance to GND and VIN. Size the inductor, output capacitors and compensation components to provide the desired inductor ripple and a stable output. Refer to

the LTC3311-1 data sheet and LTPowerCAD® for more information on choosing the required components.

7. To test the transient response with a base load, add the desired resistor to produce a minimum load between  $V_{OUT}$  and  $I_{STEP}$  turrets (RL shown on Figure 2). Note that the total load resistance will be  $RL$  plus  $R10$  (10mΩ). Adjust a signal generator with a 10ms period, 10% duty cycle and an amplitude from 1V to 2V to start.
8. Measure the  $I_{STEP}$  voltage to observe the current,  $V_{I\_STEP}/10\text{m}\Omega$ . Adjust the amplitude of the pulse to provide the desired transient. Connect signal generator SG1 between SG\_INPUT and GND turrets. Adjust the rising and falling edge of the pulse to provide the desired ramp rate. Figure 8 shows a load step from 4A to 12A. Refer to the following equations:

$$I_{OUT} = \frac{V_{I\_STEP}}{10\text{m}\Omega} \quad (2)$$

$$V_{GS} = V_{SG\_INPUT} - V_{I\_STEP} \quad (3)$$

9. When done, turn off SG1, PS1 and Load. Remove all the connections to the demo board.

# DEMO MANUAL DC3113A

## TEST SETUP

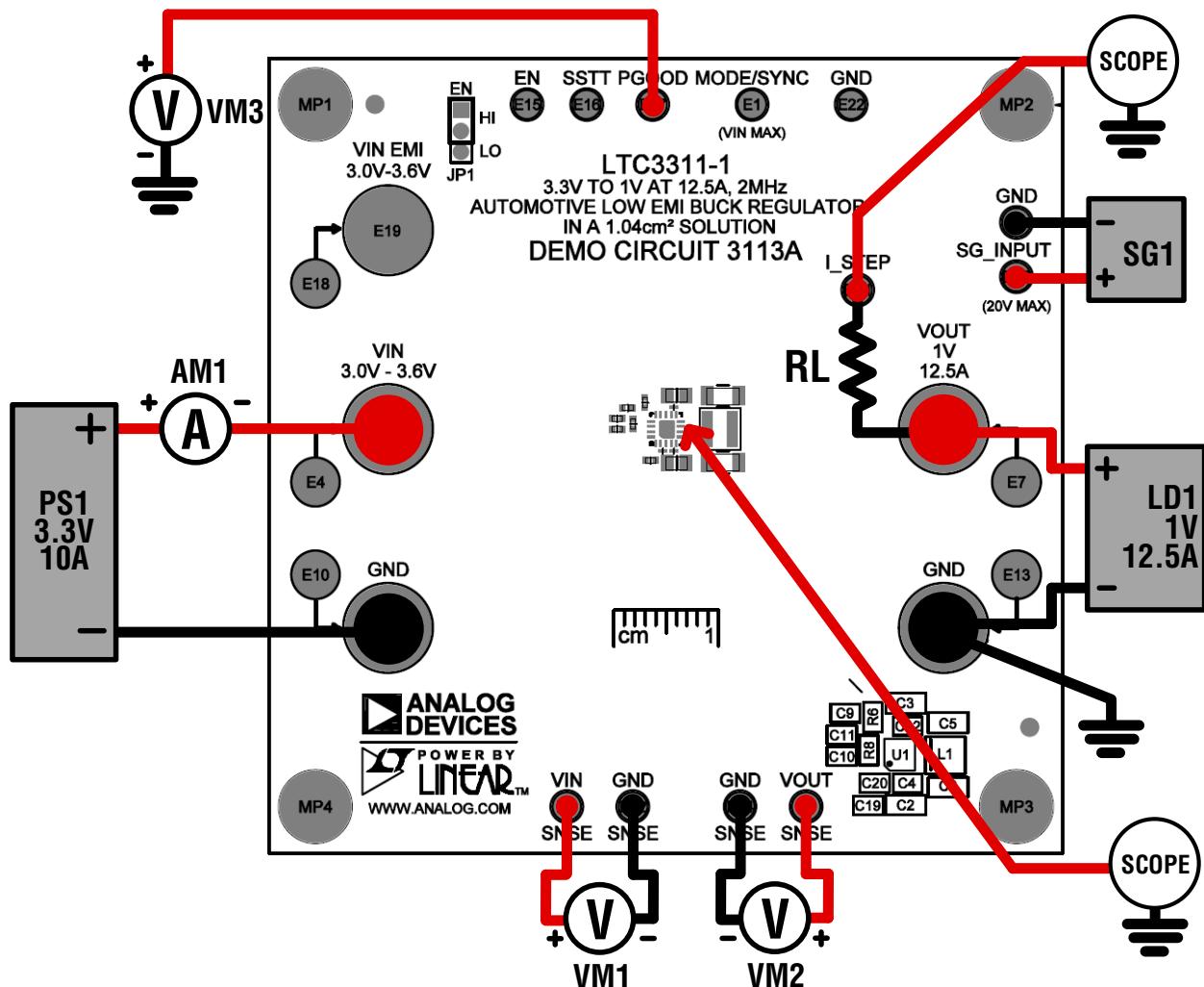


Figure 2. Test Setup for DC3113A Demo Board

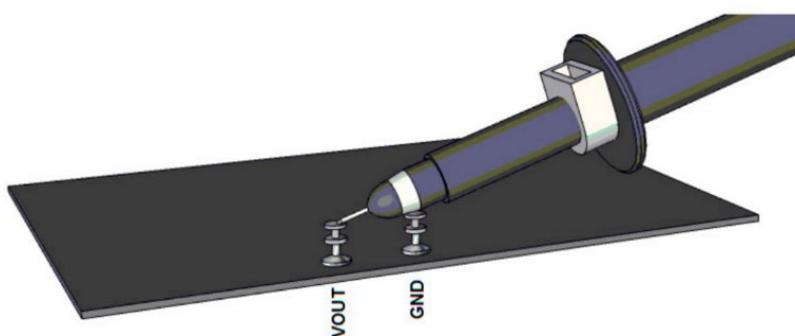


Figure 3. Technique for Measuring Output Ripple and Step Response with a Scope Probe



Figure 4. Technique for Measuring Output Ripple and Step Response with a Low Inductance Connector (Not Supplied)

## THEORY OF OPERATION

### Introduction to the DC3113A

The DC3113A demonstration circuit features the LTC3311-1, a low voltage synchronous step-down Silent Switcher. The LTC3311-1 is a monolithic, constant frequency, current mode step-down DC/DC converter. An oscillator turns on the internal top power switch at the beginning of each clock cycle. Current in the inductor then increases until the top switch comparator trips and turns off the top power switch. The peak inductor current, at which the top switch turns off, is controlled by the voltage on the ITH node. The error amplifier servos the ITH node by comparing the voltage on the internal VFB node with an internal 500mV reference. When the load current increases, it causes a reduction in the feedback voltage relative to the reference, leading the error amplifier to raise the ITH voltage until the average inductor current matches the new load current. When the top switch turns off, the synchronous bottom power switch turns on until the next clock cycle begins. In pulse-skipping mode, the bottom switch also turns off when inductor current falls to zero. If overload conditions result in excessive current flowing through the bottom switch, the next clock cycle will be delayed until the switch current returns to a safe level.

If the EN pin is low, the LTC3311-1 is in shutdown state with low quiescent current. When the EN pin is above its threshold, the switching regulator will be enabled.

The MODE/SYNC pin synchronizes the switching frequency to an external clock. It can be a clock output for multi-phase operation. It also sets the regulator operation modes. The operation modes are either forced continuous or pulse-skipping. See the LTC3311-1 data sheet for more detailed information.

The maximum allowable operating frequency is influenced by the minimum on time of the top switch, the ratio of  $V_{OUT}$  to  $V_{IN}$  and the inductance of the output inductor. The maximum allowable operating frequency may be calculated in the formula below.

$$f_{SW(MAX)} = \frac{V_{OUT}}{V_{IN(MAX)} \cdot t_{ON(MIN)}} \quad (4)$$

Select an operating switching frequency below  $f_{SW(MAX)}$ . It is desired to obtain an inductor current of 30% of the maximum LTC3311-1 operating load, 12.5A. Use the formulas below to calculate the inductor value to obtain a 30% (3A) inductor ripple for the operating frequency.

$$L \geq \frac{V_{OUT}}{3.75A \cdot f_{SW}} \cdot \left( 1 - \frac{V_{OUT}}{V_{IN(MAX)}} \right) \text{ for } \frac{V_{OUT}}{V_{IN(MAX)}} \leq 0.5 \quad (5)$$

When determining the compensation components, C10, C11 and R8, controlling the loop stability and transient response are the two main considerations. The LTC3311-1 has been designed to operate at a high bandwidth for fast transient response capabilities. This reduces required output capacitance to meet the desired transient voltage range. The mid-band gain of the loop increases with R8 and the bandwidth of the loop increases with decreasing C11. C10 along with R8 provides a high frequency pole to reduce the high frequency gain.

Loop stability is generally measured using the Bode Plot method of plotting loop gain in dB and phase shift in degrees. The 0dB crossover frequency should be less than the 1/6 of the operating frequency to reduce the effects of added phase shift of the modulator. The control loop phase margin goal should be 45° or greater and a gain margin goal of 8dB or greater.

# DEMO MANUAL DC3113A

## TYPICAL PERFORMANCE CHARACTERISTICS

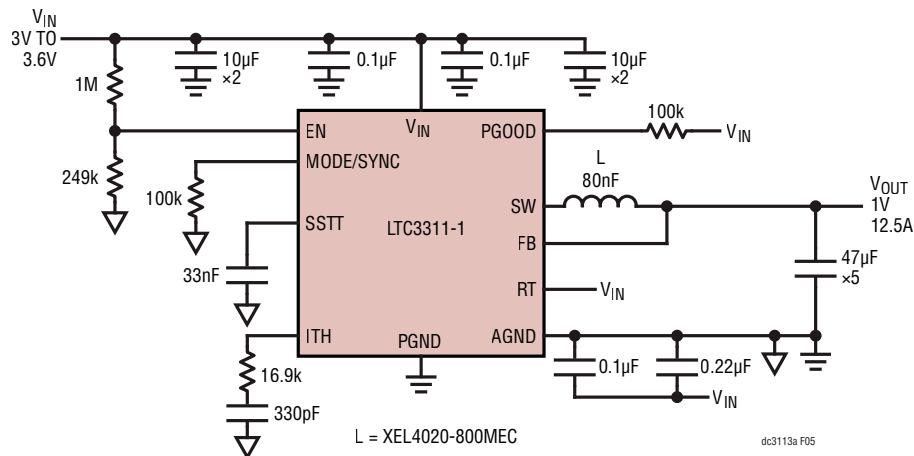


Figure 5. DC3113A, 1V/12.5A Step-Down Converter Typical Solution Schematic

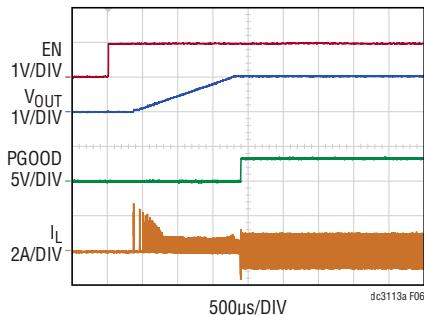


Figure 6. Start-Up Waveforms with Light Load

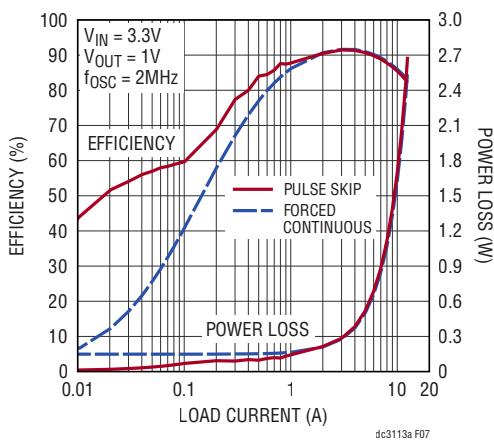


Figure 7. Efficiency vs Load Current

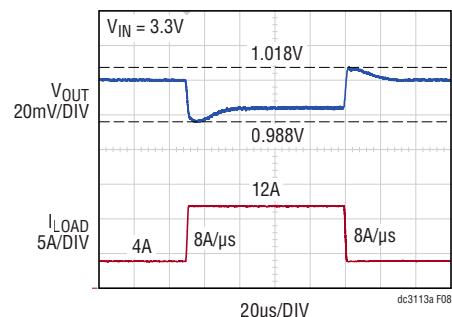
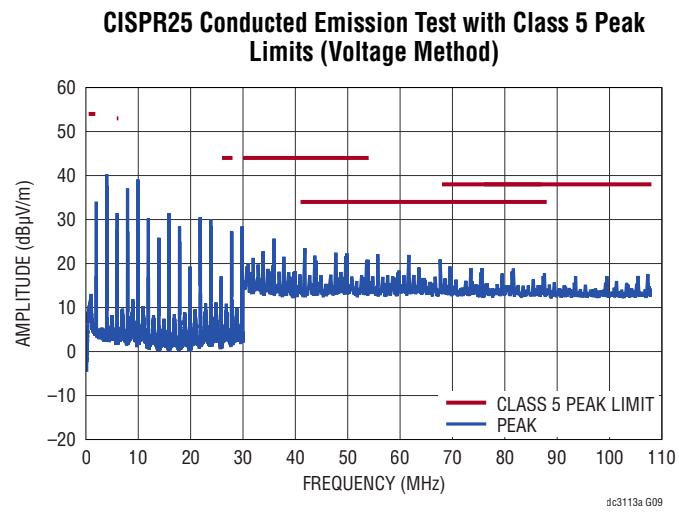
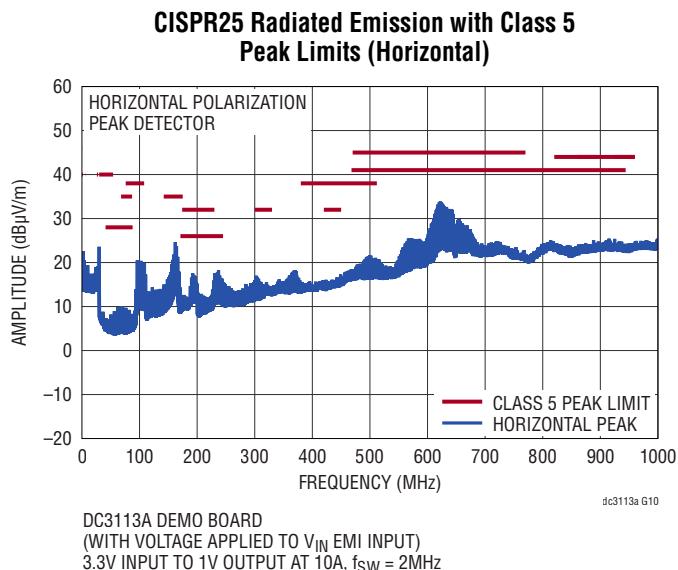


Figure 8. Load Step Response

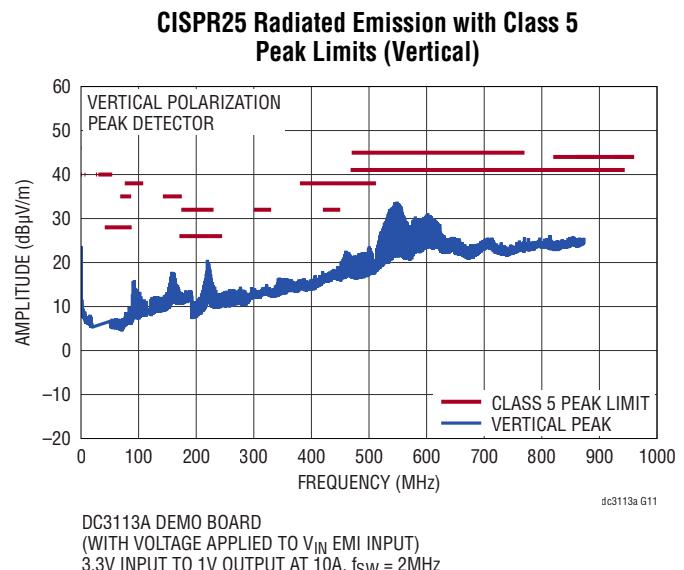
## EMI TEST RESULTS



**Figure 9. DC3113A CISPR25 Conducted Emission with Class 5 Peak Limits (Voltage Method)**



**Figure 10. DC3113A CISPR25 Radiated Emission (Horizontal)**



**Figure 11. DC3113A CISPR25 Radiated Emission (Vertical)**

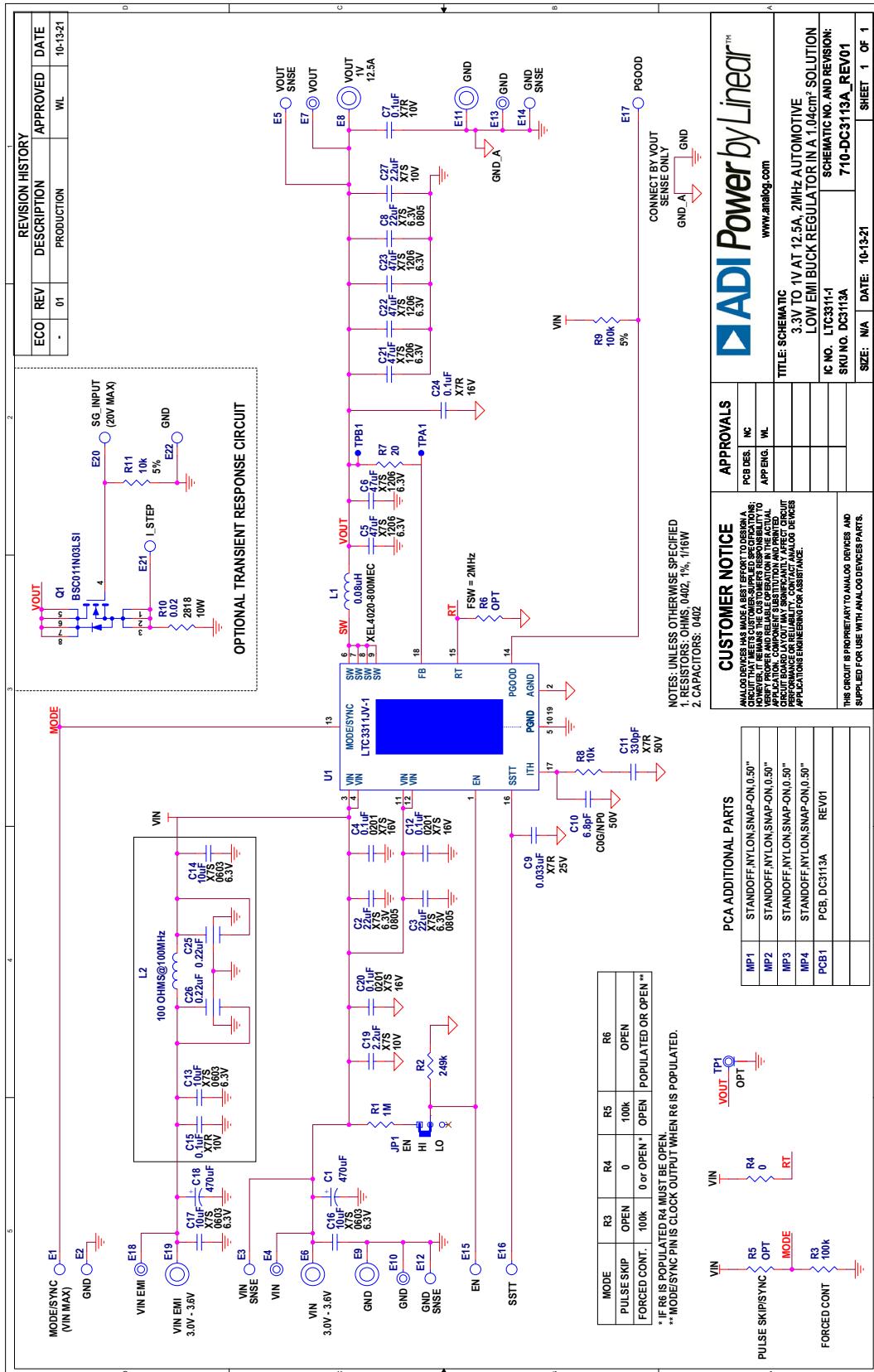
# DEMO MANUAL DC3113A

## PARTS LIST

ITEM	QTY	REFERENCE	PART DESCRIPTION	MANUFACTURER/PART NUMBER
<b>Required Circuit Components</b>				
1	2	C2, C3	CAP, 22µF, X7S, 6.3V, 20%, 0805	TDK, C2012X7S0J226M125AC
2	3	C4, C12, C20	CAP, 0.1µF, X7S, 16V, 10%, 0201	MURATA, GRM033C71C104KE14D
3	5	C5, C6, C21-C23	CAP, 47µF, X7S, 6.3V, 20%, 1206	TDK, C3216X7S0J476M160AC
4	1	C9	CAP, 0.033µF, X7R, 25V, 10%, 0402	KEMET, C0402C333K3RACTU
5	1	C10	CAP, 6.8pF, COG/NP0, 50V, ±0.5pF, 0402	AVX, 04025A6R8DAT2A
6	1	C11	CAP, 330pF, X7R, 50V, 10%, 0402, AEC-Q200	TDK, CGA2B2X7R1H331K050BA
7	1	C19	CAP, 2.2µF, X7S, 10V, 10%, 0402, AEC-Q200	MURATA, GRT155C71A225KE13D
8	1	C24	CAP, 0.1µF, X7R, 16V, 10%, 0402, AEC-Q200	MURATA, GCM155R71C104KA55D
9	1	L1	IND., 0.08µH, PWR, 20%, 21.4A, 1.8mΩ, 4mm × 4mm, AEC-Q200	COILCRAFT, XEL4020-800MEC
10	1	R8	RES., 10k, 1%, 1/16W, 0402, AEC-Q200	VISHAY, CRCW040210K0FKED
11	1	U1	IC, LOW VOLTAGE SYNCH STEPDOWN REG, LQFN-18, PRELIM-PART NOT FOR PRODUCTION	ADI APPROVED SUPPLIER, LTC3311JV-1#PBF
<b>Additional Demo Board Circuit Components</b>				
1	2	C1, C18	CAP, 470µF, TANT, POSCAP, 6.3V, 20%, 7343, 10mΩ, TCF	PANASONIC, 6TCF470MAH
2	2	C7, C15	CAP, 0.1µF, X7R, 10V, 10%, 0402, AEC-Q200	MURATA, GCM155R71A104KA55D
3	1	C8	CAP, 22µF, X7S, 6.3V, 20%, 0805	TDK, C2012X7S0J226M125AC
4	4	C13, C14, C16, C17	CAP, 10µF, X7S, 6.3V, 20%, 0603	TDK, C1608X7S0J106M080AC
5	2	C25, C26	CAP, 0.22µF, X7R, 6.3V, 20%, 0603	JOHANSON DIELECTRICS, 6R3X14W224MV4T
6	1	C27	CAP, 2.2µF, X7S, 10V, 10%, 0402, AEC-Q200	MURATA, GRT155C71A225KE13D
7	1	L2	IND., 100Ω AT 100MHz, FERRITE BEAD, 25%, 8A, 6mΩ, 1812	WURTH ELEKTRONIK, 74279226101
8	1	Q1	XSTR., MOSFET, N-CH, 30V, 37A, PG-TDSON-8	INFINEON, BSC011N03LSI
9	1	R1	RES., 1M, 1%, 1/16W, 0402, AEC-Q200	STACKPOLE ELECTRONICS, INC., RMCF0402FT1M00
10	1	R2	RES., 249k, 1%, 1/16W, 0402, AEC-Q200	NIC, NRC04F2493TRF
11	2	R3, R9	RES., 100k, 5%, 1/16W, 0402, AEC-Q200	NIC, NRC04J104TRF
12	1	R4	RES., 0Ω, 1/16W, 0402, AEC-Q200	NIC, NRC04ZOTRF
13	1	R7	RES., 20Ω, 1%, 1/16W, 0402, AEC-Q200	NIC, NRC04F20R0TRF
14	1	R10	RES., 0.02Ω, 1%, 10W, 2818, HP, METAL, SENSE, AEC-Q200	VISHAY, WSHP2818R0200FEA
15	1	R11	RES., 10k, 5%, 1/16W, 0402, AEC-Q200	NIC, NRC04J103TRF
16	0	TP1	CONN., U.FL, RECEP, ST SMD, 0Hz TO 6GHz 50Ω	HIROSE ELECTRIC, U.FL-R-SMT-1(10)
<b>Hardware: For Demo Board Only</b>				
1	12	E1-E3, E5, E12, E14-E17, E20-E22	TEST POINT, TURRET, 0.064" MTG. HOLE, PCB 0.062" THK	MILL-MAX, 2308-2-00-80-00-00-07-0
2	5	E4, E7, E10, E13, E18	TEST POINT, TURRET, 0.094" MTG. HOLE, PCB 0.062" THK	MILL-MAX, 2501-2-00-80-00-00-07-0
3	5	E6, E8, E9, E11, E19	CONN., BANANA JACK, FEMALE, THT, NON-INSULATED, SWAGE, 0.218"	KEYSTONE, 575-4
4	1	JP1	CONN., HDR, MALE, 1 × 3, 2mm, VERT, ST, THT	WURTH ELEKTRONIK, 62000311121
5	4	MP1-MP4	STANDOFF, NYLON, SNAP-ON, 0.50"	KEYSTONE, 8833
6	1	XJP1	CONN., SHUNT, FEMALE, 2-POS, 2mm	WURTH ELEKTRONIK, 60800213421

# DEMO MANUAL DC3113A

## SCHEMATIC DIAGRAM



Information furnished by Analog Devices is believed to be accurate and reliable. However, no responsibility is assumed by Analog Devices for its use, nor for any infringements of patents or other rights of third parties that may result from its use. Specifications subject to change without notice. No license is granted by implication or otherwise under any patent or patent rights of Analog Devices.

# DEMO MANUAL DC3113A

---



## ESD Caution

**ESD (electrostatic discharge) sensitive device.** Charged devices and circuit boards can discharge without detection. Although this product features patented or proprietary protection circuitry, damage may occur on devices subjected to high energy ESD. Therefore, proper ESD precautions should be taken to avoid performance degradation or loss of functionality.

---

## Legal Terms and Conditions

By using the evaluation board discussed herein (together with any tools, components documentation or support materials, the "Evaluation Board"), you are agreeing to be bound by the terms and conditions set forth below ("Agreement") unless you have purchased the Evaluation Board, in which case the Analog Devices Standard Terms and Conditions of Sale shall govern. Do not use the Evaluation Board until you have read and agreed to the Agreement. Your use of the Evaluation Board shall signify your acceptance of the Agreement. This Agreement is made by and between you ("Customer") and Analog Devices, Inc. ("ADI"), with its principal place of business at One Technology Way, Norwood, MA 02062, USA. Subject to the terms and conditions of the Agreement, ADI hereby grants to Customer a free, limited, personal, temporary, non-exclusive, non-sublicensable, non-transferable license to use the Evaluation Board FOR EVALUATION PURPOSES ONLY. Customer understands and agrees that the Evaluation Board is provided for the sole and exclusive purpose referenced above, and agrees not to use the Evaluation Board for any other purpose. Furthermore, the license granted is expressly made subject to the following additional limitations: Customer shall not (i) rent, lease, display, sell, transfer, assign, sublicense, or distribute the Evaluation Board; and (ii) permit any Third Party to access the Evaluation Board. As used herein, the term "Third Party" includes any entity other than ADI, Customer, their employees, affiliates and in-house consultants. The Evaluation Board is NOT sold to Customer; all rights not expressly granted herein, including ownership of the Evaluation Board, are reserved by ADI. CONFIDENTIALITY. This Agreement and the Evaluation Board shall all be considered the confidential and proprietary information of ADI. Customer may not disclose or transfer any portion of the Evaluation Board to any other party for any reason. Upon discontinuation of use of the Evaluation Board or termination of this Agreement, Customer agrees to promptly return the Evaluation Board to ADI. ADDITIONAL RESTRICTIONS. Customer may not disassemble, decompile or reverse engineer chips on the Evaluation Board. Customer shall inform ADI of any occurred damages or any modifications or alterations it makes to the Evaluation Board, including but not limited to soldering or any other activity that affects the material content of the Evaluation Board. Modifications to the Evaluation Board must comply with applicable law, including but not limited to the RoHS Directive. TERMINATION. ADI may terminate this Agreement at any time upon giving written notice to Customer. Customer agrees to return to ADI the Evaluation Board at that time. LIMITATION OF LIABILITY. THE EVALUATION BOARD PROVIDED HEREUNDER IS PROVIDED "AS IS" AND ADI MAKES NO WARRANTIES OR REPRESENTATIONS OF ANY KIND WITH RESPECT TO IT. ADI SPECIFICALLY DISCLAIMS ANY REPRESENTATIONS, ENDORSEMENTS, GUARANTEES, OR WARRANTIES, EXPRESS OR IMPLIED, RELATED TO THE EVALUATION BOARD INCLUDING, BUT NOT LIMITED TO, THE IMPLIED WARRANTY OF MERCHANTABILITY, TITLE, FITNESS FOR A PARTICULAR PURPOSE OR NONINFRINGEMENT OF INTELLECTUAL PROPERTY RIGHTS. IN NO EVENT WILL ADI AND ITS LICENSORS BE LIABLE FOR ANY INCIDENTAL, SPECIAL, INDIRECT, OR CONSEQUENTIAL DAMAGES RESULTING FROM CUSTOMER'S POSSESSION OR USE OF THE EVALUATION BOARD, INCLUDING BUT NOT LIMITED TO LOST PROFITS, DELAY COSTS, LABOR COSTS OR LOSS OF GOODWILL. ADI'S TOTAL LIABILITY FROM ANY AND ALL CAUSES SHALL BE LIMITED TO THE AMOUNT OF ONE HUNDRED US DOLLARS (\$100.00). EXPORT. Customer agrees that it will not directly or indirectly export the Evaluation Board to another country, and that it will comply with all applicable United States federal laws and regulations relating to exports. GOVERNING LAW. This Agreement shall be governed by and construed in accordance with the substantive laws of the Commonwealth of Massachusetts (excluding conflict of law rules). Any legal action regarding this Agreement will be heard in the state or federal courts having jurisdiction in Suffolk County, Massachusetts, and Customer hereby submits to the personal jurisdiction and venue of such courts. The United Nations Convention on Contracts for the International Sale of Goods shall not apply to this Agreement and is expressly disclaimed.

Rev. 0

07/22

---

10

# Mouser Electronics

Authorized Distributor

Click to View Pricing, Inventory, Delivery & Lifecycle Information:

[Analog Devices Inc.:](#)

[DC3113A](#)